

**Single/Dual/Quad 100MHz,
Rail-to-Rail Input and Output,
Ultralow $1.9\text{nV}/\sqrt{\text{Hz}}$ Noise, Low Power Op Amps**

FEATURES

- **Low Noise Voltage: $1.9\text{nV}/\sqrt{\text{Hz}}$ (100kHz)**
- **Low Supply Current: 3mA/Amp Max**
- **Gain Bandwidth Product: 100MHz**
- **Dual LT6203 in Tiny DFN Package**
- Low Distortion: -80dB at 1MHz
- Low Offset Voltage: $500\mu\text{V}$ Max
- Wide Supply Range: 2.5V to 12.6V
- Input Common Mode Range Includes Both Rails
- Output Swings Rail-to-Rail
- Common Mode Rejection Ratio 90dB Typ
- Unity Gain Stable
- Low Noise Current: $1.1\text{pA}/\sqrt{\text{Hz}}$
- Output Current: 30mA Min
- Operating Temperature Range -40°C to 85°C
- Low Profile (1mm) SOT-23 (ThinSOT™) Package

APPLICATIONS

- Low Noise, Low Power Signal Processing
- Active Filters
- Rail-to-Rail Buffer Amplifiers
- Driving A/D Converters
- DSL Receivers
- Battery Powered/Battery Backed Equipment

DESCRIPTION

The LTC®6202/LT6203/LT6204 are single/dual/quad low noise, rail-to-rail input and output unity gain stable op amps that feature $1.9\text{nV}/\sqrt{\text{Hz}}$ noise voltage and draw only 2.5mA of supply current per amplifier. These amplifiers combine very low noise and supply current with a 100MHz gain bandwidth product, a $25\text{V}/\mu\text{s}$ slew rate, and are optimized for low supply signal conditioning systems.

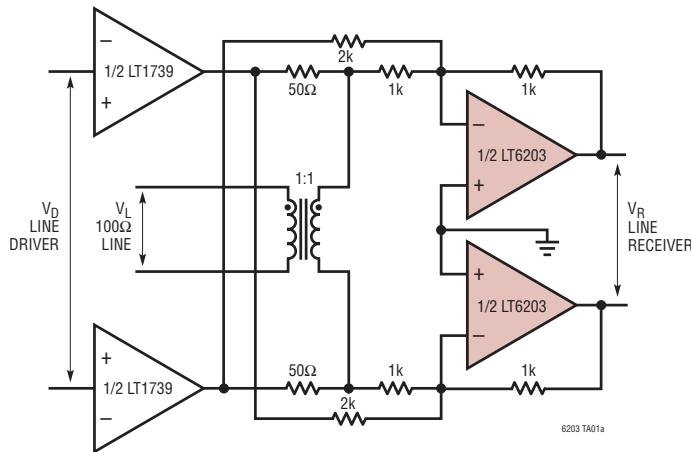
These amplifiers maintain their performance for supplies from 2.5V to 12.6V and are specified at 3V, 5V and $\pm 5\text{V}$ supplies. Harmonic distortion is less than -80dBc at 1MHz making these amplifiers suitable in low power data acquisition systems.

The LT6202 is available in the 5-pin TSOT-23 and the 8-pin SO, while the LT6203 comes in 8-pin SO and MSOP packages with standard op amp pinouts. For compact layouts the LT6203 is also available in a tiny fine line leadless package (DFN), while the quad LT6204 is available in the 16-pin SSOP and 14-pin SO packages. These devices can be used as plug-in replacements for many op amps to improve input/output range and noise performance.

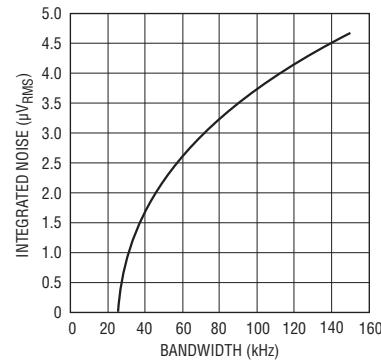
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TYPICAL APPLICATION

Low Noise 4- to 2-Wire Local Echo Cancellation Differential Receiver



Line Receiver Integrated Noise 25kHz to 150kHz



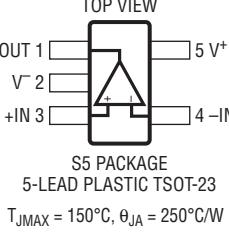
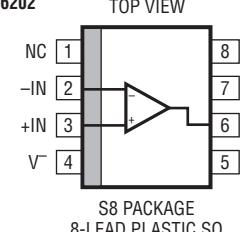
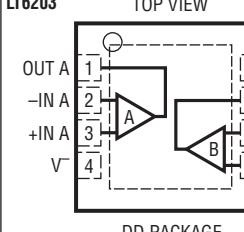
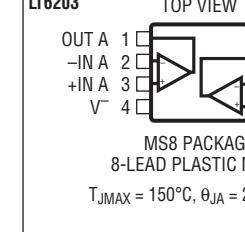
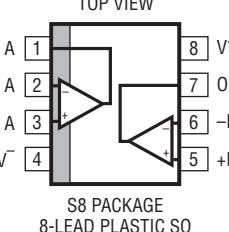
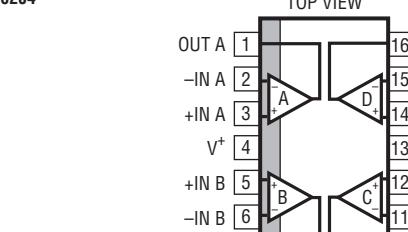
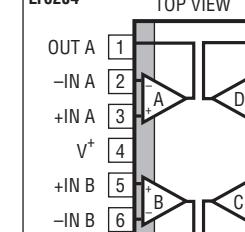
LT6202/LT6203/LT6204

ABSOLUTE MAXIMUM RATINGS

(Note 1)

Total Supply Voltage (V ⁺ to V ⁻)	12.6V	Junction Temperature (DD Package)	125°C
Input Current (Note 2)	±40mA	Storage Temperature Range	−65°C to 150°C
Output Short-Circuit Duration (Note 3)	Indefinite	Storage Temperature Range	
Operating Temperature Range (Note 4)	−40°C to 85°C	(DD Package)	−65°C to 125°C
Specified Temperature Range (Note 5)	−40°C to 85°C	Lead Temperature (Soldering, 10 sec)	300°C
Junction Temperature	150°C		

PIN CONFIGURATION

LT6202  <p>TOP VIEW S5 PACKAGE 5-LEAD PLASTIC TSOT-23 T_{JMAX} = 150°C, θ_{JA} = 250°C/W</p>	LT6202  <p>TOP VIEW S8 PACKAGE 8-LEAD PLASTIC SO T_{JMAX} = 150°C, θ_{JA} = 190°C/W</p>	LT6203  <p>TOP VIEW DD PACKAGE 8-LEAD (3mm × 3mm) PLASTIC DFN T_{JMAX} = 125°C, θ_{JA} = 160°C/W UNDERSIDE METAL CONNECTED TO V⁻</p>	LT6203  <p>TOP VIEW MS8 PACKAGE 8-LEAD PLASTIC MSOP T_{JMAX} = 150°C, θ_{JA} = 250°C/W</p>
LT6203  <p>TOP VIEW S8 PACKAGE 8-LEAD PLASTIC SO T_{JMAX} = 150°C, θ_{JA} = 190°C/W</p>	LT6204  <p>TOP VIEW GN PACKAGE 16-LEAD NARROW PLASTIC SSOP T_{JMAX} = 150°C, θ_{JA} = 135°C/W</p>		LT6204  <p>TOP VIEW S PACKAGE 14-LEAD PLASTIC SO T_{JMAX} = 150°C, θ_{JA} = 150°C/W</p>

ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE
LT6202CS5#PBF	LT6202CS5#TRPBF	LTG6	5-Lead Plastic TSOT-23	0°C to 70°C
LT6202IS5#PBF	LT6202IS5#TRPBF	LTG6	5-Lead Plastic TSOT-23	−40°C to 85°C
LT6202CS8#PBF	LT6202CS8#TRPBF	6202	8-Lead Plastic SO	0°C to 70°C
LT6202IS8#PBF	LT6202IS8#TRPBF	6202I	8-Lead Plastic SO	−40°C to 85°C
LT6203CDD#PBF	LT6203CDD#TRPBF	LAAP	8-Lead (3mm × 3mm) Plastic DFN	0°C to 70°C
LT6203IDD#PBF	LT6203IDD#TRPBF	LAAP	8-Lead (3mm × 3mm) Plastic DFN	−40°C to 85°C

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ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE
LT6203CMS8#PBF	LT6203CMS8#TRPBF	LTB2	8-Lead Plastic MSOP	0°C to 70°C
LT6203IMS8#PBF	LT6203IMS8#TRPBF	LTB3	8-Lead Plastic MSOP	-40°C to 85°C
LT6203CS8#PBF	LT6203CS8#TRPBF	6203	8-Lead Plastic SO	0°C to 70°C
LT6203IS8#PBF	LT6203IS8#TRPBF	6203I	8-Lead Plastic SO	-40°C to 85°C
LT6204CGN#PBF	LT6204CGN#TRPBF	6204	16-Lead Narrow Plastic SSOP	0°C to 70°C
LT6204IGN#PBF	LT6204IGN#TRPBF	6204I	16-Lead Narrow Plastic SSOP	-40°C to 85°C
LT6204CS#PBF	LT6204CS#TRPBF	LT6204CS	14-Lead Plastic SO	0°C to 70°C
LT6204IS#PBF	LT6204IS#TRPBF	LT6204IS	14-Lead Plastic SO	-40°C to 85°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container. Consult LTC Marketing for information on non-standard lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>

ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_S = 5\text{V, } 0\text{V}$; $V_S = 3\text{V, } 0\text{V}$; $V_{CM} = V_{OUT} = \text{half supply}$,
unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_S = 5\text{V, } 0\text{V}$, $V_{CM} = \text{Half Supply}$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	0.1	0.5		mV
		$V_S = 3\text{V, } 0\text{V}$, $V_{CM} = \text{Half Supply}$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	0.6	1.5		mV
		$V_S = 5\text{V, } 0\text{V}$, $V_{CM} = V^+ \text{ to } V^-$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	0.25	2.0		mV
		$V_S = 3\text{V, } 0\text{V}$, $V_{CM} = V^+ \text{ to } V^-$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	1.0	3.5		mV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^- \text{ to } V^+$	0.15	0.8		mV
			0.3	1.8		mV
I_B	Input Bias Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	-7.0 -8.8	-1.3 -3.3	2.5	μA
ΔI_B	I_B Shift	$V_{CM} = V^- \text{ to } V^+$		4.7	11.3	μA
	I_B Match (Channel-to-Channel) (Note 6)			0.1	0.6	μA
I_{OS}	Input Offset Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	0.12 0.07 0.12	1 1 1.1		μA
	Input Noise Voltage	0.1Hz to 10Hz		800		$\text{nV}_{\text{P-P}}$
e_n	Input Noise Voltage Density	$f = 100\text{kHz}$, $V_S = 5\text{V}$ $f = 10\text{kHz}$, $V_S = 5\text{V}$		2 2.9	4.5	$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Noise Current Density, Balanced Input Noise Current Density, Unbalanced	$f = 10\text{kHz}$, $V_S = 5\text{V}$		0.75 1.1		$\text{pA}/\sqrt{\text{Hz}}$
	Input Resistance	Common Mode Differential Mode		4 12		$\text{m}\Omega$ $\text{k}\Omega$

