DS92LV1224

DS92LV1224 30-66 MHz 10 Bit Bus LVDS Deserializer



Literature Number: SNLS189



DS92LV1224

30-66 MHz 10 Bit Bus LVDS Deserializer

General Description

The DS92LV1224 is a 300 to 660 Mb/s deserializer for high-speed unidirectional serial data transmission over FR-4 printed circuit board backplanes and balanced copper cables. It receives the Bus LVDS serial data stream from a compatible 10-bit serializer, transforms it back into a 10-bit wide parallel data bus and recovers parallel clock. This single serial data stream simplifies PCB design and reduces PCB cost by narrowing data paths that in turn reduce PCB size and number of layers. The single serial data stream also reduces cable size, the number of connectors, and eliminates clock-to-data and data-to-data skew.

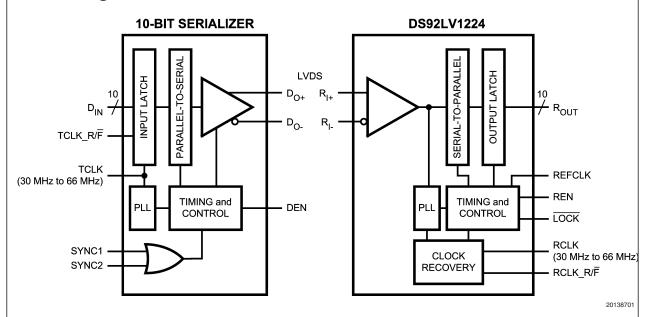
The DS92LV1224 works well with any National Semiconductor's Bus LVDS 10-bit serializer within its specified frequency operating range. It features low power consumption, and high impedance outputs in power down mode.

The DS92LV1224 was designed with the flow-through pinout and is available in a space saving 28-lead SSOP package.

Features

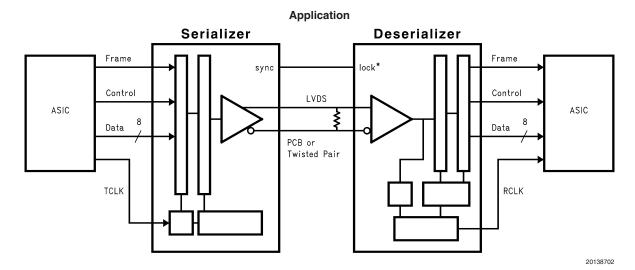
- 30-66 MHz Single 1:10 Deserializer with 300-660 Mb/s troughput
- Robust Bus LVDS serial data transmission with embedded clock with embedded clock for exceptional noise immunity and low EMI
- Clock recovery from PLL lock to random data patterns.
- Guaranteed transition every data transfer cycle
- Low power consumption < 300 mW (typ) @ 66 MHz
- Single differential pair eliminates multi-channel skew
- Flow-through pinout for easy PCB layout
- Synchronization mode and LOCK indicator
- Programmable edge trigger on clock
- High impedance on receiver inputs when power is off
- Small 28-lead SSOP package

Block Diagrams



TRI-STATE® is a registered trademark of National Semiconductor Corporation.

Block Diagrams (Continued)



Functional Description

The DS92LV1224 is a 10-bit Deserializer device which together with a compatible serializer (i.e. DS92LV1023E) forms a chipset designed to transmit data over FR-4 printed circuit board backplanes and balanced copper cables at clock speeds from 30 MHz to 66 MHz.

The chipset has three active states of operation: Initialization, Data Transfer, and Resynchronization; and two passive states: Powerdown and TRI-STATE®.

The following sections describe each operation and passive state.

Initialization

Initialization of both devices must occur before data transmission begins. Initialization refers to synchronization of the Serializer and Deserializer PLL's to local clocks, which may be the same or separate. Afterwards, synchronization of the Deserializer to Serializer occurs.

Step 1: When you apply $V_{\rm CC}$ to both Serializer and/or Deserializer, the respective outputs enter TRI-STATE®, and onchip power-on circuitry disables internal circuitry. When $V_{\rm CC}$ reaches $V_{\rm CC}$ OK (2.5V) the PLL in each device begins locking to a local clock. For the Serializer, the local clock is the transmit clock (TCLK) provided by the source ASIC or other device. For the Deserializer, you must apply a local clock to the REFCLK pin.

The Serializer outputs remain in TRI-STATE while the PLL locks to the TCLK. After locking to TCLK, the Serializer is now ready to send data or SYNC patterns, depending on the levels of the SYNC1 and SYNC2 inputs or a data stream. The SYNC pattern sent by the Serializer consists of six ones and six zeros switching at the input clock rate.

Note that the Deserializer $\overline{\text{LOCK}}$ output will remain high while its PLL locks to the incoming data or to SYNC patterns on the input.

