

No.2087C

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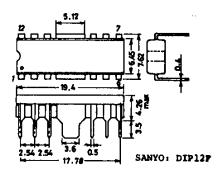
2-Channel AF Power Amp for Radio, Tape Recorder Use

Features

- . Low quiescent current
- . On-chip 2 channels permitting use in stereo and bridge amplifier applications.
- . High output
- . Minimum number of external parts required (9 pcs. minimum)
- Good ripple rejection (55dB)
- . Soft tone at the output saturation mode
- . Good channel separation
- . Easy thermal design
- . Small pop noise at the time of power supply ON/OFF
- . On-chip muting

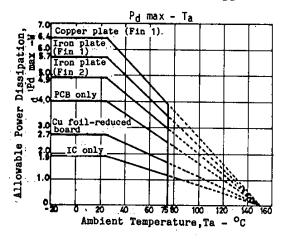
Maximum Ratings at Ta=25°C Maximum Supply Voltage Allowable Power Dissipation Operating Temperature Storage Temperature	V _{CC} max Pdmax Topr Tstg	Quiescent Operating With recommended	15 12 PCB 4 -20 to +75 -55 to +150	unit V V W oC C
Operating Conditions at Ta=25°C Recommended Supply Voltage Load Resistance Operating Voltage Range	V _{CC} R _L V _{CC} op	Stereo BTL	7.5 to 9.0 3 to 8 8 4.5 to 12	unit V ohm ohm V

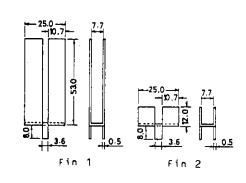
Package Dimensions 3022A-D12FIC (unit: mm)



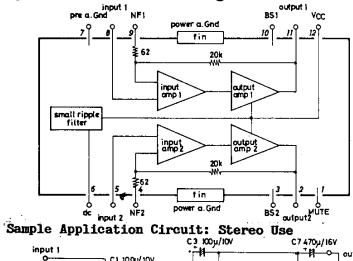
Operating Characteristics at Ta=25°C, V_{CC} =9V,f=1kHz,Rg=600ohms,R_L=4ohms,VG=50dB, See specified Test Circuit.

		min	typ	max	unit
Icco	Rg=0		20	40	mA
VG	$V_{TN} = -50 dBm$	48	50	52	dΒ
△VG	V _{TN} =-50dBm			±1	đΒ
Po	TĤD=10≸,stereo	1.7	2.4		W
	THD=10%,BTL(R_{I} =80hms)		4.5		W
THD	Po=250mW		0.3	1.0	%
ri		21	30		kohm
v _{no}	Rg=0,stereo		0.5	1.0	mV
110	Rg=10kohms,stereo		0.8	2.0	mV
\mathtt{Rr}	$Rg=0, f_{p}=100Hz, V_{CCp}=0dBm$	45	55		dΒ
CT	Rg=10kohms, Vo=0dBm	45	55		ďΒ
A_{TT}	Vo=0dBm,pin1=9V	70			dB
	VG △VG Po THD ri VNO Rr CT	VG V _{IN} =-50dBm △VG V _{IN} =-50dBm Po THD=10%, stereo THD=10%, BTL(R _L =8ohms) THD Po=250mW ri V _{NO} Rg=0, stereo Rg=10kohms, stereo Rr Rg=0, f _R =100Hz, V _{CCR} =0dBm CT Rg=10kohms, vo=0dBm	Icco Rg=0 VG V _{IN} =-50dBm 48 △VG V _{IN} =-50dBm 1.7 Po THD=10%, stereo 1.7 THD = 10%, BTL(R _L =80hms) 21 VNO Rg=0, stereo 21 Rg=10kohms, stereo Rg=10kohms, stereo 45 CT Rg=10kohms, Vo=0dBm 45	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$





