

# STEP-UP DC/DC CONVERTER FOR WHITE LED BACK LIGHT

NO.EA-297-170905

# **OUTLINE**

The R1206N071B is PWM control type step-up DC/DC converter IC with low supply current.

The R1206N071B is fully dedicated to drive White LEDs with constant current. Each of these ICs consists of an NMOS FET, an oscillator, a PWM comparator, a voltage reference unit, an error amplifier, a current limit circuit, an under voltage lockout circuit (UVLO), and an over-voltage protection circuit (OVP).

The R1206N071B can drive white LEDs in constant current with high efficiency by using an inductor, a diode, a resistor and capacitors as external components.

The LEDs current can be set by an external resistance value and can adjust the dimming of LEDs by CE pin according to the signal of PWM. Feedback voltage is 0.2V, therefore power loss by current setting resistance is small and efficiency is good. Maximum duty cycle is internally fixed, Typ. 91%. LEDs can be driven from low voltage. Protection circuits are the current limit of Lx peak current, the over voltage limit of output, and the under voltage lockout function.

It is controllable the dimming of LEDs quickly when the PWM signal (between 200Hz to 300kHz) input to CE pin. If the CE pin input is "L" in the fixed time (Typ. 0.5ms), the IC becomes the standby mode and turns OFF LEDs.

# **FEATURES**

<ul> <li>Supply Current</li> <li>Standby Current</li> <li>Input Voltage Range</li> <li>Feedback Voltage</li> </ul>	Max. 5μA 1.8V to 5.5V
Feedback Voltage Accuracy	
• Temperature-Drift Coefficient of Feedback Voltage	±150ppm/°C
Oscillator Frequency	Typ. 1.2MHz
Maximum Duty Cycle	Typ. 91%
Switch ON Resistance	Typ. 1.35Ω
UVLO Detector Threshold	Typ. 1.6V
Lx Current Limit Protection	Typ. 700mA
OVP Detector Threshold	Typ. 29.5V
Switching Control	PWM
LED dimming control	by external PWM signal (Frequency 200Hz to 300kHz)
Packages	SOT-23-6
Ceramic capacitors are recommended	0.22μF

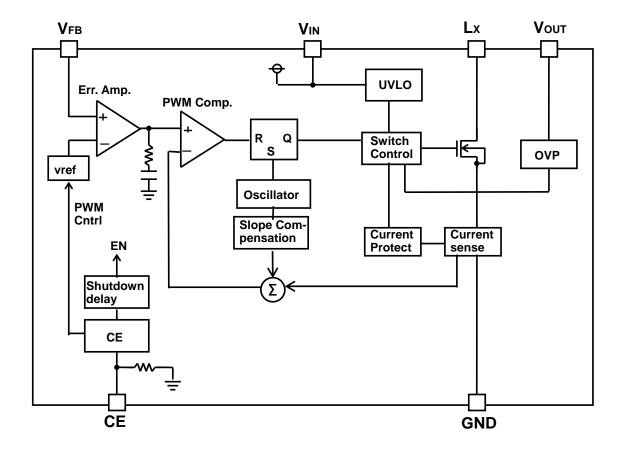
# **APPLICATIONS**

• White LED Backlight for portable equipment

# **SELECTION GUIDE**

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
R1206N071B-TR-FE	SOT-23-6	3,000 pcs	Yes	Yes

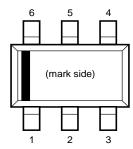
# **BLOCK DIAGRAM**



NO.EA-297-170905

# **PIN DESCRIPTIONS**





# • SOT-23-6

Pin No	Symbol	Pin Description
1	Lx	Switching Pin (Open Drain Output)
2	GND	Ground Pin
3	V <sub>FB</sub>	Feedback Pin
4	CE	Chip Enable Pin ("H" Active)
5	Vouт	Output Pin
6	Vin	Input Pin

#### R1206N071B

NO.EA-297-170905

# **ABSOLUTE MAXIMUM RATINGS**

(GND=0V)