Step 2: The Deserializer PLL must synchronize to the Serializer to complete initialization. The Deserializer will lock to non-repetitive data patterns. However, the transmission of SYNC patterns enables the Deserializer to lock to the Serializer signal within a specified time.

The user's application determines control of the SYNC1 and SYNC 2 pins. One recommendation is a direct feedback loop from the $\overline{\text{LOCK}}$ pin. Under all circumstances, the Serializer stops sending SYNC patterns after both SYNC inputs return low.

When the Deserializer detects edge transitions at the Bus LVDS input, it will attempt to lock to the embedded clock information. When the Deserializer locks to the Bus LVDS clock, the LOCK output will go low. When LOCK is low, the Deserializer outputs represent incoming Bus LVDS data.

Data Transfer

After initialization, the Serializer will accept data from inputs DIN0–DIN9. The Serializer uses the TCLK input to latch incoming Data. The TCLK_R/F pin selects which edge the Serializer uses to strobe incoming data. TCLK_R/F high selects the rising edge for clocking data and low selects the falling edge. If either of the SYNC inputs is high for 5*TCLK cycles, the data at DIN0-DIN9 is ignored regardless of clock edge.

After determining which clock edge to use, a start and stop bit, appended internally, frame the data bits in the register. The start bit is always high and the stop bit is always low. The start and stop bits function as the embedded clock bits in the serial stream.

The Serializer transmits serialized data and clock bits (10+2 bits) from the serial data output (DO \pm) at 12 times the TCLK frequency. For example, if TCLK is 66 MHz, the serial rate is 66 x 12 = 792 Mega-bits-per-second. Since only 10 bits are from input data, the serial "payload" rate is 10 times the TCLK frequency. For instance, if TCLK = 66 MHz, the payload data rate is 66 x 10 = 660 Mbps. The data source provides TCLK and must be in the range of 30 MHz to 66 MHz nominal.

The Serializer outputs (DO \pm) can drive a point-to-point connection or in limited multi-point or multi-drop backplanes. The outputs transmit data when the enable pin (DEN) is high, $\overline{PWRDN} = \text{high}$, and SYNC1 and SYNC2 are low. When DEN is driven low, the Serializer output pins will enter TRI-STATE.

When the Deserializer synchronizes to the Serializer, the $\overline{\text{LOCK}}$ pin is low. The Deserializer locks to the embedded

Data Transfer (Continued)

clock and uses it to recover the serialized data. ROUT data is valid when $\overline{\text{LOCK}}$ is low. Otherwise ROUT0-ROUT9 is invalid.

The ROUT0-ROUT9 pins use the RCLK pin as the reference to data. The polarity of the RCLK edge is controlled by the RCLK_R/F input. See *Figure 6*.

ROUT(0-9), LOCK and RCLK outputs will drive a maximum of three CMOS input gates (15 pF load) with a 66 MHz clock.

Resynchronization

When the Deserializer PLL locks to the embedded clock edge, the Deserializer $\overline{\text{LOCK}}$ pin asserts a low. If the Deserializer loses lock, the $\overline{\text{LOCK}}$ pin output will go high and the outputs (including RCLK) will enter TRI-STATE.

The user's system monitors the \overline{LOCK} pin to detect a loss of synchronization. Upon detection, the system can arrange to pulse the Serializer SYNC1 or SYNC2 pin to resynchronize. Multiple resynchronization approaches are possible. One recommendation is to provide a feedback loop using the \overline{LOCK} pin itself to control the sync request of the Serializer (SYNC1 or SYNC2). Dual SYNC pins are provided for multiple control in a multi-drop application. Sending sync patterns for resynchronization is desirable when lock times within a specific time are critical. However, the Deserializer can lock to random data, which is discussed in the next section.

Random Lock Initialization and Resynchronization

The initialization and resynchronization methods described in their respective sections are the fastest ways to establish the link between the Serializer and Deserializer. However. the DS92LV1224 can attain lock to a data stream without requiring the Serializer to send special SYNC patterns. This allows the DS92LV1224 to operate in "open-loop" applications. Equally important is the Deserializer's ability to support hot insertion into a running backplane. In the open loop or hot insertion case, we assume the data stream is essentially random. Therefore, because lock time varies due to data stream characteristics, we cannot possibly predict exact lock time. However, please see Table 1 for some general random lock times under specific conditions. The primary constraint on the "random" lock time is the initial phase relation between the incoming data and the REFCLK when the Deserializer powers up. As described in the next paragraph, the data contained in the data stream can also affect lock time. If a specific pattern is repetitive, the Deserializer could enter "false lock" - falsely recognizing the data pattern as the clocking bits. We refer to such a pattern as a repetitive multi-transition, RMT. This occurs when more than one Low-High transition takes place in a clock cycle over multiple cycles. This occurs when any bit, except DIN 9, is held at a low state and the adjacent bit is held high, creating a 0-1 transition. In the worst case, the Deserializer could become locked to the data pattern rather than the clock. Circuitry within the DS92LV1224 can detect that the possibility of "false lock" exists. The circuitry accomplishes this by detecting more than one potential position for clocking bits. Upon detection, the circuitry will prevent the LOCK output from becoming active until the potential "false lock" pattern changes. The false lock detect circuitry expects the data will eventually change, causing the Deserializer to lose lock to the data pattern and then continue searching for clock bits in the serial data stream. Graphical representations of RMT are shown in Figure 1. Please note that RMT only applies to bits DINO-DIN8.