Equivalent Circuit Block Diagram



Unit (resistance: Ω , capacitance: F)

input 1

C3 000µ/10V

C7 470µ/16V

output 1

C3 000µ/10V

C7 470µ/16V

output 1

C5 0.1µ

RL

33.3

RL

33.3

FRL

C4 100µ/10V

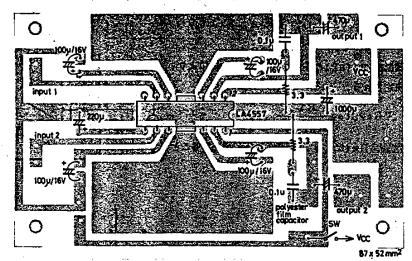
C6 C8

O.1µ

33.3

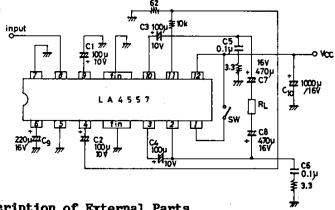
RL

Sample Printed Circuit Pattern (Cu-foiled side)



Sample Application Circuit: Bridge Amp Use

Unit (resistance: Ω, capacitance: F)



Description of External Parts

C1(C2): Feedback capacitor. The low cutoff frequency is determined by the following formula.

> f_T: Low cutoff frequency f L= 2πc1Rf Rr: Feedback resistor

Since this capacitor as well as decoupling capacitor affects the starting time, the capacitor value must be fixed with the necessary low frequency band fully considered.

C3(C4): Bootstrap capacitor. The output at low frequencies depends on this capacitor. Decreasing the capacitor value lowers the output at low frequencies. A capacitor value of 47uF or more is required.

C5(C6): Oscillation blocking capacitor. Use a polyester film capacior that is good in high frequency response and temperature characteristic. The use of an electrolytic capacitor, ceramic capacitor may cause oscillation to occur at low temperatures.

C7(C8): Output capacitor. The low cutoff frequency is determined by the following formula.

> f_L: Low cutoff frequency 2πc7RL R: Load resistance

To make the low frequency response in the bridge amplifier mode identical with that in the stereo mode, the capacitor value must be doubled.

C9:

Decoupling capacitor. Used for the ripple filter. Since the rejection effect is saturated at a certain capacitor value, it is meaningless to increase the capacitor values more than needed. This capacitor, being also used for the time constant of the muting circuit, affects the starting time.

C10:

Power source capacitor

Application Circuits

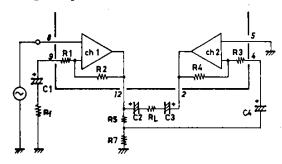
Voltage gain adjust

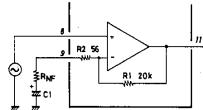
. Stereo mode

The voltage gain is determined by on-chip resistor R1(R2) and external feedback resistor $R_{\rm Nf}$ as follows:

Any voltage gain can be obtained by external resistor R_{Nf} .

. Bridge amplifier mode





Unit (resistance: Ω)

The CH1 is a noninverting amplifier and the CH2 is an inverting amplifier. The total voltage gain, being apparently higher than that of the CH1 by 6dB, is approximately calculated by the following formula.

VG=20log R2/R1+6(dB)

To reduce the voltage gain, Rf is connected and the following formula is used. VG=20log R2/Rf+R1+6(dB)

Proper cares in using LA4557-applied set

- Slider contact noise of variable resistor
 Since the input circuit uses PNP transistors, no input coupling capacitor
 is required. However, if slider contact noise of the variable resistor
 presents any problem, connect a capacitor in series with input.
- 2. Pop noise

If pop noise generated at the time of power ON/OFF disturbs you, connect a resistor of 500ohms to 1kohm across the middle point and GND.