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LT6202/LT6203/LT6204

ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_S = 5\text{V, } 0\text{V}$; $V_S = 3\text{V, } 0\text{V}$; $V_{CM} = V_{OUT} = \text{half supply}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
C_{IN}	Input Capacitance	Common Mode Differential Mode		1.8 1.5		pF pF
A_{VOL}	Large Signal Gain	$V_S = 5\text{V, } V_0 = 0.5\text{V to } 4.5\text{V, } R_L = 1\text{k to } V_S/2$ $V_S = 5\text{V, } V_0 = 1\text{V to } 4\text{V, } R_L = 100 \text{ to } V_S/2$ $V_S = 3\text{V, } V_0 = 0.5\text{V to } 2.5\text{V, } R_L = 1\text{k to } V_S/2$	40 8.0 17	70 14 40		V/mV V/mV V/mV
CMRR	Common Mode Rejection Ratio	$V_S = 5\text{V, } V_{CM} = V^- \text{ to } V^+$ $V_S = 5\text{V, } V_{CM} = 1.5\text{V to } 3.5\text{V}$ $V_S = 3\text{V, } V_{CM} = V^- \text{ to } V^+$	60 80 56	83 100 80		dB dB dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_S = 5\text{V, } V_{CM} = 1.5\text{V to } 3.5\text{V}$	85	120		dB
PSRR	Power Supply Rejection Ratio	$V_S = 2.5\text{V to } 10\text{V, } V_{CM} = 0\text{V}$	60	74		dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = 2.5\text{V to } 10\text{V, } V_{CM} = 0\text{V}$	70	100		dB
	Minimum Supply Voltage (Note 7)		2.5			V
V_{OL}	Output Voltage Swing LOW Saturation (Note 8)	No Load $I_{SINK} = 5\text{mA}$ $V_S = 5\text{V, } I_{SINK} = 20\text{mA}$ $V_S = 3\text{V, } I_{SINK} = 15\text{mA}$		5 85 240 185	50 190 460 350	mV mV mV mV
V_{OH}	Output Voltage Swing HIGH Saturation (Note 8)	No Load $I_{SOURCE} = 5\text{mA}$ $V_S = 5\text{V, } I_{SOURCE} = 20\text{mA}$ $V_S = 3\text{V, } I_{SOURCE} = 15\text{mA}$		25 90 325 225	75 210 600 410	mV mV mV mV
I_{SC}	Short-Circuit Current	$V_S = 5\text{V}$ $V_S = 3\text{V}$	± 30 ± 25	± 45 ± 40		mA mA
I_S	Supply Current per Amp	$V_S = 5\text{V}$ $V_S = 3\text{V}$		2.5 2.3	3.0 2.85	mA mA
GBW	Gain Bandwidth Product	Frequency = 1MHz, $V_S = 5\text{V}$		90		MHz
SR	Slew Rate	$V_S = 5\text{V, } A_V = -1, R_L = 1\text{k, } V_0 = 4\text{V}$	17	24		V/μs
FPBW	Full Power Bandwidth (Note 10)	$V_S = 5\text{V, } V_{OUT} = 3\text{V}_{P-P}$	1.8	2.5		MHz
t_S	Settling Time	0.1%, $V_S = 5\text{V, } V_{STEP} = 2\text{V, } A_V = -1, R_L = 1\text{k}$	85			ns

The ● denotes the specifications which apply over $0^\circ\text{C} < T_A < 70^\circ\text{C}$ temperature range. $V_S = 5\text{V, } 0\text{V}$; $V_S = 3\text{V, } 0\text{V}$; $V_{CM} = V_{OUT} = \text{half supply}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_S = 5\text{V, } 0\text{V, } V_{CM} = \text{Half Supply}$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	0.2 0.2	0.7 0.9	mV mV
		$V_S = 3\text{V, } 0\text{V, } V_{CM} = \text{Half Supply}$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	0.6 0.6	1.7 1.9	mV mV
		$V_S = 5\text{V, } 0\text{V, } V_{CM} = V^+ \text{ to } V^-$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	0.7 0.7	2.5 2.7	mV mV
		$V_S = 3\text{V, } 0\text{V, } V_{CM} = V^+ \text{ to } V^-$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	1.2 1.2	4.0 4.2	mV mV
$V_{OS\ TC}$	Input Offset Voltage Drift (Note 9)	$V_{CM} = \text{Half Supply}$	●	3.0	9.0	μV/°C
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^- \text{ to } V^+$	● ●	0.15 0.5	0.9 2.3	mV mV

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ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over $0^{\circ}\text{C} < T_A < 70^{\circ}\text{C}$
 temperature range. $V_S = 5\text{V, } 0\text{V}$; $V_S = 3\text{V, } 0\text{V}$; $V_{CM} = V_{OUT}$ = half supply, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
I_B	Input Bias Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	● ● ●	-7.0 1.3 -8.8	-1.3 2.5 -3.3	μA μA μA
ΔI_B	I_B Shift	$V_{CM} = V^-$ to V^+	●	4.7	11.3	μA
	I_B Match (Channel-to-Channel) (Note 6)		●	0.1	0.6	μA
I_{OS}	Input Offset Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	● ● ●	0.15 0.10 0.15	1 1 1.1	μA μA μA
A_{VOL}	Large Signal Gain	$V_S = 5\text{V, } V_0 = 0.5\text{V to } 4.5\text{V, } R_L = 1\text{k to } V_S/2$ $V_S = 5\text{V, } V_0 = 1.5\text{V to } 3.5\text{V, } R_L = 100$ to $V_S/2$ $V_S = 3\text{V, } V_0 = 0.5\text{V to } 2.5\text{V, } R_L = 1\text{k to } V_S/2$	● ● ●	35 6.0 15	60 12 36	V/mV V/mV V/mV
CMRR	Common Mode Rejection Ratio	$V_S = 5\text{V, } V_{CM} = V^-$ to V^+ $V_S = 5\text{V, } V_{CM} = 1.5\text{V to } 3.5\text{V}$ $V_S = 3\text{V, } V_{CM} = V^-$ to V^+	● ● ●	60 78 56	83 97 75	dB dB dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_S = 5\text{V, } V_{CM} = 1.5\text{V to } 3.5\text{V}$	●	83	100	dB
PSRR	Power Supply Rejection Ratio	$V_S = 3\text{V to } 10\text{V, } V_{CM} = 0\text{V}$	●	60	70	dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = 3\text{V to } 10\text{V, } V_{CM} = 0\text{V}$	●	70	100	dB
	Minimum Supply Voltage (Note 7)		●	3.0		V
V_{OL}	Output Voltage Swing LOW Saturation (Note 8)	No Load $I_{SINK} = 5\text{mA}$ $I_{SINK} = 15\text{mA}$	● ● ●	5.0 95 260	60 200 365	mV mV mV
V_{OH}	Output Voltage Swing HIGH Saturation (Note 8)	No Load $I_{SOURCE} = 5\text{mA}$ $V_S = 5\text{V, } I_{SOURCE} = 20\text{mA}$ $V_S = 3\text{V, } I_{SOURCE} = 15\text{mA}$	● ● ● ●	50 115 360 260	100 230 635 430	mV mV mV mV
I_{SC}	Short-Circuit Current	$V_S = 5\text{V}$ $V_S = 3\text{V}$	● ●	± 20 ± 20	± 33 ± 30	mA mA
I_S	Supply Current per Amp	$V_S = 5\text{V}$ $V_S = 3\text{V}$	● ●	3.1 2.75	3.85 3.50	mA mA
GBW	Gain Bandwidth Product	Frequency = 1MHz	●	87		MHz
SR	Slew Rate	$V_S = 5\text{V, } A_V = -1, R_L = 1\text{k, } V_0 = 4\text{V}$	●	15	21	$\text{V}/\mu\text{s}$
FPBW	Full Power Bandwidth (Note 10)	$V_S = 5\text{V, } V_{OUT} = 3\text{V}_{\text{P-P}}$	●	1.6	2.2	MHz

The ● denotes the specifications which apply over $-40^{\circ}\text{C} < T_A < 85^{\circ}\text{C}$ temperature range. $V_S = 5\text{V, } 0\text{V}$; $V_S = 3\text{V, } 0\text{V}$; $V_{CM} = V_{OUT}$ = half supply, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_S = 5\text{V, } 0\text{V, } V_{CM} = \text{Half Supply}$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	0.2 0.2	0.8 1.0	mV mV
		$V_S = 3\text{V, } 0\text{V, } V_{CM} = \text{Half Supply}$ LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	0.6 0.6	2.0 2.2	mV mV
		$V_S = 5\text{V, } 0\text{V, } V_{CM} = V^+$ to V^- LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	1.0 1.0	3.0 3.5	mV mV
		$V_S = 3\text{V, } 0\text{V, } V_{CM} = V^+$ to V^- LT6203, LT6204, LT6202S8 LT6202 TSOT-23	● ●	1.4 1.4	4.5 4.7	mV mV

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over $-40^{\circ}\text{C} < T_A < 85^{\circ}\text{C}$
 temperature range. $V_S = 5\text{V, } 0\text{V}$; $V_S = 3\text{V, } 0\text{V}$; $V_{CM} = V_{OUT} = \text{half supply}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$V_{OS\ TC}$	Input Offset Voltage Drift (Note 9)	$V_{CM} = \text{Half Supply}$	●	3.0	9.0	$\mu\text{V}/^{\circ}\text{C}$
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^- \text{ to } V^+$	● ●	0.3 0.7	1.0 2.5	mV mV
I_B	Input Bias Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	● ● ●	-7.0 1.3 -8.8	-1.3 2.5 -3.3	μA μA μA
ΔI_B	I_B Shift	$V_{CM} = V^- \text{ to } V^+$	●	4.7	11.3	μA
	I_B Match (Channel-to-Channel) (Note 6)		●	0.1	0.6	μA
I_{OS}	Input Offset Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	● ● ●	0.2 0.2 0.2	1 1.1 1.2	μA μA μA
A_{VOL}	Large Signal Gain	$V_S = 5\text{V, } V_0 = 0.5\text{V to } 4.5\text{V, } R_L = 1\text{k to } V_S/2$ $V_S = 5\text{V, } V_0 = 1.5\text{V to } 3.5\text{V, } R_L = 100 \text{ to } V_S/2$ $V_S = 3\text{V, } V_0 = 0.5\text{V to } 2.5\text{V, } R_L = 1\text{k to } V_S/2$	● ● ●	32 4.0 13	60 10 32	V/mV V/mV V/mV
CMRR	Common Mode Rejection Ratio	$V_S = 5\text{V, } V_{CM} = V^- \text{ to } V^+$ $V_S = 5\text{V, } V_{CM} = 1.5\text{V to } 3.5\text{V}$ $V_S = 3\text{V, } V_{CM} = V^- \text{ to } V^+$	● ● ●	60 75 56	80 95 75	dB dB dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_S = 5\text{V, } V_{CM} = 1.5\text{V to } 3.5\text{V}$	●	80	100	dB
PSRR	Power Supply Rejection Ratio	$V_S = 3\text{V to } 10\text{V, } V_{CM} = 0\text{V}$	●	60	70	dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = 3\text{V to } 10\text{V, } V_{CM} = 0\text{V}$	●	70	100	dB
	Minimum Supply Voltage (Note 7)		●	3.0		V
V_{OL}	Output Voltage Swing LOW Saturation (Note 8)	No Load $I_{SINK} = 5\text{mA}$ $I_{SINK} = 15\text{mA}$	● ● ●	6 95 210	70 210 400	mV mV mV
V_{OH}	Output Voltage Swing HIGH Saturation (Note 8)	No Load $I_{SOURCE} = 5\text{mA}$ $V_S = 5\text{V, } I_{SOURCE} = 15\text{mA}$ $V_S = 3\text{V, } I_{SOURCE} = 15\text{mA}$	● ● ● ●	55 125 370 270	110 240 650 650	mV mV mV mV
I_{SC}	Short-Circuit Current	$V_S = 5\text{V}$ $V_S = 3\text{V}$	● ●	± 15 ± 15	± 25 ± 23	mA mA
I_S	Supply Current per Amp	$V_S = 5\text{V}$ $V_S = 3\text{V}$	● ●	3.3 3.0	4.1 3.65	mA mA
GBW	Gain Bandwidth Product	Frequency = 1MHz	●	83		MHz
SR	Slew Rate	$V_S = 5\text{V, } A_V = -1, R_L = 1\text{k, } V_0 = 4\text{V}$	●	12	17	$\text{V}/\mu\text{s}$
FPBW	Full Power Bandwidth (Note 10)	$V_S = 5\text{V, } V_{OUT} = 3\text{V}_{\text{P-P}}$	●	1.3	1.8	MHz