Powerdown

When no data transfer occurs, you can use the Powerdown state. The Serializer and Deserializer use the Powerdown state, a low power sleep mode, to reduce power consumption. The Deserializer enters Powerdown when you drive PWRDN and REN low. The Serializer enters Powerdown when you drive PWRDN low. In Powerdown, the PLL stops and the outputs enterTRI-STATE, which disables load current and reduces supply current to the milliampere range. To exit Powerdown, you must drive the PWRDN pin high.

Before valid data exchanges between the Serializer and Deserializer, you must reinitialize and resynchronize the devices to each other. Initialization of the Serializer takes 510 TCLK cycles. The Deserializer will initialize and assert LOCK high until lock to the Bus LVDS clock occurs.

TRI-STATE

The Serializer enters TRI-STATE when the DEN pin is driven low. This puts both driver output pins (DO+ and DO-) into TRI-STATE. When you drive DEN high, the Serializer returns to the previous state, as long as all other control pins remain static (SYNC1, SYNC2, PWRDN, TCLK_R/F).

When you drive the REN pin low, the Deserializer enters TRI-STATE. Consequently, the receiver output pins (ROUT0-ROUT9) and RCLK will enter TRI-STATE. The LOCK output remains active, reflecting the state of the PLL.

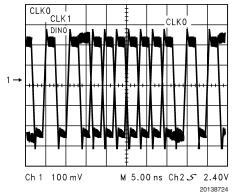
TABLE 1.

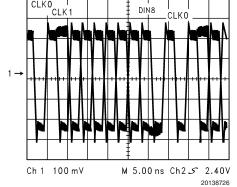
Random Lock Times for the DS92LV1224						
	40 MHz 66 MHz Units					
Maximum	26	18	μs			
Mean	4.5	3.0	μs			
Minimum	0.77	0.43	μs			
Conditions:	PRBS 2 ¹⁵ , V _{CC} = 3.3V					

¹⁾ Difference in lock times are due to different starting points in the data pattern with multiple parts.

Ordering Information

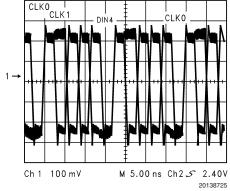
NSID	Function	Package
DS92LV1224TMSA	Deserializer	MSA28





DIN0 Held Low-DIN1 Held High Creates an RMT Pattern

DIN8 Held Low-DIN9 Held High Creates an RMT Pattern



DIN4 Held Low-DIN5 Held High Creates an RMT Pattern

FIGURE 1. RMT Patterns Seen on the Bus LVDS Serial Output

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage (V_{CC}) -0.3V to +4V

LVCMOS/LVTTL Input

Voltage -0.3V to $(V_{CC} + 0.3V)$

LVCMOS/LVTTL Output

Voltage -0.3V to $(V_{CC} + 0.3V)$

Bus LVDS Receiver Input

Voltage -0.3V to +3.9VJunction Temperature $+150^{\circ}C$

Storage Temperature -65°C to +150°C

Lead Temperature

(Soldering, 4 seconds) +260°C
Maximum Package Power Dissipation Capacity

@ 25°C Package:

28L SSOP 1.27 W

Package Derating:

 $\begin{array}{c} & 10.3 \text{ mW/}^{\circ}\text{C above} \\ 28\text{L SSOP} & +25^{\circ}\text{C} \\ \theta_{ja} & 97^{\circ}\text{C/W} \\ \theta_{jc} & 27^{\circ}\text{C/W} \end{array}$

ESD Rating

HBM (1.5kOhm, 100pF) >2kV

MM > 250V

Recommended Operating Conditions

	Min	Nom	Max	Units	
Supply Voltage (V_{CC})	3.0	3.3	3.6	V	
Operating Free Air Temperature (T _A)	-40	+25	+85	°C	
Receiver Input Range	0		2.4	V	
Supply Noise Voltage (V _{CC})			100	mV _{P-P}	

Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions			Тур	Max	Units
DESERIALIZ	ER LVCMOS/LVTTL DC SPECIFICAT	TONS (apply to pins PWR	DN, RCLK_R/F, F	EN, RE	FCLK =	inputs;	apply
to pins ROU	T, RCLK, \overline{LOCK} = outputs)						
V _{IH}	High Level Input Voltage			2.0		V _{CC}	V
V _{IL}	Low Level Input Voltage			GND		0.8	V
V _{CL}	Input Clamp Voltage	$I_{CL} = -18 \text{ mA}$			-0.62	-1.5	V
I _{IN}	Input Current	V _{IN} = 0V or 3.6V		-10	±1	+15	μA
V _{OH}	High Level Output Voltage	$I_{OH} = -9 \text{ mA}$		2.2	3.0	V _{CC}	V
V _{OL}	Low Level Output Voltage	I _{OL} = 9 mA		GND	0.25	0.5	V
I _{os}	Output Short Circuit Current	VOUT = 0V		-15	-47	-85	mA
I _{oz}	TRI-STATE Output Current	PWRDN or REN = 0.8V, V	OUT = 0V or VCC	-10	±0.1	+10	μA
DESERIALIZ	ER Bus LVDS DC SPECIFICATIONS	(apply to pins RI+ and RI-	-)				
VTH	Differential Threshold High Voltage	VCM = +1.1V			+6	+50	mV
VTL	Differential Threshold Low Voltage	V CIVI = +1.1 V		-50	-12		mV
	Innut Cumput	$V_{IN} = +2.4V, V_{CC} = 3.6V o$	r 0V	-10	±1	+15	μA
I _{IN}	Input Current	$V_{IN} = 0V, V_{CC} = 3.6V \text{ or } 0$	V	-10	±0.05	+10	μΑ
DESERIALIZ	ER SUPPLY CURRENT (apply to pin	s DVCC and AVCC)					
	Deserializer Supply Current Worst	0 15 = 5	f = 30 MHz		58	75	mA
I _{CCR}	Case	$C_L = 15 \text{ pF}$	f = 40 MHz		58	75	mA
		Figure 2	f = 66 MHz		90	110	mA
I _{CCXR}	Deserializer Supply Current	PWRDN = 0.8V, REN = 0.8V			0.36	1.0	mA
	Powerdown						

Deserializer Timing Requirements for REFCLKOver recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
t _{RFCP}	REFCLK Period		15.15	Т	33.33	ns
t _{RFDC}	REFCLK Duty Cycle		30	50	70	%
t _{RFCP} /	Ratio of REFCLK to		95	1	105	
t _{TCP}	TCLK		95	1	103	
t _{RFTT}	REFCLK Transition Time			3	6	ns

Deserializer Switching Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Тур	Max	Units
t _{RCP}	Receiver out Clock Period	$t_{RCP} = t_{TCP}$	RCLK	15.15		33.33	ns
t _{CLH}	CMOS/TTL Low-to-High Transition Time	CL = 15 pF	Rout(0-9), LOCK, RCLK		1.2	4	ns
t _{CHL}	CMOS/TTL High-to-Low Transition Time	Figure 3			1.1	4	ns
		All Temp./ All Freq.		1.75*t _{RCP} +1.25	1.75*t _{RCP} +3.75	1.75*t _{RCP} +6.25	ns
	Deserializer Delay	Room Temp./ 3.3V/30MHz		1.75*t _{RCP} +2.25	1.75*t _{RCP} +3.75	1.75*t _{RCP} +5.25	ns
t _{DD}	Figure 5	Room Temp./ 3.3V/40MHz		1.75*t _{RCP} +2.25	1.75*t _{RCP} +3.75	1.75*t _{RCP} +5.25	ns
		Room Temp./ 3.3V/66MHz		1.75*t _{RCP} +2.75	1.75*t _{RCP} +3.75	1.75*t _{RCP} +4.75	ns
			RCLK 30MHz	0.4*t _{RCP}	0.5*t _{RCP}		ns
t _{ROS}	ROUT Data Valid before RCLK	Figure 6	RCLK 40MHz	0.4*t _{RCP}	0.5*t _{RCP}		ns
			RCLK 66MHz	0.38*t _{RCP}	0.5*t _{RCP}		ns
+	ROUT Data valid after		30MHz	-0.4*t _{RCP}	-0.5*t _{RCP}		ns
t _{ROH}	RCLK	Figure 6	40MHz	-0.4*t _{RCP}	-0.5*t _{RCP}		ns
			66MHz	-0.38*t _{RCP}	-0.5*t _{RCP}		ns
t _{RDC}	RCLK Duty Cycle			45	50	55	%
t _{HZR}	HIGH to TRI-STATE Delay				2.8	10	ns
t _{LZR}	LOW to TRI-STATE Delay	Figure 7	Dout(0.0)		2.8	10	ns
t _{zhr}	TRI-STATE to HIGH Delay	rigule 7	Rout(0-9)		4.2	10	ns
t _{ZLR}	TRI-STATE to LOW Delay				4.2	10	ns
	Deserializer PLL Lock		30MHz		1.68	3	μs
t_{DSR1}	time from PWRDWN		40MHz		1.31	3	μs
	(with SYNCPAT)		66MHz		0.84	3	μs