Symbol	Item		Rating	Unit
Vin	V <sub>IN</sub> Pin Voltage		-0.3 to 6.5	V
Vce	CE Pin Voltage		-0.3 to V <sub>IN</sub> +0.3	V
V <sub>FB</sub>	V <sub>FB</sub> Pin Voltage		-0.3 to V <sub>IN</sub> +0.3	V
Vouт	Vоит Pin Voltage		-0.3 to 32	V
V <sub>L</sub> X	Lx Pin Voltage		-0.3 to 32	V
ILX	Lx Pin Current		1000	mA
PD	Power Dissipation * SOT-23-6 (JEDEC STD. 51-7 Test Land Pattern)		660	mW
Tj	Junction Temperature Range		-40 to 125	°C
Tstg	Storage Temperature Range		-55 to 125	°C

<sup>\*)</sup> Refer to POWER DISSIPATION for detailed information.

## **ABSOLUTE MAXIMUM RATINGS**

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

# RECOMMENDED OPERATING CONDITIONS

Symbol	Item	Rating	Unit
V <sub>IN</sub>	Operating Input Voltage	1.8 to 5.5	V
Та	Operating Temperature Range	-40 to 85	°C

#### RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.



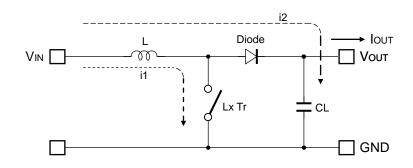
# **ELECTRICAL CHARACTERISTICS**

• R1206N071B (Ta=25°C)

Symbol	Item	Conditions	Min.	Тур.	Max.	Unit
IDD	Supply Current	V <sub>IN</sub> =5.5V, V <sub>FB</sub> =0V, Lx at no load		0.5	1.0	mA
Istandby	Standby Current	VIN=5.5V, VCE=0V		1.0	5.0	μΑ
V <sub>UVLO1</sub>	UVLO Detector Threshold	V <sub>IN</sub> falling	1.5	1.6	1.7	V
V <sub>UVLO2</sub>	UVLO Released Voltage	V <sub>IN</sub> rising		Vuvlo1 +0.1	1.8	V
Vсен	CE Input Voltage "H"	V <sub>IN</sub> =5.5V	1.5			V
Vcel	CE Input Voltage "L"	V <sub>IN</sub> =1.8V			0.5	V
Rce	CE Pull Down Resistance	V <sub>IN</sub> =3.6V	600	1200	2200	kΩ
V <sub>FB</sub>	V <sub>FB</sub> Voltage Accuracy	VIN=VCE=3.6V	0.19	0.20	0.21	V
ΔV <sub>FB</sub> / ΔTa	V <sub>FB</sub> Voltage Temperature Coefficient	$V_{\text{IN}}=V_{\text{CE}}=3.6V$ , $-40^{\circ}C \leq T_{a} \leq 85^{\circ}C$		±150		ppm /°C
lгв	V <sub>FB</sub> Input Current	VIN=5.5V, VFB=0V or VIN	-0.1		0.1	μΑ
Ron	Switch ON Resistance	VIN=3.6V, ILX=100mA		1.35		Ω
<b>I</b> Lxleak	Switch Leakage Current	V <sub>L</sub> x=30V		0	3.0	μΑ
l⊥xlim	Switch Current Limit	VIN=3.6V	400	700	1000	mA
fosc	Oscillator Frequency	VIN=3.6V, VOUT=VFB=0V	1.0	1.2	1.4	MHz
Maxduty	Maximum Duty Cycle	VIN=3.6V, VOUT=VFB=0V	86	91		%
V <sub>OVP1</sub>	OVP Detector Threshold	VIN=3.6V, VOUT rising	28.7	29.5	30.3	V
ΔV <sub>OVP1</sub> / ΔTa	V <sub>OVP1</sub> Voltage Temperature Coefficient	$V_{\text{IN}}=V_{\text{CE}}=3.6V$ , $-40^{\circ}C \leq T_{\text{a}} \leq 85^{\circ}C$		±150		ppm /°C
V <sub>OVP2</sub>	OVP Released Voltage	VIN=3.6V, VOUT falling		V <sub>OVP1</sub> -1.55		V