Thermal design

Since the DIP12F package is such that the Cu-foiled area of the printed circuit board is used to dissipate heat, make the Cu-foiled area in the vicinity of the heat sink of the IC as large as possible when designing the printed circuit board. Power dissipation Pd is increased depending on the supply voltage and load. So, it is recommended to use the printed circuit board together with the heat sink. The following is a formula to be used to calculate Pd (for stereo use). For AC power supply, however, it is recommended to actually measure Pd on the tramsformer of each set. For bridge amplifier use, Pd is calculated at 1/2 of the load.

(1) DC power supply

Pd max =
$$\frac{VCC^2}{\pi^2 RL}$$
 + I cco · VCC (for stereo use)----(1)

(2) AC power supply

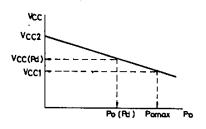
 V_{CC2} : Supply voltage at quiescent mode

V_{CC(Pd)}: Supply voltage at Pd max

V_{CC1}: Supply voltage at maximum output

r : Voltage regulation $\frac{Vcc2-Vcc1}{Vcc1}$

Icco : Quiescent current



Line regulation $V_{CC(Pd)^2}$

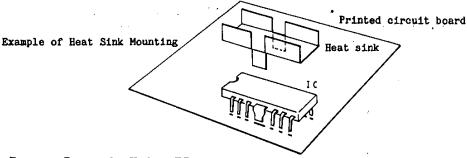
Pd max=
$$\frac{\sqrt{CC(1d)}}{\pi^2 RL}$$
 + I cco·VCC(Pd)(for stereo use)····(2)

where

$$Vcc(Pd) = \frac{(1+r)Vcc1}{r \cdot Vcc1} \times \sqrt{\frac{RL}{Po \text{ max}}}$$