Descrializer Switching Characteristics (Continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Тур	Max	Units
Deserializer PLL Lock time from SYNCPAT	Decerializer DLL Lock		30MHz		0.62	1	μs
		40MHz		0.47	1	μs	
	unie nom STNOPAT		66MHz		0.29	0.8	μs
t _{ZHLK}	TRI-STATE to HIGH Delay (power-up)		LOCK		3.7	12	ns
	Deserializer Noise	Figure 10	30 MHz	650	950		ps
t _{RNM}	Margin	Figure 10 (Note 7)	40 MHz	450	730		ps
	Waigiii	(Note 7)	66 MHz	250	400		ps

Note 1: "Absolute Maximum Ratings" are those values beyond which the safety of the device cannot be guaranteed. They are not meant to imply that the devices should be operated at these limits. The table of "Electrical Characteristics" specifies conditions of device operation.

Note 2: Typical values are given for $V_{CC} = 3.3V$ and $T_A = +25^{\circ}C$.

Note 3: Current into device pins is defined as positive. Current out of device pins is defined as negative. Voltages are referenced to ground except VOD, ΔVOD , VTH and VTL which are differential voltages.

Note 4: t_{LLHT} and t_{LHLT} specifications are Guranteed By Design (GBD) using statistical analysis.

Note 5: Because the Serializer is in TRI-STATE mode, the Deserializer will lose PLL lock and have to resynchronize before data transfer.

Note 6: For the purpose of specifying deserializer PLL performance, tDSR1 and tDSR2 are specified with the REFCLK running and stable, and with

specific conditions for the incoming data stream (SYNCPATs). It is recommended that the derserializer be initialized using either $t_{\rm DSR1}$ timing or $t_{\rm DSR2}$ timing. $t_{\rm DSR1}$ is the time required for the deserializer to indicate lock upon power-up or when leaving the power-down mode. Synchronization patterns should be sent to the device before initiating either condition. $t_{\rm DSR2}$ is the time required to indicate lock for the powered-up and enabled deserializer when the input (RI+ and RI-) conditions change from not receiving data to receiving synchronization patterns (SYNCPATs).

Note 7: t_{RNM} is a measure of how much phase noise (jitter) the deserializer can tolerate in the incoming data stream before bit errors occur. The Deserializer Noise Margin is Guaranteed By Design (GBD) using statistical analysis.

AC Timing Diagrams and Test Circuits

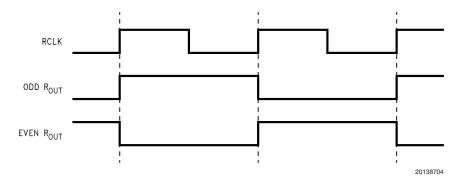


FIGURE 2. "Worst Case" Deserializer ICC Test Pattern

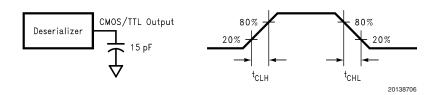


FIGURE 3. Deserializer CMOS/TTL Output Load and Transition Times

AC Timing Diagrams and Test Circuits (Continued)

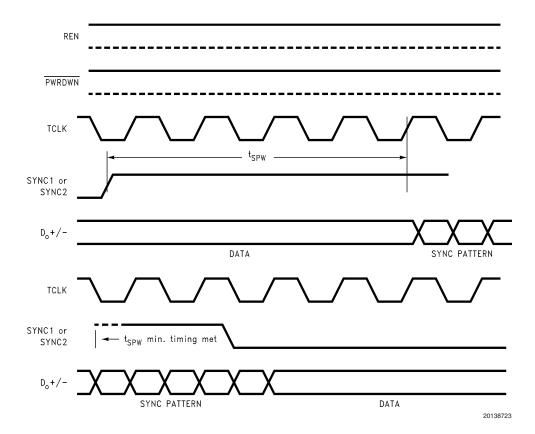


FIGURE 4. SYNC Timing Delays

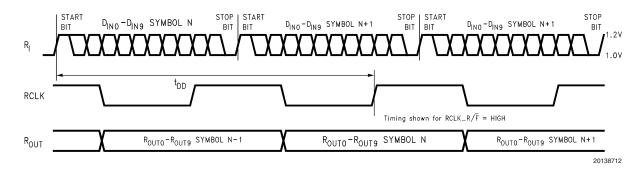
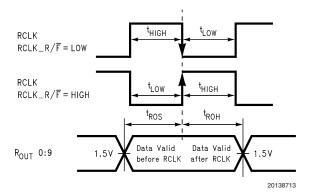


FIGURE 5. Deserializer Delay

AC Timing Diagrams and Test Circuits (Continued)