ELECTRICAL CHARACTERISTICS

 $T_A = 25^\circ\text{C}$, $V_S = \pm 5\text{V}$; $V_{CM} = V_{OUT} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6203, LT6204, LT6202S8 $V_{CM} = 0\text{V}$ $V_{CM} = V^+$ $V_{CM} = V^-$		1.0	2.5	mV
		LT6202 SOT-23 $V_{CM} = 0\text{V}$ $V_{CM} = V^+$ $V_{CM} = V^-$		2.6	5.5	mV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)			2.3	5.0	mV
		$V_{CM} = 0\text{V}$ $V_{CM} = V^- \text{ to } V^+$		1.0	2.7	mV
I_B	Input Bias Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$	-7.0	-1.3	3.0	μA
				1.3	-3.8	μA
ΔI_B	I_B Shift	$V_{CM} = V^- \text{ to } V^+$		5.3	12.5	μA
	I_B Match (Channel-to-Channel) (Note 6)			0.1	0.6	μA
I_{OS}	Input Offset Current	$V_{CM} = \text{Half Supply}$ $V_{CM} = V^+$ $V_{CM} = V^-$		0.15	1	μA
				0.2	1.2	μA
	Input Noise Voltage	0.1Hz to 10Hz		0.35	1.3	μA
e_n	Input Noise Voltage Density	$f = 100\text{kHz}$		800		$\text{nV}_{\text{P-P}}$
		$f = 10\text{kHz}$		1.9		$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Noise Current Density, Balanced Input Noise Current Density, Unbalanced	$f = 10\text{kHz}$		2.8	4.5	$\text{nV}/\sqrt{\text{Hz}}$
				0.75		$\text{pA}/\sqrt{\text{Hz}}$
	Input Resistance	Common Mode Differential Mode		1.1		$\text{pA}/\sqrt{\text{Hz}}$
C_{IN}	Input Capacitance	Common Mode Differential Mode		4		$\text{M}\Omega$
				12		$\text{k}\Omega$
A_{VOL}	Large Signal Gain	$V_0 = \pm 4.5\text{V}$, $R_L = 1\text{k}$ $V_0 = \pm 2.5\text{V}$, $R_L = 100$	75	130		V/mV
			11	19		V/mV
CMRR	Common Mode Rejection Ratio	$V_{CM} = V^- \text{ to } V^+$ $V_{CM} = -2\text{V} \text{ to } 2\text{V}$	65	85		dB
			85	98		dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_{CM} = -2\text{V} \text{ to } 2\text{V}$		85	120	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.25\text{V} \text{ to } \pm 5\text{V}$	60	74		dB
		$V_S = \pm 1.25\text{V} \text{ to } \pm 5\text{V}$		70	100	dB
V_{OL}	Output Voltage Swing LOW Saturation (Note 8)	No Load $I_{SINK} = 5\text{mA}$ $I_{SINK} = 20\text{mA}$		5	50	mV
				87	190	mV
V_{OH}	Output Voltage Swing HIGH Saturation (Note 8)	No Load $I_{SOURCE} = 5\text{mA}$ $I_{SOURCE} = 20\text{mA}$		245	460	mV
				40	95	mV
I_{SC}	Short-Circuit Current			95	210	mV
				320	600	mV
I_S	Supply Current per Amp			40	95	mA
				95	210	mA
GBW	Gain Bandwidth Product	Frequency = 1MHz	70	100		MHz
				18	25	$\text{V}/\mu\text{s}$
SR	Slew Rate	$A_V = -1$, $R_L = 1\text{k}$, $V_0 = 4\text{V}$	1.9	2.6		MHz
				78		ns
FPBW	Full Power Bandwidth (Note 10)	$V_{OUT} = 3\text{V}_{\text{P-P}}$		0.05		$\%$
				0.03		DEG
t_S	Settling Time	0.1% , $V_{STEP} = 2\text{V}$, $A_V = -1$, $R_L = 1\text{k}$		0.05		$\%$
				0.03		DEG
dG	Differential Gain (Note 11)	$A_V = 2$, $R_F = R_G = 499\Omega$, $R_L = 2\text{k}$		0.05		$\%$
				0.03		DEG

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LT6202/LT6203/LT6204

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over $0^{\circ}\text{C} < T_A < 70^{\circ}\text{C}$ temperature range. $V_S = \pm 5\text{V}$; $V_{CM} = V_{OUT} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6203, LT6204, LT6202S8				
		$V_{CM} = 0\text{V}$	●	1.6	2.8	mV
		$V_{CM} = V^+$	●	3.2	6.8	mV
		$V_{CM} = V^-$	●	2.8	5.8	mV
		LT6202 SOT-23				
		$V_{CM} = 0\text{V}$	●	1.6	3.0	mV
		$V_{CM} = V^+$	●	3.2	7.3	mV
		$V_{CM} = V^-$	●	2.8	6.3	mV
$V_{OS\,TC}$	Input Offset Voltage Drift (Note 9)	$V_{CM} = \text{Half Supply}$	●	7.5	24	$\mu\text{V}/^{\circ}\text{C}$
I_B	Input Offset Voltage Match (Channel-to-Channel) (Note 6)	$V_{CM} = 0\text{V}$	●	0.2	1.0	mV
I_B	Input Bias Current	$V_{CM} = \text{Half Supply}$	●	-7.0	-1.4	μA
ΔI_B	I_B Shift	$V_{CM} = V^-$ to V^+	●	1.8	3.6	μA
ΔI_B	I_B Match (Channel-to-Channel) (Note 6)		●	-10	-4.3	μA
I_{OS}	Input Offset Current	$V_{CM} = \text{Half Supply}$	●	5.4	13	μA
V_{OS}	Input Offset Current	$V_{CM} = V^+$	●	0.1	1	μA
V_{OS}	Input Offset Current	$V_{CM} = V^-$	●	0.2	1.2	μA
A_{VOL}	Large Signal Gain	$V_0 = \pm 4.5\text{V}$, $R_L = 1\text{k}$	●	70	120	V/mV
V_{CMRR}	Common Mode Rejection Ratio	$V_{CM} = V^-$ to V^+	●	65	84	dB
V_{CMRR}	CMRR Match (Channel-to-Channel) (Note 6)	$V_{CM} = -2\text{V}$ to 2V	●	83	95	dB
$PSRR$	Power Supply Rejection Ratio	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	●	60	70	dB
$PSRR$	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	●	70	100	dB
V_{OL}	Output Voltage Swing LOW Saturation (Note 8)	No Load	●	6	70	mV
V_{OH}	Output Voltage Swing HIGH Saturation (Note 8)	No Load	●	95	200	mV
I_{SC}	Short-Circuit Current		●	210	400	mV
I_S	Supply Current per Amp		●	65	120	mV
GBW	Gain Bandwidth Product	Frequency = 1MHz	●	125	240	mV
SR	Slew Rate	$A_V = -1$, $R_L = 1\text{k}$, $V_0 = 4\text{V}$	●	350	625	mV
$FPBW$	Full Power Bandwidth (Note 10)	$V_{OUT} = 3\text{V}_{\text{P-P}}$	●	1.7	2.3	MHz

The ● denotes the specifications which apply over $-40^{\circ}\text{C} < T_A < 85^{\circ}\text{C}$ temperature range. $V_S = \pm 5\text{V}$; $V_{CM} = V_{OUT} = 0\text{V}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6203, LT6204, LT6202S8				
		$V_{CM} = 0\text{V}$	●	1.7	3.0	mV
		$V_{CM} = V^+$	●	3.8	7.5	mV
		$V_{CM} = V^-$	●	3.5	6.6	mV
		LT6202 SOT-23				
		$V_{CM} = 0\text{V}$	●	1.7	3.2	mV
		$V_{CM} = V^+$	●	3.8	7.7	mV
		$V_{CM} = V^-$	●	3.5	6.7	mV

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ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over $-40^{\circ}\text{C} < T_A < 85^{\circ}\text{C}$
 temperature range. $V_S = \pm 5\text{V}$; $V_{CM} = V_{OUT} = 0\text{V}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$V_{OS\ TC}$	Input Offset Voltage Drift (Note 9)	$V_{CM} = \text{Half Supply}$	●	7.5	24	$\mu\text{V}/^{\circ}\text{C}$
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)	$V_{CM} = 0\text{V}$ $V_{CM} = V^- \text{ to } V^+$	● ●	0.3 0.6	1.0 2.5	mV mV
I_B	Input Bias Current	$V_{CM} = \text{Half Supply}$	●	-7.0	-1.4	μA
		$V_{CM} = V^+$	●		1.8	μA
		$V_{CM} = V^-$	●	-10	-4.5	μA
ΔI_B	I_B Shift	$V_{CM} = V^- \text{ to } V^+$	●	5.4	13	μA
	I_B Match (Channel-to-Channel) (Note 6)		●	0.15	0.7	μA
I_{OS}	Input Offset Current	$V_{CM} = \text{Half Supply}$	●	0.15	1	μA
		$V_{CM} = V^+$	●	0.3	1.2	μA
		$V_{CM} = V^-$	●	0.5	1.6	μA
A_{VOL}	Large Signal Gain	$V_0 = \pm 4.5\text{V}$, $R_L = 1\text{k}$	●	60	110	V/mV
		$V_0 = \pm 1.5\text{V}$, $R_L = 100$	●	6.0	13	V/mV
CMRR	Common Mode Rejection Ratio	$V_{CM} = V^- \text{ to } V^+$	●	65	84	dB
		$V_{CM} = -2\text{V} \text{ to } 2\text{V}$	●	80	95	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.5\text{V} \text{ to } \pm 5\text{V}$	●	60	70	dB
		$V_S = \pm 1.5\text{V} \text{ to } \pm 5\text{V}$	●	70	100	dB
V_{OL}	Output Voltage Swing LOW Saturation (Note 8)	No Load	●	7	75	mV
		$I_{SINK} = 5\text{mA}$	●	98	205	mV
		$I_{SINK} = 15\text{mA}$	●	260	500	mV
V_{OH}	Output Voltage Swing HIGH Saturation (Note 8)	No Load	●	70	130	mV
		$I_{SOURCE} = 5\text{mA}$	●	130	250	mV
		$I_{SOURCE} = 15\text{mA}$	●	360	640	mV
I_{sc}	Short-Circuit Current		●	± 15	± 25	mA
I_S	Supply Current per Amp		●	3.8	4.5	mA
GBW	Gain Bandwidth Product	Frequency = 1MHz	●	90		MHz
SR	Slew Rate	$A_V = -1$, $R_L = 1\text{k}$, $V_0 = 4\text{V}$	●	13	18	$\text{V}/\mu\text{s}$
FPBW	Full Power Bandwidth (Note 10)	$V_{OUT} = 3\text{V}_{\text{P-P}}$	●	1.4	1.9	MHz

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: Inputs are protected by back-to-back diodes and diodes to each supply. If the inputs are taken beyond the supplies or the differential input voltage exceeds 0.7V, the input current must be limited to less than 40mA.

Note 3: A heat sink may be required to keep the junction temperature below the absolute maximum rating when the output is shorted indefinitely.

Note 4: The LT6202C/LT6202I, LT6203C/LT6203I and LT6204C/LT6204I are guaranteed functional over the temperature range of -40°C and 85°C .

Note 5: The LT6202C/LT6203C/LT6204C are guaranteed to meet specified performance from 0°C to 70°C . The LT6202C/LT6203C/LT6204C are designed, characterized and expected to meet specified performance from -40°C to 85°C , but are not tested or QA sampled at these temperatures.

The LT6202I/LT6203I/LT6204I are guaranteed to meet specified performance from -40°C to 85°C .

Note 6: Matching parameters are the difference between the two amplifiers A and D and between B and C of the LT6204; between the two amplifiers of the LT6203. CMRR and PSRR match are defined as follows: CMRR and PSRR are measured in $\mu\text{V}/\text{V}$ on the identical amplifiers. The difference is calculated between the matching sides in $\mu\text{V}/\text{V}$. The result is converted to dB .

Note 7: Minimum supply voltage is guaranteed by power supply rejection ratio test.