# THEORY OF OPERATION

# Operation of Step-Up DC/DC Converter and Output Current <Basic Circuit>

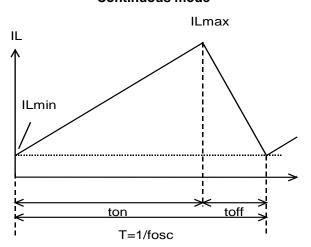


<Current through L>

#### Discontinuous mode

# ILmax ILmax topen ton T=1/fosc

### Continuous mode



There are two operation modes of the step-up PWM control-DC/DC converter. That is the continuous mode and discontinuous mode by the continuousness inductor.

When the transistor turns ON, the voltage of inductor L becomes equal to  $V_{IN}$  voltage. The increase value of inductor current (i1) will be

$$\Delta i1 = V_{IN} \times ton / L$$
 Formula 1

As the step-up circuit, during the OFF time (when the transistor turns OFF) the voltage is continually supply from the power supply. The decrease value of inductor current (i2) will be

$$\Delta i2 = (V_{OUT} - V_{IN}) \times t_{OPEN} / L_{IN}$$
 Formula 2

At the PWM control-method, the inductor current become continuously when topen=toff, the DC/DC converter

operate as the continuous mode.

In the continuous mode, the variation of current of i1 and i2 is same at regular condition.

$$V_{IN} \times ton / L = (V_{OUT} - V_{IN}) \times toff / L$$
 Formula 3

The duty at continuous mode will be

The average of inductor current at tf = toff will be

$$IL(Ave.) = V_{IN} \times ton / (2 \times L)$$
 Formula 5

If the input voltage = output voltage, the lout will be

$$I_{OUT} = V_{IN}^2 \times ton / (2 \times L \times V_{OUT})$$
......Formula 6

If the I<sub>OUT</sub> value is large than above the calculated value (Formula 6), it will become the continuous mode, at this status, the peak current (IL<sub>max</sub>) of inductor will be

The peak current value is larger than the lout value. In case of this, selecting the condition of the input and the output and the external components by considering of ILmax value.

The explanation above is based on the ideal calculation, and the loss caused by Lx switch and the external components are not included.

The actual maximum output current will be between 50% and 80% by the above calculations. Especially, when the IL is large or V<sub>IN</sub> is low, the loss of V<sub>IN</sub> is generated with on resistance of the switch. Moreover, it is necessary to consider Vf of the diode (approximately 0.8V) about V<sub>OUT</sub>.

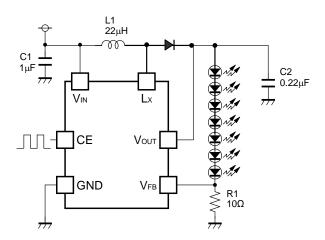
#### Soft-Start

The output of the error amplifier starts from 0V and the inrush current is suppressed when starting by the CE pin "H" input.

Moreover, the inrush current can be suppressed by gradually enlarging Duty of the PWM signal to the CE pin.

# **APPLICATION INFORMATION**

# • Typical Application Circuit



C1	CM105B105K06
C2	GRM21BR71H224
L1	LQH32CN220K53

#### Selection of Inductors

The peak current of the inductor at normal mode can be estimated as the next formula when the efficiency is 80%.

$$ILmax = 1.25 \times I_{OUT} \times V_{OUT} / V_{IN} + 0.5 \times V_{IN} \times (V_{OUT} - V_{IN}) / (L \times V_{OUT} \times fosc)$$

When the start-up or dimming control by CE pin, transient current flows, the peak current must be equal or less than the current limit of the IC. The peak current should not beyond the rating current of the inductor.

When 4-7 LEDs are driven with  $V_{IN}$ =3.6V, the recommended inductance value is  $10\mu H$  -22 $\mu H$ .