Timing shown for RCLK_R/ \overline{F} = LOW Duty Cycle (t_{RDC}) = $\frac{t_{HIGH}}{t_{HIGH} + t_{LOW}}$

FIGURE 6. Deserializer Data Valid Out Times

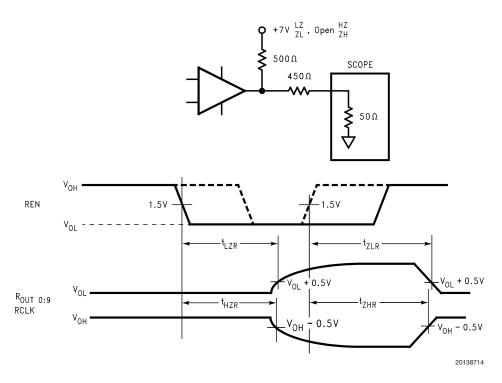
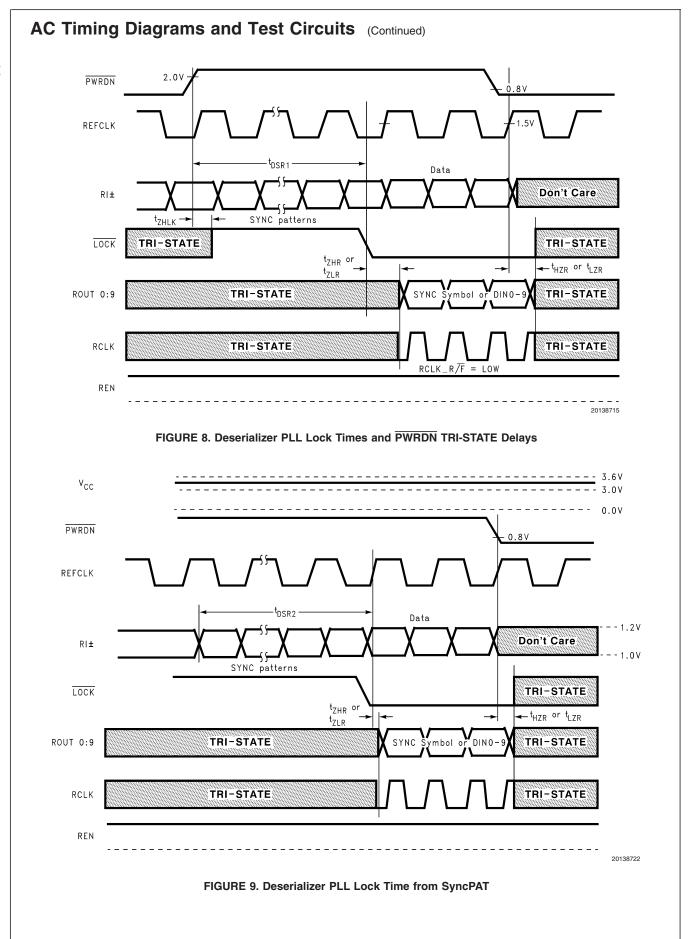
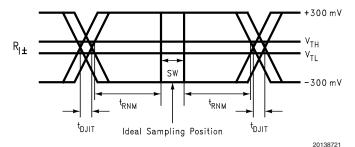


FIGURE 7. Deserializer TRI-STATE Test Circuit and Timing



AC Timing Diagrams and Test Circuits (Continued)



SW - Setup and Hold Time (Internal Data Sampling Window) t_{DJIT} - Serializer Output Bit Position Jitter that results from Jitter on TCLK

 t_{RNM} = Receiver Noise Margin Time

FIGURE 10. Receiver Bus LVDS Input Skew Margin

Deserializer Truth Table

INPUTS				
PWRDN	REN	ROUT [0:9]	LOCK	RCLK
Н	Н	Z	Н	Z
Н	Н	Active	L	Active
L	Х	Z	Z	Z
Н	L	Z	Active	Z

¹⁾ LOCK Active indicates the LOCK output will reflect the state of the Deserializer with regard to the selected data stream.
2) RCLK Active indicates the RCLK will be running if the Deserializer is locked. The Timing of RCLK with respect to ROUT is determined by RCLK_R/F.
3) ROUT and RCLK are TRI-STATED when LOCK is asserted High.

Application Information

USING THE SERIALIZER AND DESERIALIZER CHIPSET

The Serializer and Deserializer chipset is an easy to use transmitter and receiver pair that sends 10 bits of parallel LVTTL data over a serial Bus LVDS link up to 660 Mbps. An on-board PLL serializes the input data and embeds two clock bits within the data stream. The Deserializer uses a separate reference clock (REFCLK) and an onboard PLL to extract the clock information from the incoming data stream and then deserialize the data. The Deserializer monitors the incoming clock information, determines lock status, and asserts the LOCK output high when loss of lock occurs.