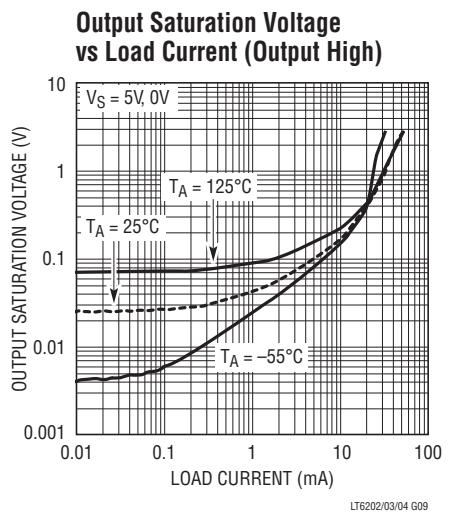
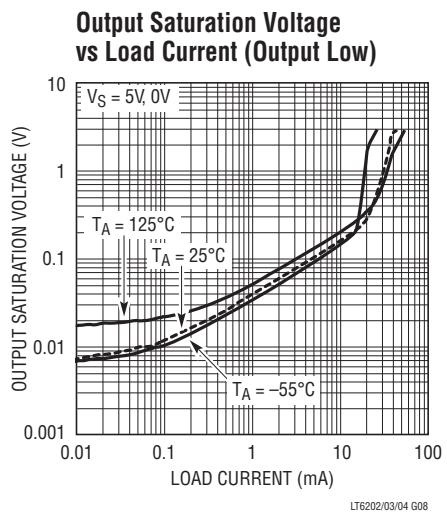
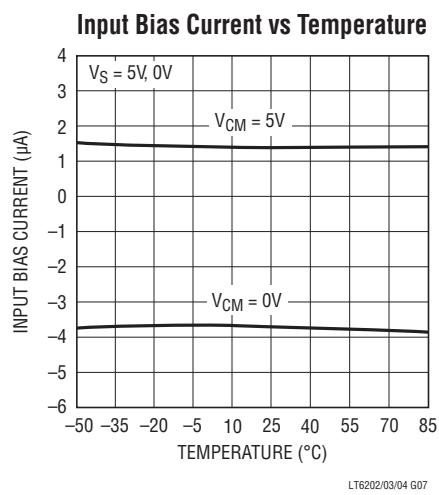
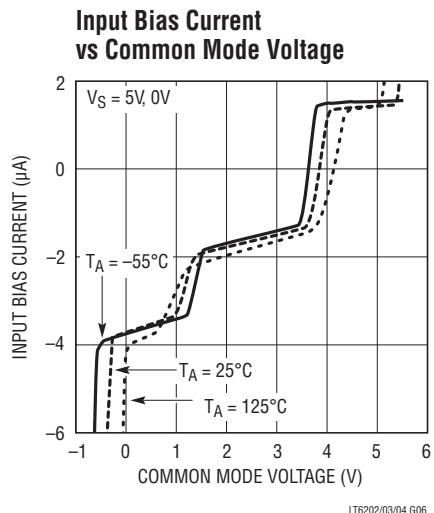
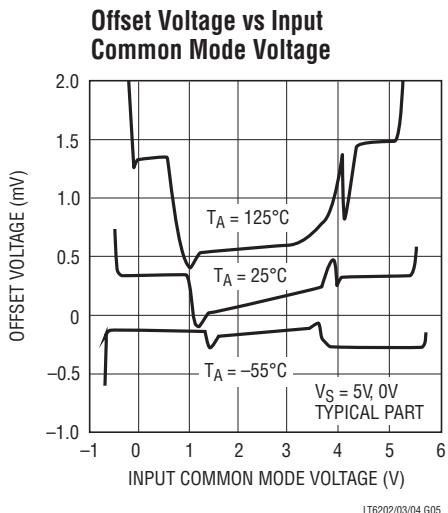
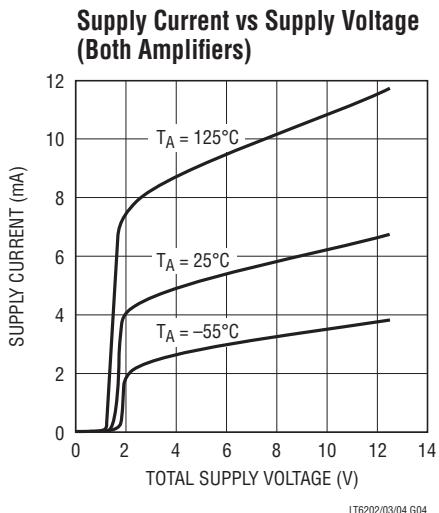
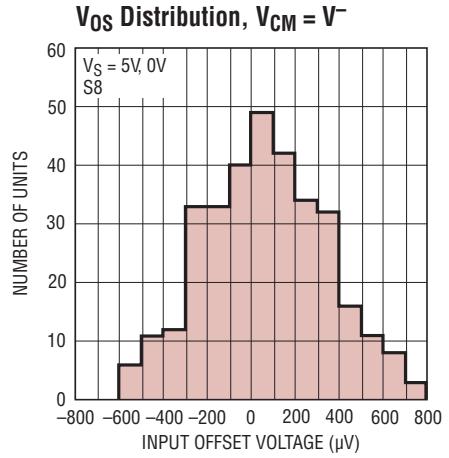
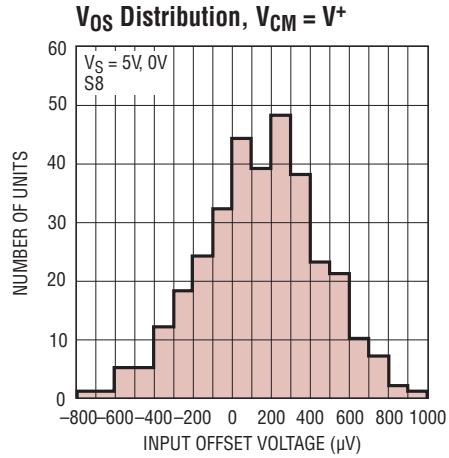
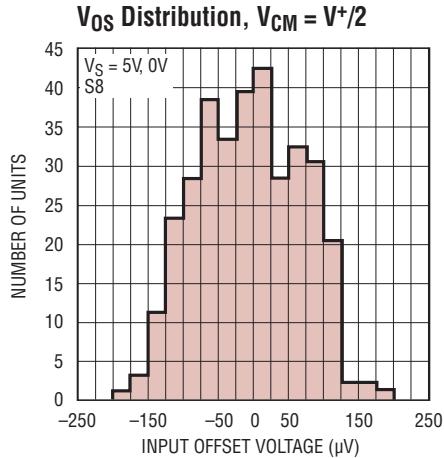
Note 8: Output voltage swings are measured between the output and power supply rails.

Note 9: This parameter is not 100% tested.

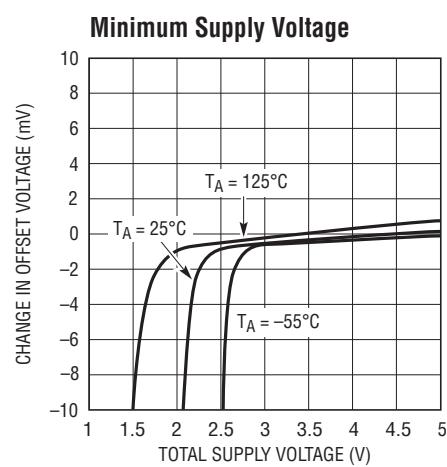
Note 10: Full-power bandwidth is calculated from the slew rate: $\text{FPBW} = \text{SR}/2\pi V_P$

Note 11: Differential gain and phase are measured using a Tektronix TSG120YC/NTSC signal generator and a Tektronix 1780R Video Measurement Set. The resolution of this equipment is 0.1% and 0.1°. Ten identical amplifier stages were cascaded giving an effective resolution of 0.01% and 0.01°.

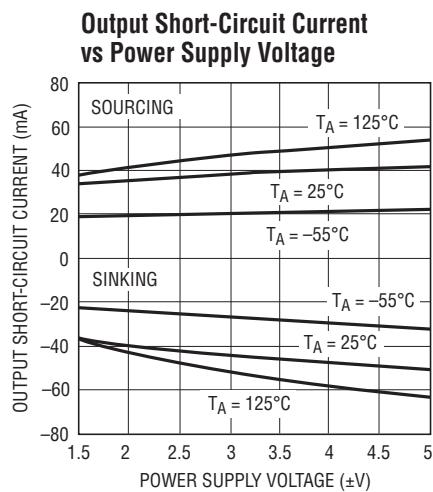
TYPICAL PERFORMANCE CHARACTERISTICS



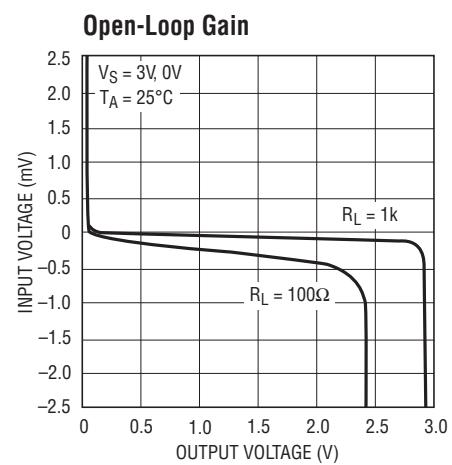
TYPICAL PERFORMANCE CHARACTERISTICS



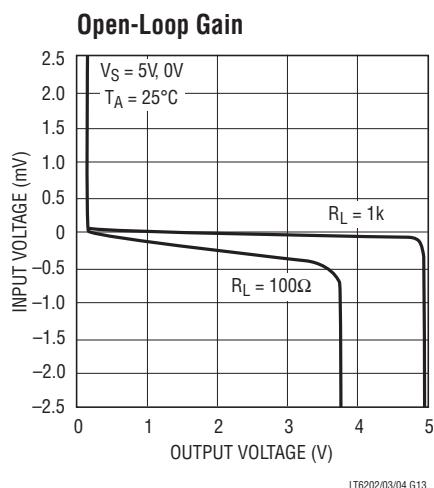
LT6202/03/04 G10



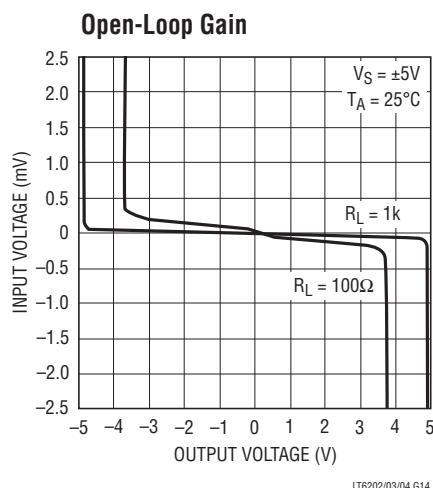
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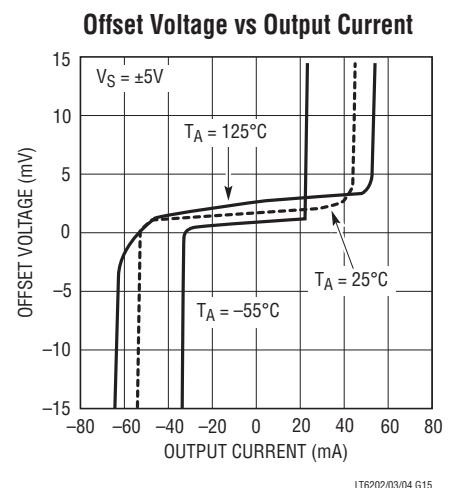
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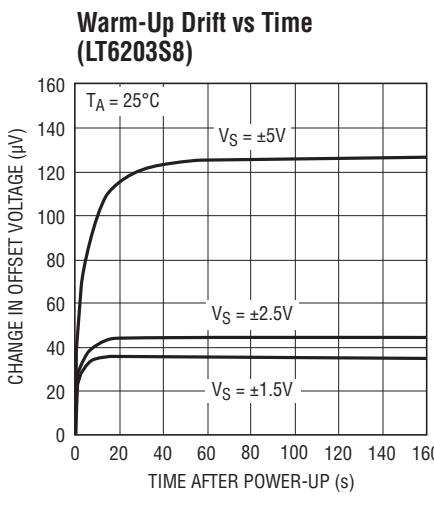
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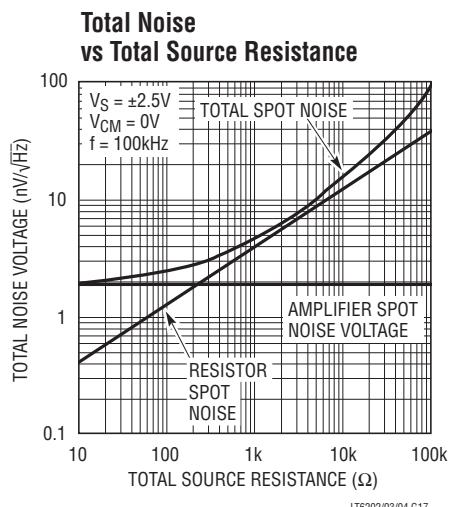
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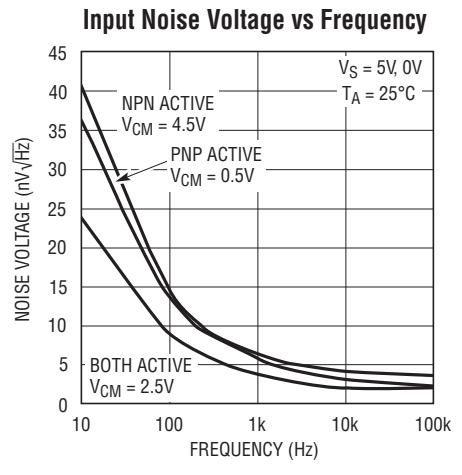
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LT6202/03/04 G16



LT6202/03/04 G17

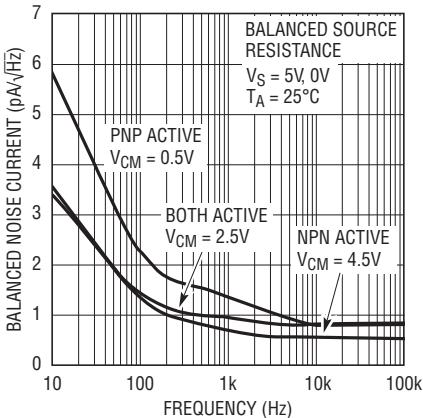


LT6202/03/04 G18

LT6202/LT6203/LT6204

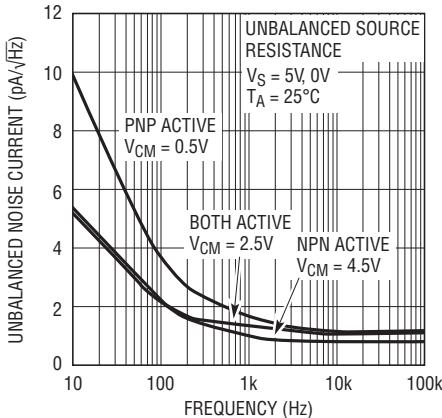
TYPICAL PERFORMANCE CHARACTERISTICS

Balanced Noise Current vs Frequency



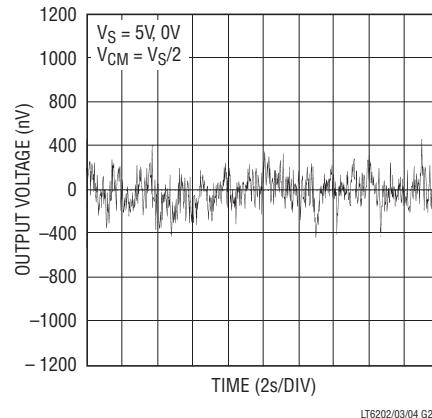
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Unbalanced Noise Current vs Frequency