Table 1 Peak current value in each condition

Condition				
VIN (V)	Vout (V)	lout (mA)	L (μH)	ILmax (mA)
3	14	20	10	215
3	14	20	22	160
3	21	20	10	280
3	21	20	22	225

**Table 2 Recommended inductors** 

L	Part No.	Rated	Size
(μH)	i ait ivo.	Current (mA)	(mm)
10	LQH32CN100K53	450	3.2x2.5x1.55
10	LQH2MC100K02	225	2.0x1.6x0.9
10	VLF3010A-100	490	2.8x2.6x0.9
10	VLS252010-100	520	2.5x2.0x1.0
22	LQH32CN220K53	250	3.2x2.5x1.55
22	LQH2MC220K02	185	2.0x1.6x0.9
22	VLF3010A-220	330	2.8x2.6x0.9

#### Selection of Capacitors

Set  $1\mu F$  or more value bypass capacitor C1 between VIN pin and GND pin as close as possible. Set  $0.22\mu F$  or more capacitor C2 between VOUT pin and GND pin.

Note the V<sub>OUT</sub> that depends on LED used, and select the rating of V<sub>OUT</sub> or more.

#### Selection of SBD (Schottky Barrier Diode )

Select the diode with low V<sub>F</sub> such as Schottky type with low reverse current I<sub>R</sub>, and with low capacitance.

 Rated voltage(V)
 Part No.

 C1
 6.3
 CM105B105K06

 C2
 25
 GRM21BR11E224

 50
 GRM21BR71H224

 D1
 30
 CRS10I30A

 30
 RSX051VA-30

**Table 3 Recommended components for R1206** 

## • LED Current setting

When CE pin input is "H" (Duty=100%), LED current can be set with feedback resistor (R1)

$$I_{LED} = V_{FB} / R1$$

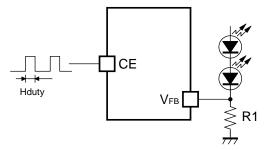
#### LED Dimming Control

The LED brightness can be controlled by inputting the PWM signal to the CE pin. If the CE pin input is "L" in the fixed time (Typ.0.5ms), the IC becomes the standby mode and turns OFF LEDs.

The current of LEDs when the CE pin is "H" input (Duty=100%) is shown by the above expression. The current of LEDs can be controlled by Duty of the PWM signal of the input CE pin. The current of LEDs when High-Duty of the CE input is Hduty reaches the value as calculatable following formula.

The frequency of the PWM signal is using the range between 200Hz to 300kHz.

When controlling the LED brightness by the PWM signal of 20kHz or less; The increasing or decreasing of the inductor current might be make a sounds in the hearable sound wave area. In that case, please use the PWM signal in the high frequency area.



Dimming control by CE pin input

## R1206N071B

NO.EA-297-170905

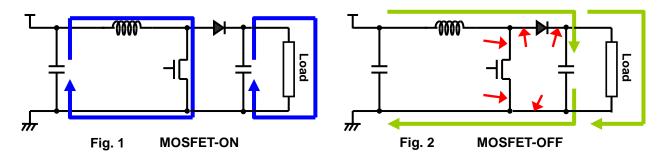
# **TECHNICAL NOTES**

#### Current Path on PCB

The current paths in an application circuit are shown in Fig. 1 and 2. A current flows through the paths shown in Fig. 1 at the time of MOSFET-ON, and shown in Fig. 2 at the time of MOSFET-OFF. In the paths pointed with red arrows in Fig. 2, current flows just in MOSFET-ON period or just in MOSFET-OFF period. Parasitic impedance/inductance and the capacitance of these paths influence stability of the system and cause noise outbreak. So please minimize this side effect. In addition, please shorten the wiring of other current paths shown in Fig. 1 and 2 except for the paths of LED load.

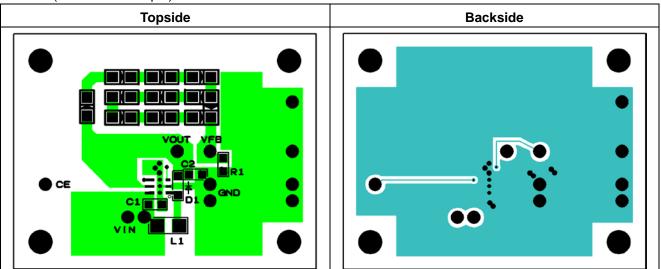
## Layout Guide for PCB

- Please shorten the wiring of the input capacitor (C1) between V<sub>IN</sub> pin and GND pin of IC. The GND pin should be connected to the strong GND plane.
- · The area of Lx land pattern should be smaller.
- The wiring between Lx pin and inductor and diode should be short and please put output capacitor (C2) close to the cathode of diode.
- · Please make the GND side of output capacitor (C2) close to the GND pin of IC.