POWER CONSIDERATIONS

An all CMOS design of the Serializer and Deserializer makes them inherently low power devices. In addition, the constant current source nature of the Bus LVDS outputs minimizes the slope of the speed vs. $\rm I_{CC}$ curve of conventional CMOS designs.

POWERING UP THE DESERIALIZER

The DS92LV1224 can be powered up at any time by following the proper sequence. The REFCLK input can be running before the Deserializer powers up, and it must be running in order for the Deserializer to lock to incoming data. The Deserializer outputs will remain in TRI-STATE until the Deserializer detects data transmission at its inputs and locks to the incoming data stream.

TRANSMITTING DATA

Once you power up the Serializer and Deserializer, they must be phase locked to each other to transmit data. Phase locking occurs when the Deserializer locks to incoming data or when the Serializer sends patterns. The Serializer sends SYNC patterns whenever the SYNC1 or SYNC2 inputs are high. The LOCK output of the Deserializer remains high until it has locked to the incoming data stream. Connecting the LOCK output of the Deserializer to one of the SYNC inputs of the Serializer will guarantee that enough SYNC patterns are sent to achieve Deserializer lock.

The Deserializer can also lock to incoming data by simply powering up the device and allowing the "random lock" circuitry to find and lock to the data stream.

While the Deserializer LOCK output is low, data at the Deserializer outputs (ROUT0-9) is valid, except for the specific case of loss of lock during transmission which is further discussed in the "Recovering from LOCK Loss" section below

NOISE MARGIN

The Deserializer noise margin is the amount of input jitter (phase noise) that the Deserializer can tolerate and still reliably receive data. Various environmental and systematic factors include:

Serializer: TCLK jitter, V_{CC} noise (noise bandwidth and out-of-band noise)

Media: ISI, Large V_{CM} shifts Deserializer: V_{CC} noise

RECOVERING FROM LOCK LOSS

In the case where the Deserializer loses lock during data transmission, up to 3 cycles of data that were previously received can be invalid. This is due to the delay in the lock detection circuit. The lock detect circuit requires that invalid

clock information be received 4 times in a row to indicate loss of lock. Since clock information has been lost, it is possible that data was also lost during these cycles. Therefore, after the Deserializer relocks to the incoming data stream and the Deserializer $\overline{\text{LOCK}}$ pin goes low, at least three previous data cycles should be suspect for bit errors.

The Deserializer can relock to the incoming data stream by making the Serializer resend SYNC patterns, as described above, or by random locking, which can take more time, depending on the data patterns being received.

HOT INSERTION

All the BLVDS devices are hot pluggable if you follow a few rules. When inserting, ensure the Ground pin(s) makes contact first, then the VCC pin(s), and then the I/O pins. When removing, the I/O pins should be unplugged first, then the VCC, then the Ground. Random lock hot insertion is illustrated in Figure 13

PCB CONSIDERATIONS

The Bus LVDS Serializer and Deserializer should be placed as close to the edge connector as possible. In multiple Deserializer applications, the distance from the Deserializer to the slot connector appears as a stub to the Serializer driving the backplane traces. Longer stubs lower the impedance of the bus, increase the load on the Serializer, and lower the threshold margin at the Deserializers. Deserializer devices should be placed much less than one inch from slot connectors. Because transition times are very fast on the Serializer Bus LVDS outputs, reducing stub lengths as much as possible is the best method to ensure signal integrity.

TRANSMISSION MEDIA

The Serializer and Deserializer can also be used in point-to-point configuration of a backplane, through a PCB trace, or through twisted pair cable. In point-to-point configuration, the transmission media need only be terminated at the receiver end. Please note that in point-to-point configuration, the potential of offsetting the ground levels of the Serializer vs. the Deserializer must be considered. Also, Bus LVDS provides a +/- 1.2V common mode range at the receiver inputs.

Failsafe Biasing for the DS92LV1224

The DS92LV1224 has an improved input threshold sensitivity of +/- 50mV versus +/- 100mV for the DS92LV1210 or DS92LV1212. This allows for greater differential noise margin in the DS92LV1224. However, in cases where the receiver input is not being actively driven, the increased sensitivity of the DS92LV1224 can pickup noise as a signal and cause unintentional locking . For example, this can occur when the input cable is disconnected.

External resistors can be added to the receiver circuit board to prevent noise pick-up. Typically, the non-inverting receiver input is pulled up and the inverting receiver input is pulled down by high value resistors. the pull-up and pull-down resistors (R₁ and R₂) provide a current path through the termination resistor (R_L) which biases the receiver inputs when they are not connected to an active driver. The value of the pull-up and pull-down resistors should be chosen so that enough current is drawn to provide a +15mV drop across the termination resistor. Please see *Figure 11* for the Failsafe Biasing Setup.