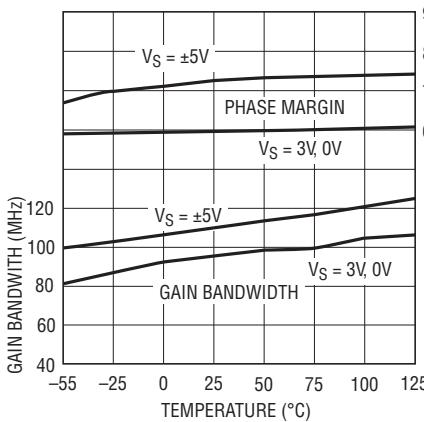
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0.1Hz to 10Hz Output Voltage Noise



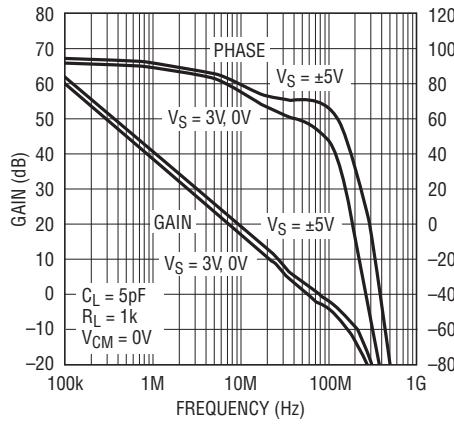
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Gain Bandwidth and Phase Margin vs Temperature



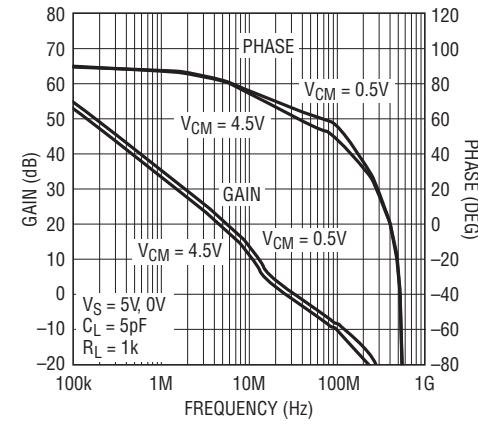
LT6202/03/04 G21

Open-Loop Gain vs Frequency



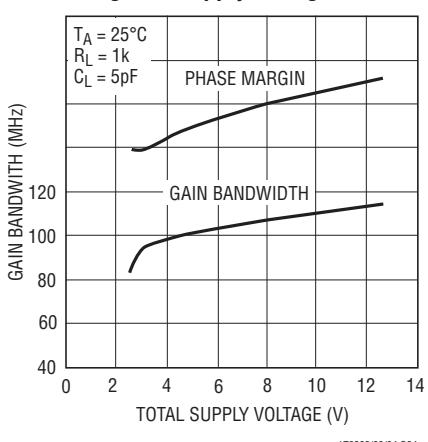
LT6202/03/04 G22

Open-Loop Gain vs Frequency



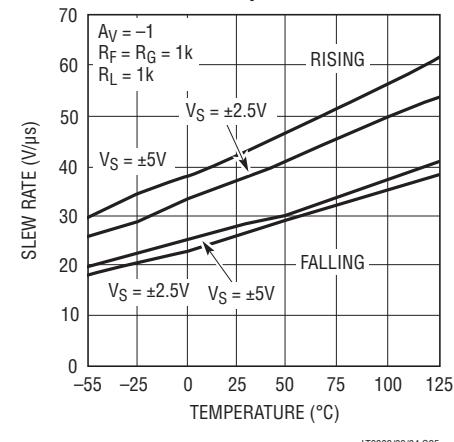
LT6202/03/04 G23

Gain Bandwidth and Phase Margin vs Supply Voltage



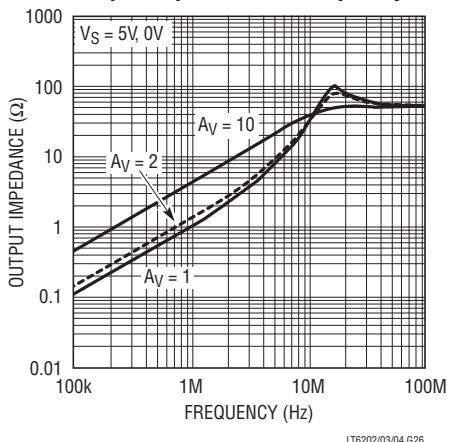
LT6202/03/04 G24

Slew Rate vs Temperature



LT6202/03/04 G25

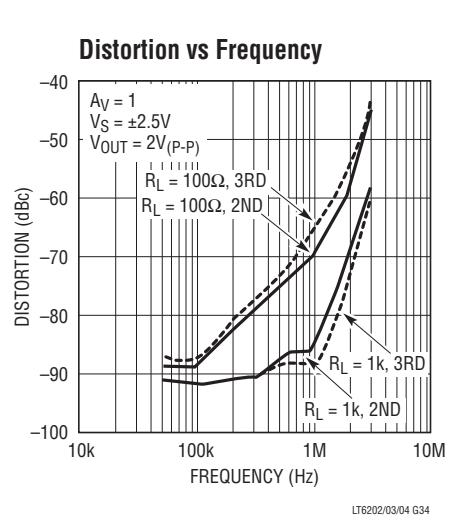
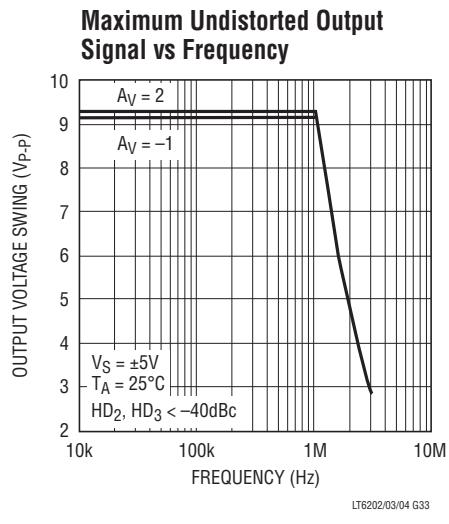
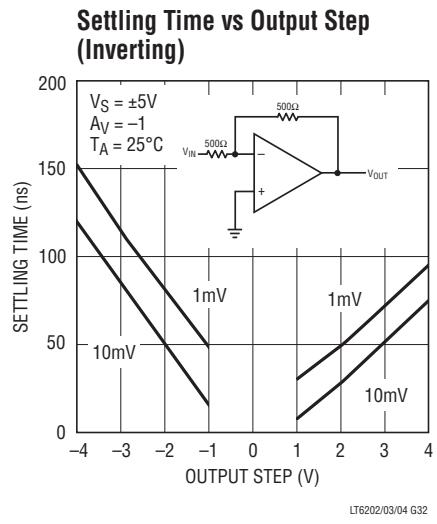
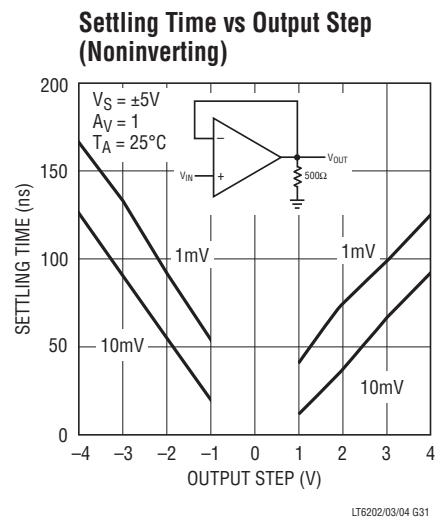
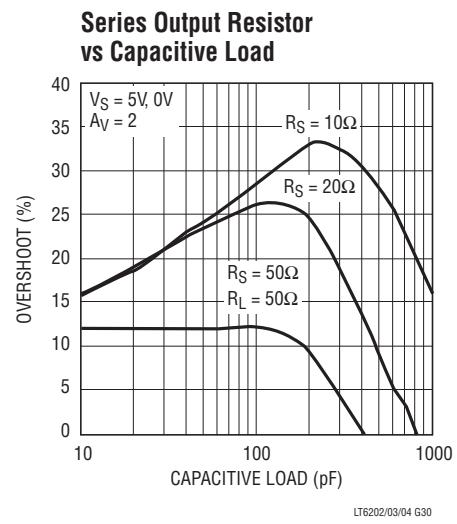
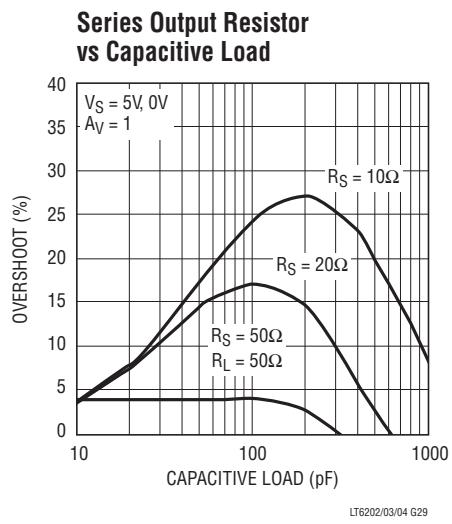
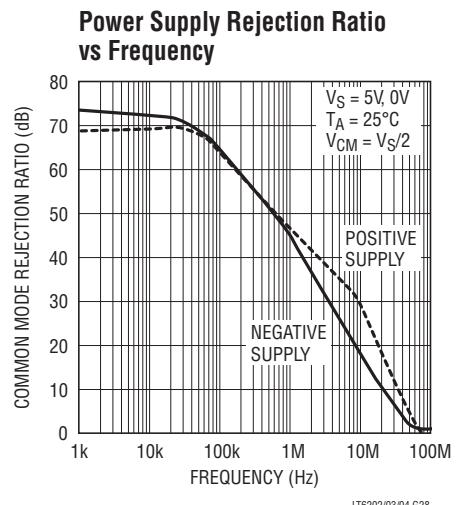
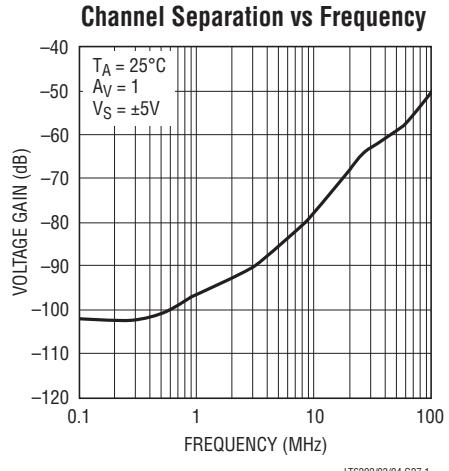
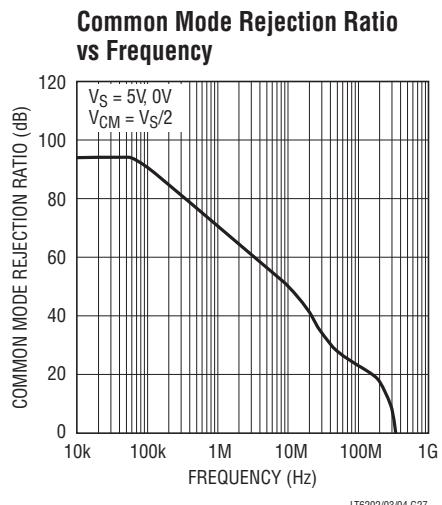
Output Impedance vs Frequency



LT6202/03/04 G26

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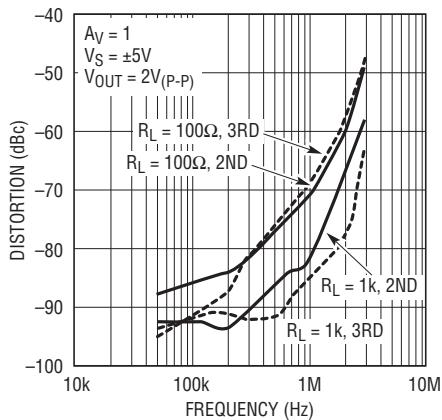
TYPICAL PERFORMANCE CHARACTERISTICS