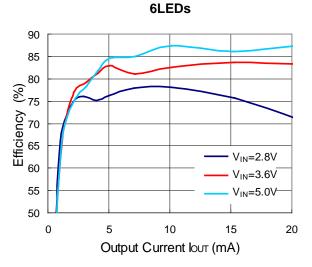
#### PCB Layout

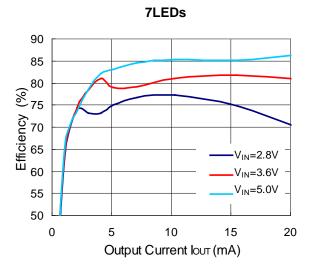
R1206N (PKG: SOT-23-6pin)



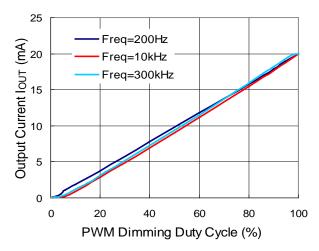
# **TYPICAL CHARACTERISTICS**

# 1) Efficiency vs. Output Current Characteristics

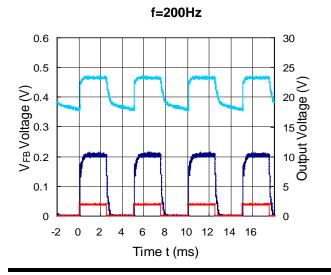


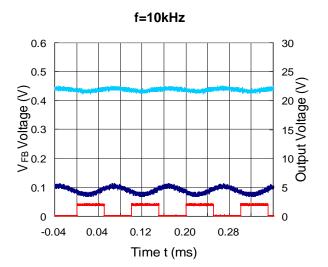


# 2) PWM Dimming Duty Cycle vs. Output Current(R1=10Ω)



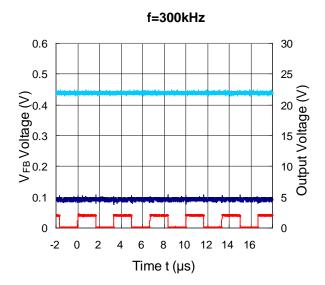
# 3) Output Current Ripple during PWM Dimming



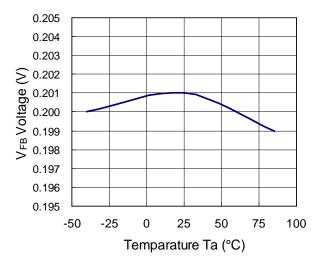


# R1206N071B

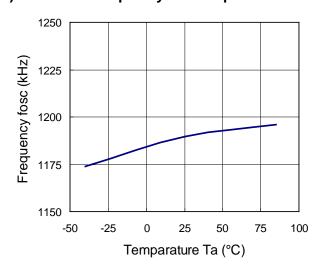
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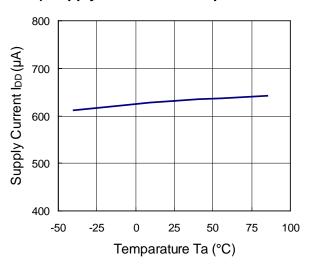
# 4) VFB Voltage vs. Temperature



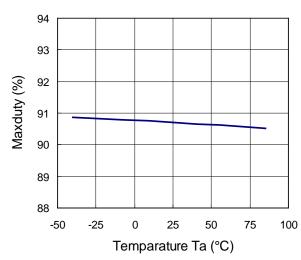
# 6) Oscillator Frequency vs. Temperature