Application Information (Continued)

USING T_{DJIT} and T_{RNM} to validate signal quality

The parameters $t_{\rm DJIT}$ and $t_{\rm RNM}$ can be used to generate an eye pattern mask to validate signal quality in an actual application or in simulation.

The parameter t_{DJIT} measures the transmitter's ability to place data bits in the ideal position to be sampled by the receiver. The typical t_{DJIT} parameter of -80 ps indicates that the crossing point of the Tx data is 80 ps ahead of the ideal crossing point. The $t_{DJIT(min)}$ and $t_{DJIT(max)}$ parameters specify the earliest and latest, respectively, time that a crossing will occur relative to the ideal position.

The parameter $t_{\sf RNM}$ is calculated by first measuring how much of the ideal bit the receiver needs to ensure correct sampling. After determining this amount, what remains of the ideal bit that is available for external sources of noise is called $t_{\sf RNM}$. It is the offset from $t_{\sf DJIT(min\ or\ max)}$ for the test mask within the eye opening.

The vertical limits of the mask are determined by the DS92LV1224 receiver input threshold of +/- 50 mV.

Please refer to the eye mask pattern of Figure 11 for a graphic representation of $t_{\rm DJIT}$ and $t_{\rm RNM}.$

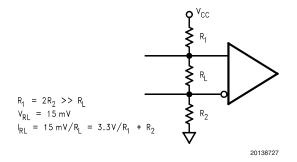


FIGURE 11. Failsafe Biasing Setup

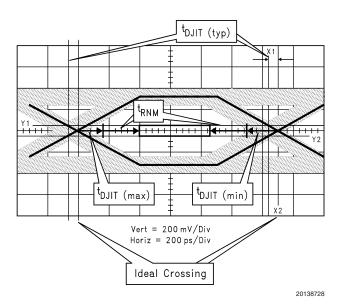


FIGURE 12. Using t_{DJIT} and t_{RNM} to Generate an Eye Pattern Mask and Validate Signal Quality

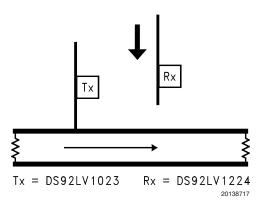
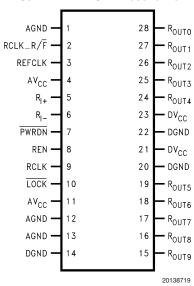


FIGURE 13. Random Lock Hot Insertion

Pin Diagrams

DS92LV1224TMSA - Deserializer



Description

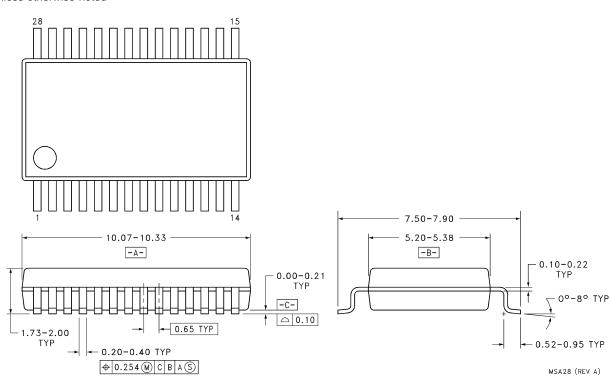
Pin Name	I/O	No.	Description
ROUT	0	15–19, 24–28	Data Output. ±9 mA CMOS level outputs.
RCLK_R/F	I	2	Recovered Clock Rising/Falling strobe select. TTL level input.
			Selects RCLK active edge for strobing of ROUT data. High selects
			rising edge. Low selects falling edge.
RI+	I	5	+ Serial Data Input. Non-inverting Bus LVDS differential input.
RI-	I	6	- Serial Data Input. Inverting Bus LVDS differential input.
PWRDN	I	7	Powerdown. TTL level input. PWRDN driven low shuts down the PLL
			and TRI-STATEs outputs putting the device into a low power sleep
			mode.
LOCK	0	10	LOCK goes low when the Deserializer PLL locks onto the embedded
			clock edge. CMOS level output. Totem pole output structure, does
			not directly support wire OR connection.
RCLK	0	9	Recovered Clock. Parallel data rate clock recovered from embedded
			clock. Used to strobe ROUT, CMOS level output.
REN	I	8	Output Enable. TTL level input. TRI-STATEs ROUT0-ROUT9, LOCK
			and RCLK when driven low.

Description (Continued)

Pin Name	I/O	No.	Description
DVCC	I	21, 23	Digital Circuit power supply.
DGND	I	14, 20, 22	Digital Circuit ground.
AVCC	I	4, 11	Analog power supply (PLL and Analog Circuits).
AGND	I	1, 12, 13	Analog ground (PLL and Analog Circuits).
REFCLK	I	3	Use this pin to supply a REFCLK signal for the internal PLL
			frequency.

Physical Dimensions inches (millimeters)

unless otherwise noted



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