LT6202/LT6203/LT6204

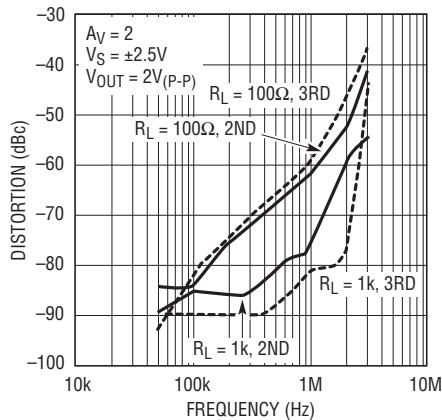
TYPICAL PERFORMANCE CHARACTERISTICS

Distortion vs Frequency



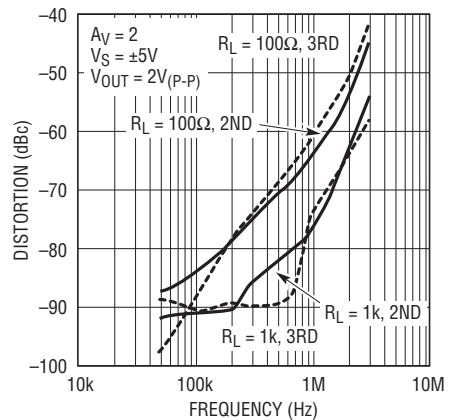
LT6202/03/04 G35

Distortion vs Frequency



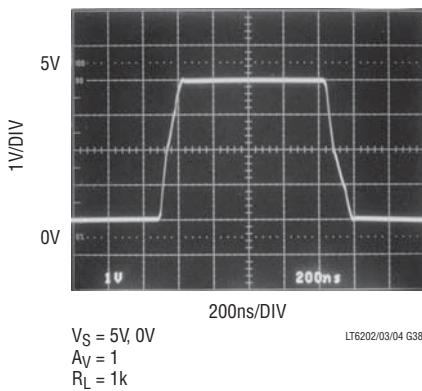
LT6202/03/04 G36

Distortion vs Frequency



LT6202/03/04 G37

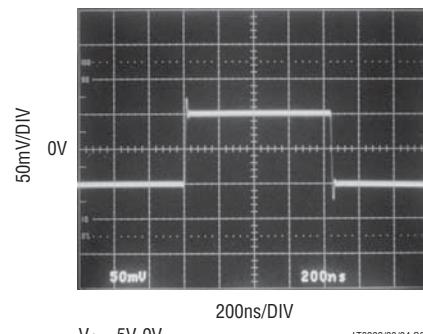
5V Large-Signal Response



$V_S = 5V, 0V$
 $A_V = 1$
 $R_L = 1k$

LT6202/03/04 G38

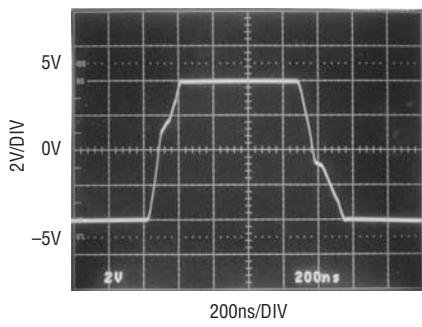
5V Small-Signal Response



$V_S = 5V, 0V$
 $A_V = 1$
 $R_L = 1k$

LT6202/03/04 G39

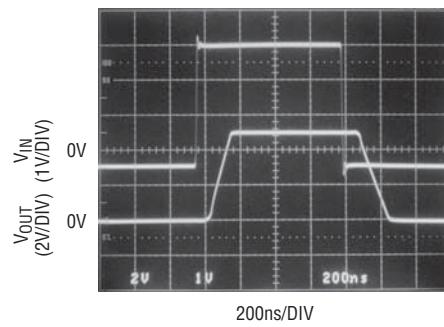
±5V Large-Signal Response



$V_S = \pm 5V$
 $A_V = 1$
 $R_L = 1k$

LT6202/03/04 G40

Output-Overdrive Recovery



$V_S = 5V, 0V$
 $A_V = 2$

LT6202/03/04 G41

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APPLICATIONS INFORMATION

Amplifier Characteristics

Figure 1 shows a simplified schematic of the LT6202/LT6203/LT6204, which has two input differential amplifiers in parallel that are biased on simultaneously when the common mode voltage is at least 1.5V from either rail. This topology allows the input stage to swing from the positive supply voltage to the negative supply voltage. As the common mode voltage swings beyond $V_{CC} - 1.5V$, current source I_1 saturates and current in Q1/Q4 is zero. Feedback is maintained through the Q2/Q3 differential amplifier, but with an input g_m reduction of 1/2. A similar effect occurs with I_2 when the common mode voltage swings within 1.5V of the negative rail. The effect of the g_m reduction is a shift in the V_{OS} as I_1 or I_2 saturate.

Input bias current normally flows out of the + and – inputs. The magnitude of this current increases when the input common mode voltage is within 1.5V of the negative rail, and only Q1/Q4 are active. The polarity of this current reverses when the input common mode voltage is within 1.5V of the positive rail and only Q2/Q3 are active.

The second stage is a folded cascode and current mirror that converts the input stage differential signals to a single ended output. Capacitor C1 reduces the unity cross frequency and improves the frequency stability without degrading the gain bandwidth of the amplifier. The differential drive generator supplies current to the output transistors that swing from rail-to-rail.

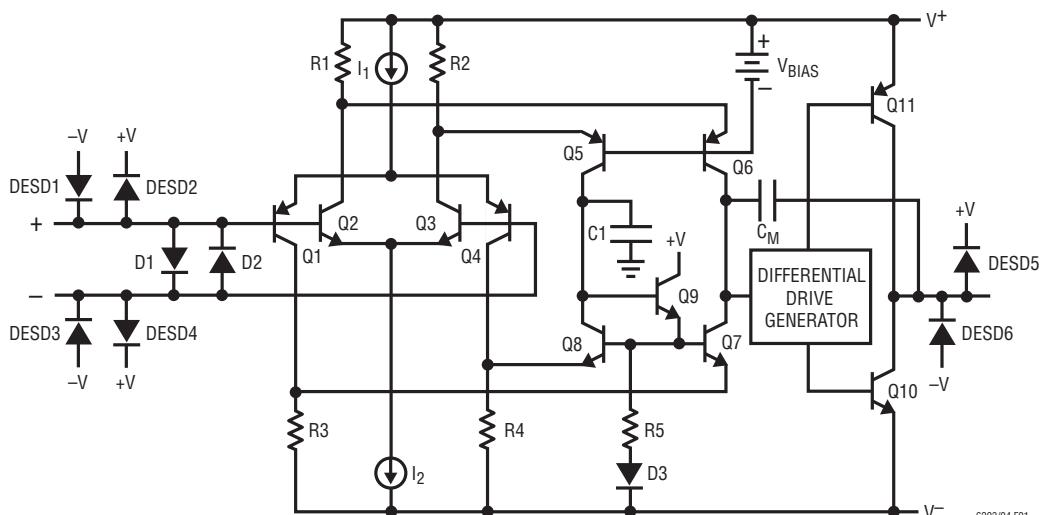


Figure 1. Simplified Schematic

APPLICATIONS INFORMATION

Input Protection

There are back-to-back diodes, D1 and D2, across the + and – inputs of these amplifiers to limit the differential input voltage to $\pm 0.7V$. The inputs of the LT6202/LT6203/LT6204 do not have internal resistors in series with the input transistors. This technique is often used to protect the input devices from over voltage that causes excessive currents to flow. The addition of these resistors would significantly degrade the low noise voltage of these amplifiers. For instance, a 100Ω resistor in series with each input would generate $1.8nV/\sqrt{Hz}$ of noise, and the total amplifier noise voltage would rise from $1.9nV/\sqrt{Hz}$ to $2.6nV/\sqrt{Hz}$. Once the input differential voltage exceeds $\pm 0.7V$, steady state current conducted through the protection diodes should be limited to $\pm 40mA$. This implies 25Ω of protection resistance per volt of continuous overdrive beyond $\pm 0.7V$. The input diodes are rugged enough to handle transient currents due to amplifier slew rate overdrive or momentary clipping without these resistors.

Figure 2 shows the input and output waveforms of the amplifier driven into clipping while connected in a gain of $A_V = 1$. When the input signal goes sufficiently beyond the power supply rails, the input transistors will saturate. When saturation occurs, the amplifier loses a stage of phase inversion and the output tries to change states. Diodes D1 and D2 forward bias and hold the output within

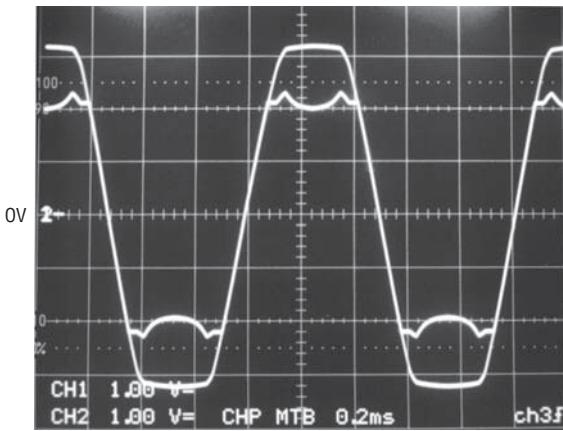


Figure 2. $V_S = \pm 2.5V$, $A_V = 1$ with Large Overdrive

a diode drop of the input signal. In this photo, the input signal generator is clipping at $\pm 35mA$, and the output transistors supply this generator current through the protection diodes.

With the amplifier connected in a gain of $A_V \geq 2$, the output can invert with very heavy input overdrive. To avoid this inversion, limit the input overdrive to $0.5V$ beyond the power supply rails.

ESD

The LT6202/LT6203/LT6204 have reverse-biased ESD protection diodes on all inputs and outputs as shown in Figure 1. If these pins are forced beyond either supply, unlimited current will flow through these diodes. If the current is transient and limited to one hundred milliamps or less, no damage to the device will occur.

Noise

The noise voltage of the LT6202/LT6203/LT6204 is equivalent to that of a 225Ω resistor, and for the lowest possible noise it is desirable to keep the source and feedback resistance at or below this value, i.e. $R_S + R_G||R_{FB} \leq 225\Omega$. With $R_S + R_G||R_{FB} = 225\Omega$ the total noise of the amplifier is: $e_n = \sqrt{(1.9nV)^2 + (1.9nV)^2} = 2.7nV$. Below this resistance value, the amplifier dominates the noise, but in the resistance region between 225Ω and approximately $10k\Omega$, the noise is dominated by the resistor thermal noise. As the total resistance is further increased, beyond $10k$, the noise current multiplied by the total resistance eventually dominates the noise.

The product of $e_n \cdot \sqrt{I_{SUPPLY}}$ is an interesting way to gauge low noise amplifiers. Many low noise amplifiers with low e_n have high I_{SUPPLY} current. In applications that require low noise with the lowest possible supply current, this product can prove to be enlightening. The LT6202/LT6203/LT6204 have an $e_n \cdot \sqrt{I_{SUPPLY}}$ product of 3.2 per amplifier, yet it is common to see amplifiers with similar noise specifications have an $e_n \cdot \sqrt{I_{SUPPLY}}$ product of 4.7 to 13.5.

For a complete discussion of amplifier noise, see the LT1028 data sheet.

TYPICAL APPLICATIONS

Low Noise, Low Power 1MΩ AC Photodiode Transimpedance Amplifier

Figure 3 shows the LT6202 applied as a transimpedance amplifier (TIA). The LT6202 forces the BF862 ultralow-noise JFET source to 0V, with R3 ensuring that the JFET has an I_{DRAIN} of 1mA. The JFET acts as a source follower, buffering the input of the LT6202 and making it suitable for the high impedance feedback elements R1 and R2. The BF862 has a minimum I_{DSS} of 10mA and a pinchoff voltage between $-0.3V$ and $-1.2V$. The JFET gate and the LT6202 output

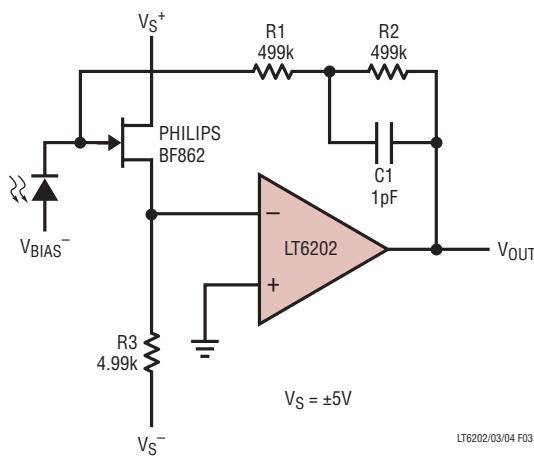


Figure 3. Low Noise, Low Power 1MΩ AC Photodiode Transimpedance Amplifier

therefore sit at a point slightly higher than one pinchoff voltage below ground (typically about $-0.6V$). When the photodiode is illuminated, the current must come from the LT6202's output through R1 and R2, as in a normal TIA. Amplifier input noise density and gain-bandwidth product were measured at 2.4nV/Hz and 100MHz , respectively. Note that because the JFET has a high g_m , approximately $1/80\Omega$, its attenuation looking into R3 is only about 2%. Gain-bandwidth product was measured at 100MHz and the closed-loop bandwidth using a 3pF photodiode was approximately 1.4MHz .