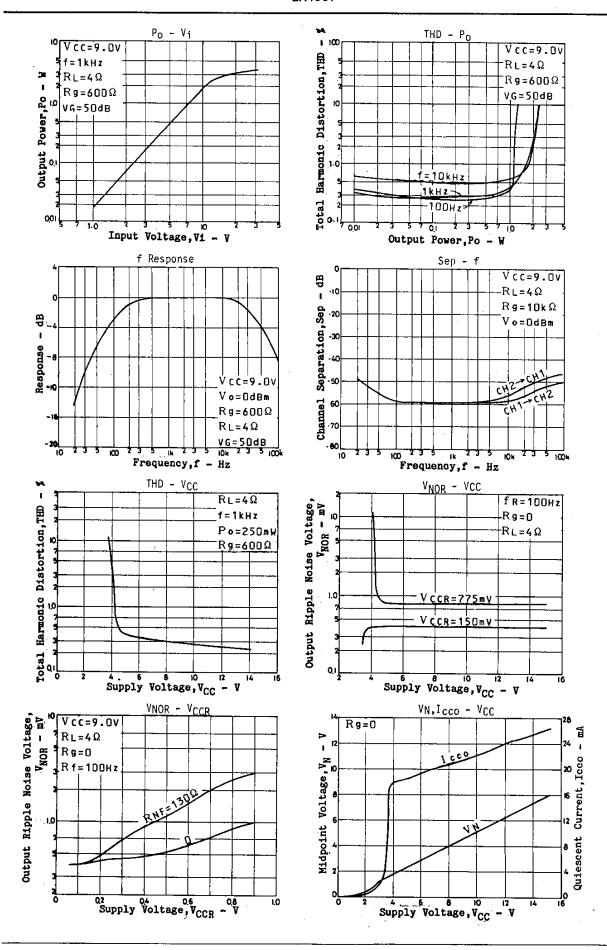
Example of Heat Sink Mounting Method

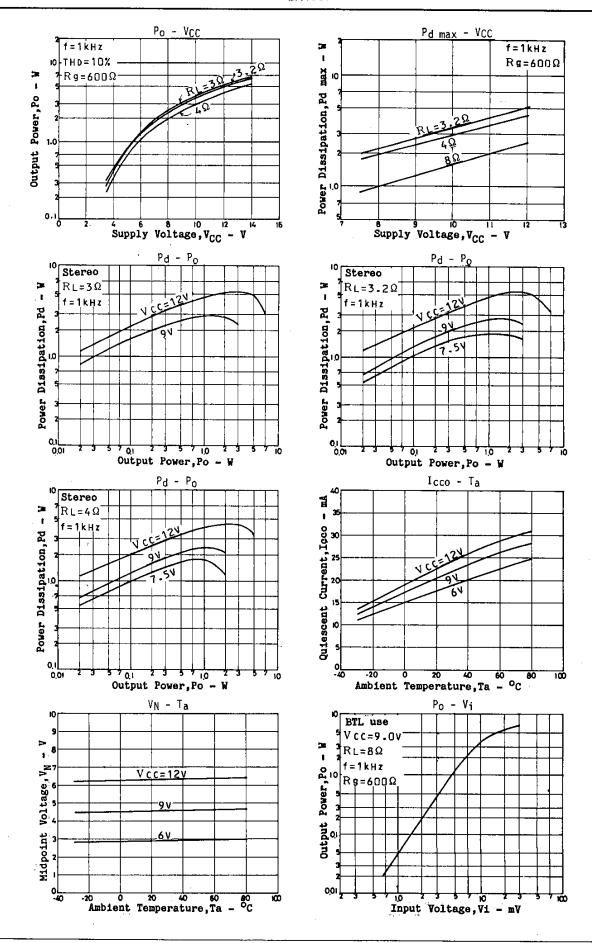
The heat sink must be of such a shape as to be able to dissipate heat from the IC plastic area and fin area and is soldered to the printed circuit board as shown below. For the size of the heat sink, refer to the Pd-Ta characteristic. The material of the heat sink is recommended to be copper or iron which is solderable. It is recommended to apply silicone grease to the IC plastic area to reduce thermal resistance between the heat sink and the IC plastic area.

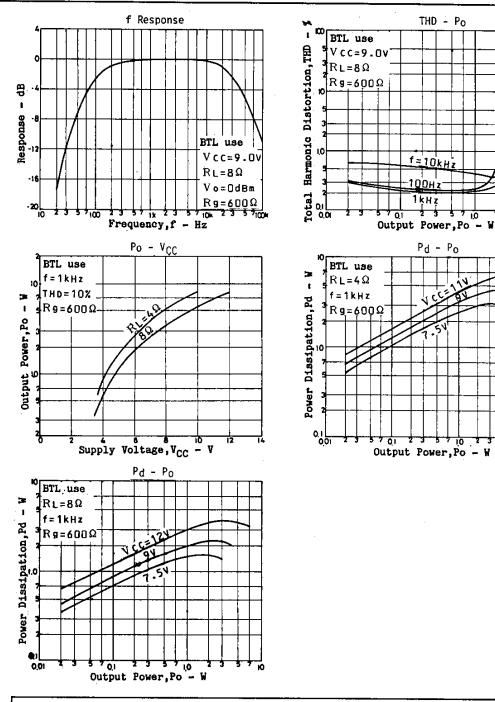


Proper Cares in Using IC

- 1. If the IC is used in the vicinity of the maximum ratings, even a slight variation in conditions may cause the maximum ratings to be exceeded, thereby leading to breakdown. Allow an ample margin of variation for supply voltage, etc. and use the IC in the range where the maximum ratings are not exceeded.
- 2. Pin-to-pin short: If power is applied when the space between pins is shorted, breakdown or deterioration may occur. When mounting the IC on the board or applying power, make sure that the space between pins is not shorted with solder, etc.
- 3. Load short: If the IC is used with the load shorted for a long time, breakdown or deterioration may occur. Be sure not to short the load.
- 4. When the IC is used in radios or radio cassette tape recorders, keep a good distance between IC and bar antenna.
- 5. When making the board, refer to the sample printed circuit pattern.
- 6. It should be noted that some plug jacks to be used for connecting to the external speaker are such that both poles are shorted once when connecting.







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