# 5) Supply Current vs. Temperature

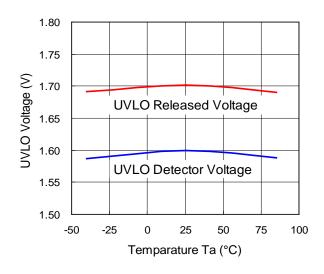


# 7) Maxduty vs. Temperature

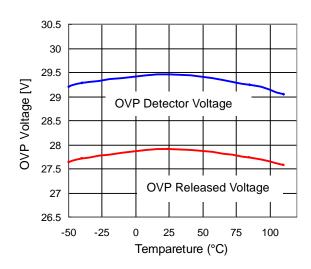


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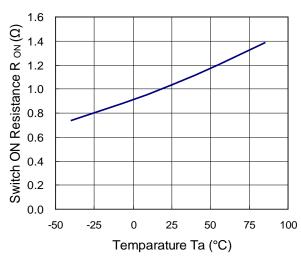
# 8) UVLO Output Voltage vs. Temperature



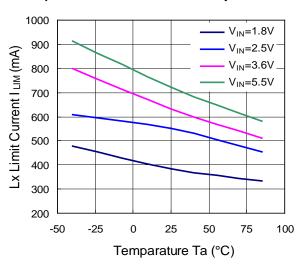
# 9) OVP Voltage vs. Temperature



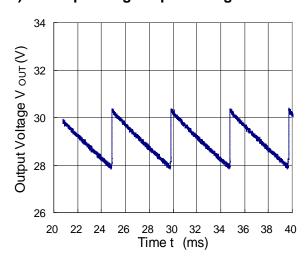
# 10) Switch ON Resistance vs. Temperature



11) Lx Current Limit vs. Temperature



# 12) OVP Operating Output Voltage Waveform



\/or /

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following conditions are used in this measurement.

#### **Measurement Conditions**

Item	Measurement Conditions (JEDEC STD. 51-7)
Environment	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)
Board Dimensions	76.2 mm × 114.3 mm × 0.8 mm
Copper Ratio	1st Layer : Less than 95% of 50 mm Square 2nd, 3rd, 4th Layers: Approx. 100% of 50 mm Square
Through-holes	φ 0.3 mm × 7 pcs

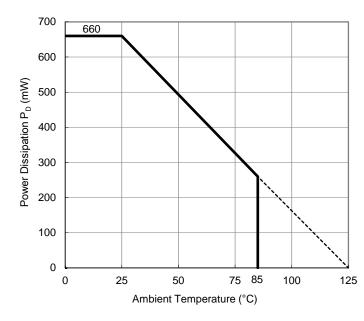
**Measurement Result** 

 $(Ta = 25^{\circ}C, Tjmax = 125^{\circ}C)$ 

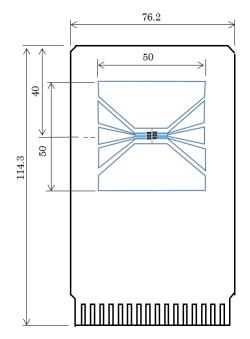
Item	Measurement Result
Power Dissipation	660 mW
Thermal Resistance (θja)	θja = 150°C/W
Thermal Characterization Parameter (ψjt)	ψjt = 51°C/W

 $\theta$ ja: Junction-to-ambient thermal resistance.

ψjt: Junction-to-top of package thermal characterization parameter

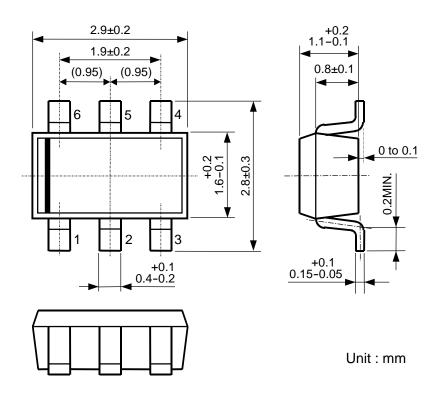


**Power Dissipation vs. Ambient Temperature** 



**Measurement Board Pattern** 

Ver. A



**SOT-23-6 Package Dimensions** 



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