Precision Low Noise, Low Power, 1MΩ Photodiode Transimpedance Amplifier

Figure 4 shows the LT6202 applied as a transimpedance amplifier (TIA), very similar to that shown in Figure 3. In this case, however, the JFET is not allowed to dictate the DC-bias conditions. Rather than being grounded, the LT6202's noninverting input is driven by the LTC2050 to the exact state necessary for zero JFET gate voltage. The noise performance is nearly identical to that of the circuit in Figure 3, with the additional benefit of excellent DC performance. Input offset was measured at under $200\mu\text{V}$ and output noise was within $2\text{mV}_\text{P-P}$ over a 20MHz bandwidth.

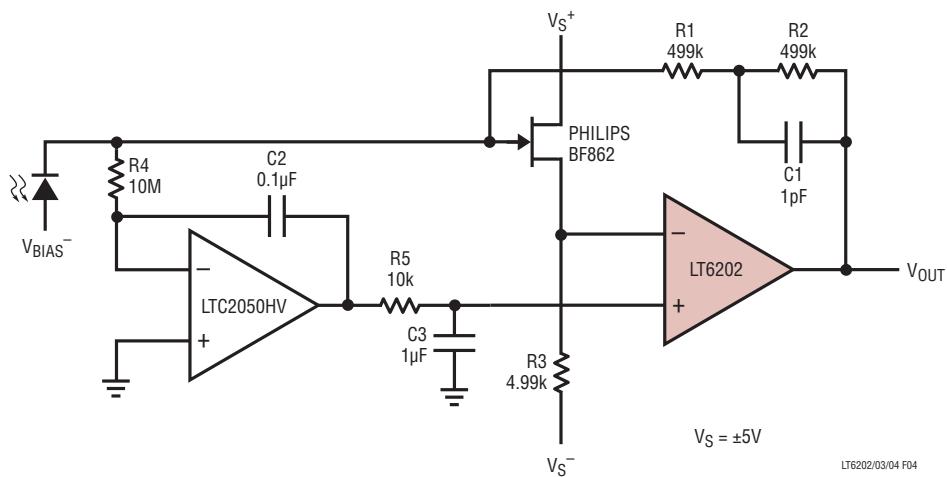


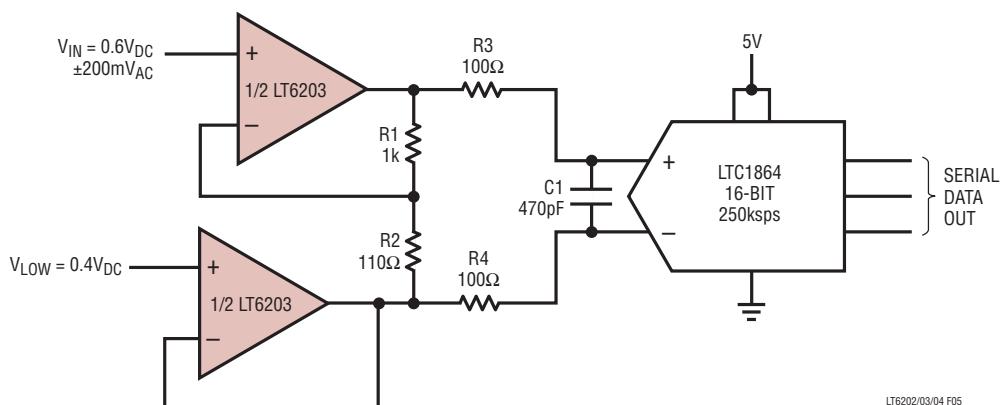
Figure 4. Precision Low Noise, Low Power Transimpedance Amplifier

TYPICAL APPLICATIONS

Single-Supply 16-Bit ADC Driver

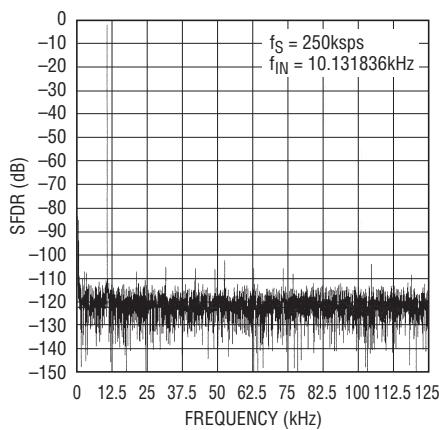
Figure 5 shows the LT6203 driving an LTC1864 unipolar 16-bit A/D converter. The bottom half of the LT6203 is in a gain-of-one configuration and buffers the 0V negative input of the LTC1864. The top half of the LT6203 is in a gain-of-ten configuration referenced to the buffered voltage V_{LOW} and drives the positive input of the LTC1864. The input range of the LTC1864 is 0V to 5V, but for best results the input range of V_{IN} should be from V_{LOW} (about 0.4V) to about 0.82V. Figure 6 shows an FFT obtained with a 10.1318kHz coherent input waveform, from 8192 samples with no windowing or averaging. Spurious free dynamic range is seen to be about 100dB.

Although the LTC1864 has a sample rate far below the gain bandwidth of the LT6203, using this amplifier is not necessarily a case of overkill. The designer is reminded that A/D converters have sample apertures that are vanishingly small (ideally, infinitesimally small) and make demands on the upstream circuitry far in excess of what is implied by the innocent-looking sample rate. In addition, when an A/D converter takes a sample, it applies a small capacitor to its inputs with a fair amount of glitch energy and expects the voltage on the capacitor to settle to the true value very quickly. Finally, the LTC1864 has a 20MHz analog input bandwidth and can be used in undersampling applications, again requiring a source bandwidth higher than Nyquist.



LT6202/03/04 F05

Figure 5. Single-Supply 16-Bit ADC Driver



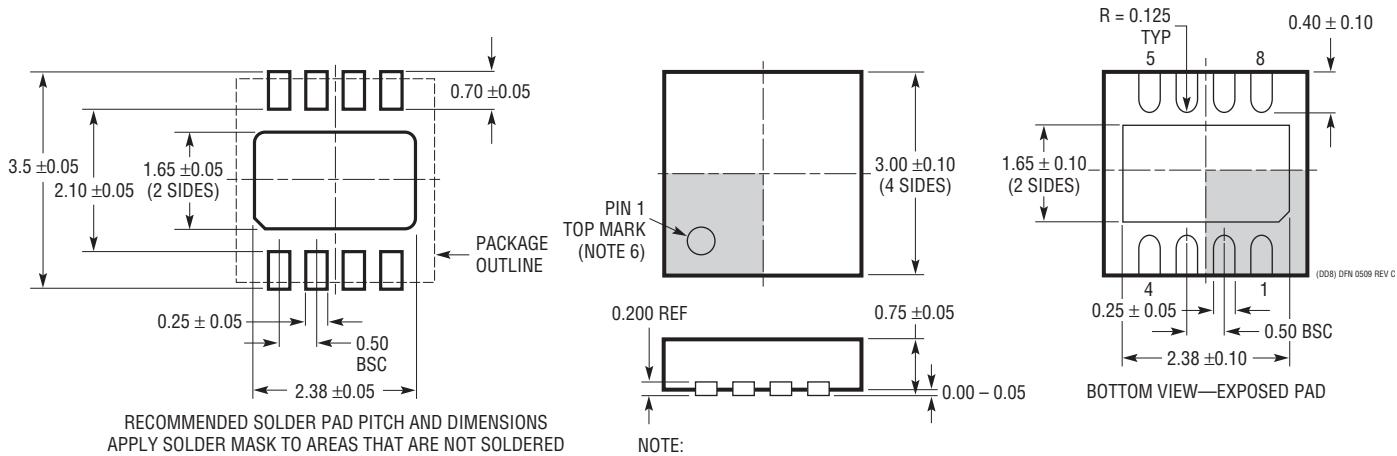
LT6202/03/04 F06

Figure 6. FFT Showing 100dB SFDR

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PACKAGE DESCRIPTION

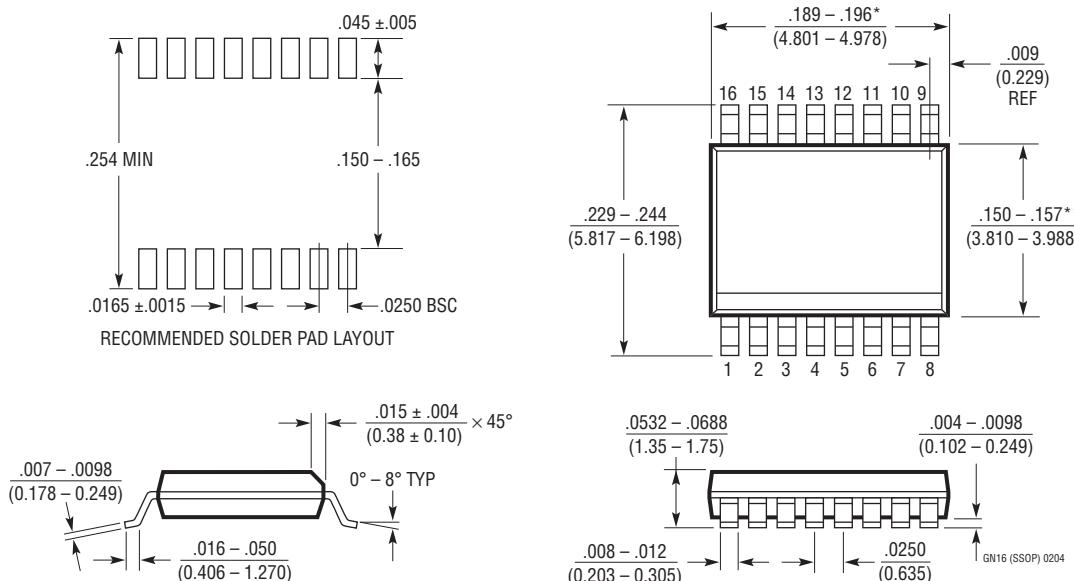
DD Package
8-Lead Plastic DFN (3mm × 3mm)
(Reference LTC DWG # 05-08-1698 Rev C)



NOTE:

1. DRAWING TO BE MADE A JEDEC PACKAGE OUTLINE MO-229 VARIATION OF (WEED-1)
2. DRAWING NOT TO SCALE
3. ALL DIMENSIONS ARE IN MILLIMETERS
4. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE
5. EXPOSED PAD SHALL BE SOLDER PLATED
6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON TOP AND BOTTOM OF PACKAGE

GN Package
16-Lead Plastic SSOP (Narrow .150 Inch)
(Reference LTC DWG # 05-08-1641)



NOTE:
1. CONTROLLING DIMENSION: INCHES
2. DIMENSIONS ARE IN INCHES
(MILLIMETERS)

3. DRAWING NOT TO SCALE

*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

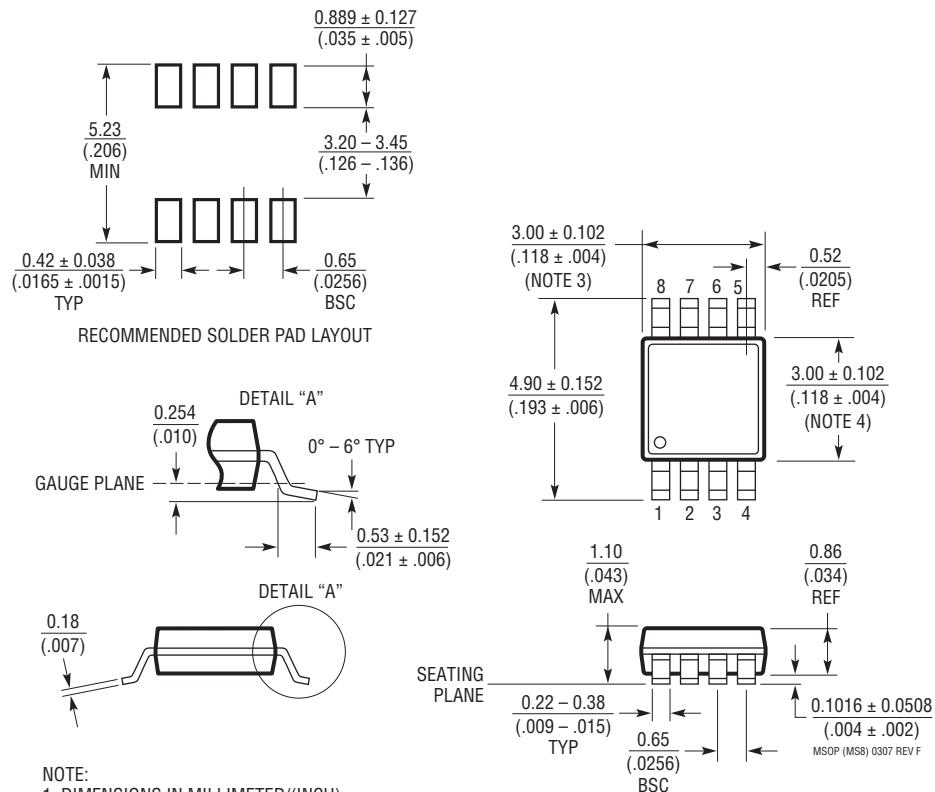
**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

PACKAGE DESCRIPTION

MS8 Package

8-Lead Plastic MSOP

(Reference LTC DWG # 05-08-1660 Rev F)

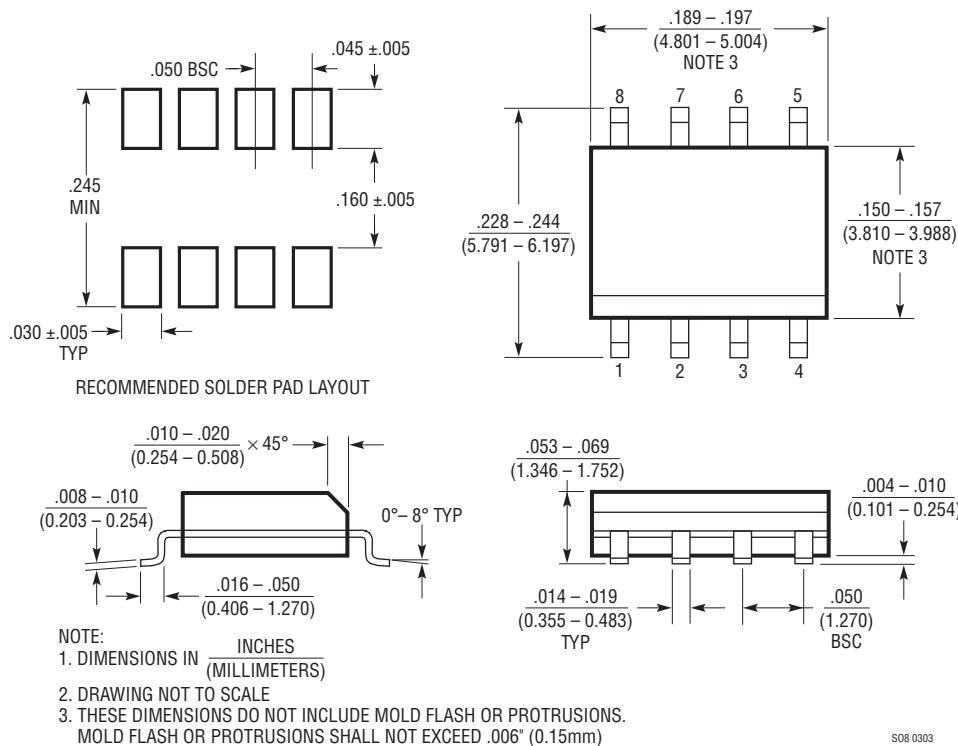


NOTE:

1. DIMENSIONS IN MILLIMETER/(INCH)
2. DRAWING NOT TO SCALE
3. DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm (.006") PER SIDE
4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006") PER SIDE
5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004") MAX

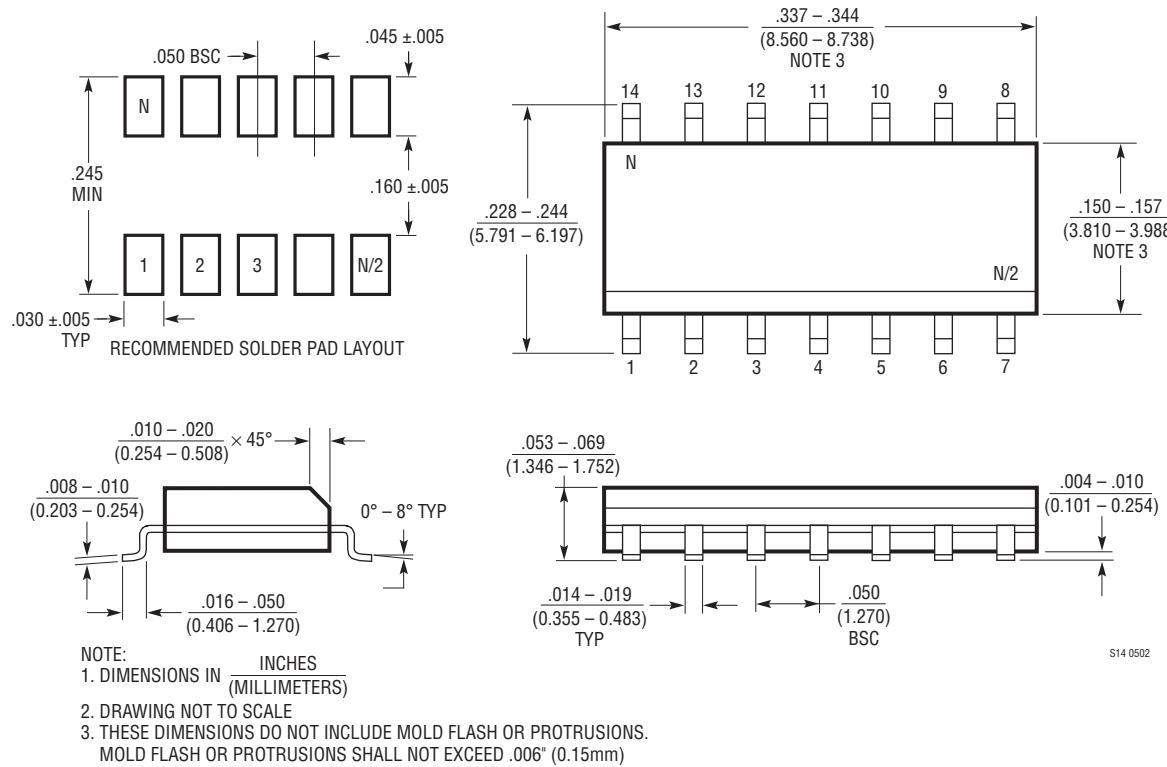
PACKAGE DESCRIPTION

S8 Package
8-Lead Plastic Small Outline (Narrow .150 Inch)
 (Reference LTC DWG # 05-08-1610)



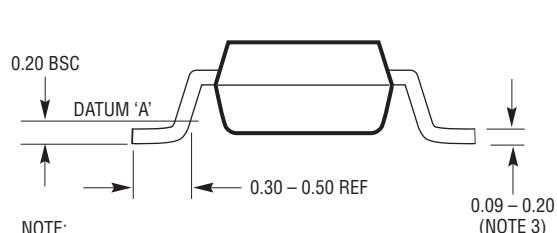
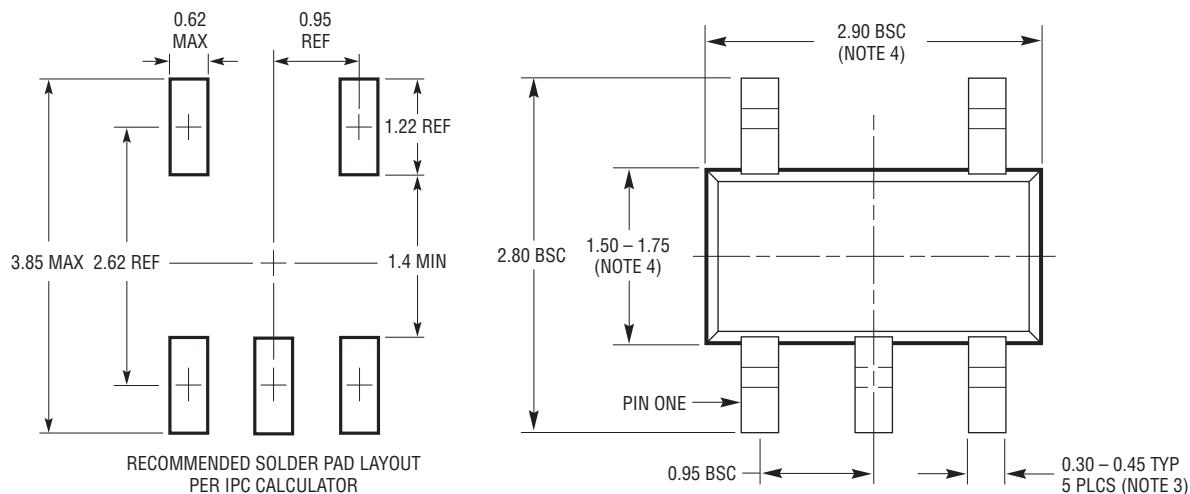
PACKAGE DESCRIPTION

S Package
14-Lead Plastic Small Outline (Narrow .150 Inch)
 (Reference LTC DWG # 05-08-1610)



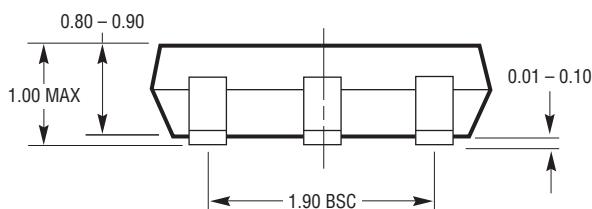
PACKAGE DESCRIPTION

S5 Package
5-Lead Plastic TSOT-23
(Reference LTC DWG # 05-08-1635)



NOTE:

1. DIMENSIONS ARE IN MILLIMETERS
2. DRAWING NOT TO SCALE
3. DIMENSIONS ARE INCLUSIVE OF PLATING
4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR
5. MOLD FLASH SHALL NOT EXCEED 0.254mm
6. JEDEC PACKAGE REFERENCE IS MO-193

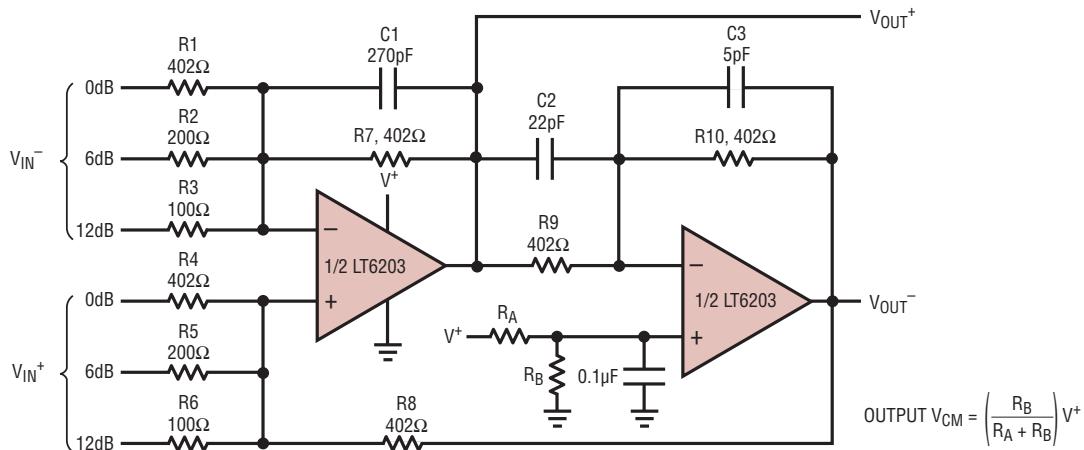


S5 TSOT-23 0302 REV B

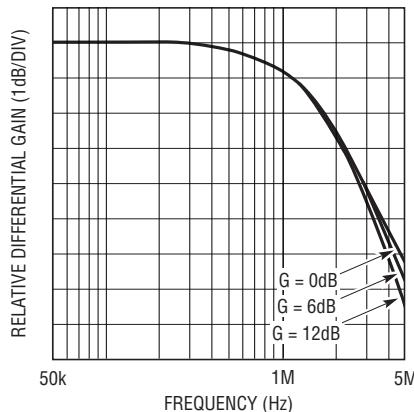
LT6202/LT6203/LT6204

TYPICAL APPLICATION

Low Noise Differential Amplifier with Gain Adjust and Common Mode Control



Low Noise Differential Amplifier Frequency Response



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1028	Single, Ultralow Noise 50MHz Op Amp	$1.1\text{nV}/\sqrt{\text{Hz}}$
LT1677	Single, Low Noise Rail-to-Rail Amplifier	3V Operation, 2.5mA, $4.5\text{nV}/\sqrt{\text{Hz}}$, $60\mu\text{V}$ Max V_{OS}
LT1722/LT1723/LT1724	Single/Dual/Quad Low Noise Precision Op Amps	$70\text{V}/\mu\text{s}$ Slew Rate, $400\mu\text{V}$ Max V_{OS} , $3.8\text{nV}/\sqrt{\text{Hz}}$, 3.7mA
LT1800/LT1801/LT1802	Single/Dual/Quad Low Power 80MHz Rail-to-Rail Op Amps	$8.5\text{nV}/\sqrt{\text{Hz}}$, 2mA Max Supply
LT1806/LT1807	Single/Dual, Low Noise 325MHz Rail-to-Rail Amplifiers	2.5V Operation, $550\mu\text{V}$ Max V_{OS} , $3.5\text{nV}/\sqrt{\text{Hz}}$
LT6200	Single Ultralow Noise Rail-to-Rail Amplifier	$0.95\text{nV}/\sqrt{\text{Hz}}$, 165MHz Gain Bandwidth

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