

General Description

The MAX4335-MAX4338 op amps deliver 40mW per channel into 32Ω from ultra-small SC70/SOT23 packages making them ideal for mono/stereo headphone drivers in portable applications. These amplifiers have a 5MHz gain-bandwidth product and are guaranteed to deliver 50mA of output current while operating from a single supply of 2.7V to 5.5V.

The MAX4336 and the MAX4338 have a shutdown/mute mode that reduces the supply current to 0.04µA per amplifier and places the outputs in a high-impedance

The MAX4335-MAX4338 have 90dB power-supply rejection ratio (PSRR), eliminating the need for costly pre-regulation in most audio applications. Both the input voltage range and the output voltage swing include both supply rails, maximizing dynamic range.

The MAX4335/MAX4336 single amplifiers are available in ultra-small 6-pin SC70 packages. The MAX4337/ MAX4338 dual amplifiers are available in an 8-pin SOT23 and a 10-pin µMAX package, respectively. All devices are specified from -40°C to +85°C.

Applications

 32Ω Headphone Drivers Portable/Battery-Powered Instruments Wireless PA Control Hands-Free Car Phones Transformer/Line Drivers DAC/ADC Buffers

Features

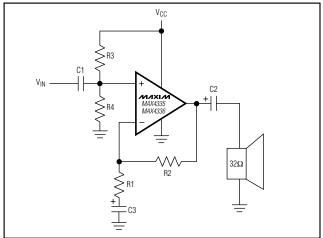
- ◆ 50mA Output Drive Capability
- ♦ Low 0.003% THD (20kHz into 10kΩ)
- ♦ Rail-to-Rail® Inputs and Outputs
- ♦ 2.7V to 5.5V Single-Supply Operation
- ♦ 5MHz Gain-Bandwidth Product
- ♦ 95dB Large-Signal Voltage Gain
- ♦ 90dB Power-Supply Rejection Ratio
- ♦ No Phase Reversal for Overdrive Inputs
- ♦ Ultra-Low Power Shutdown/Mute Mode Reduces Supply Current to 0.04µA Places Output in High-Impedance State
- **♦ Thermal Overload Protection**

Ordering Information

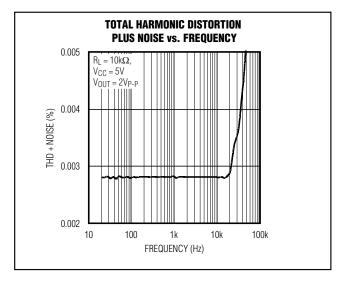
PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4335EXT-T	-40°C to +85°C	6 SC70-6	AAX
MAX4336EXT-T	-40°C to +85°C	6 SC70-6	AAW
MAX4337 EKA-T	-40°C to +85°C	8 SOT23-8	AAIK
MAX4337EUA	-40°C to +85°C	8 µMAX	_
MAX4338EUB	-40°C to +85°C	10 μMAX	

Pin Configurations appear at end of data sheet.

Typical Operating Circuit



Rail-to-Rail is a registered trademark of Nippon Motorola Ltd.



MIXIM

Maxim Integrated Products 1

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (VCC to GND).	0.3V to +6V
All Other Pins to GND	(GND - 0.3V) to (V _{CC} + 0.3V)
Output Short-Circuit Duration t	o V _{CC} or GNDContinuous
Continuous Power Dissipation	$(T_A = +70^{\circ}C)$
6-Pin SC70 (derate 3.1mW/°	C above +70°C)245mW
8-Pin SOT23 (derate 9.1mW/	°C above +70°C)727mW

8-Pin µMAX (derate 4.5mW/°C above +70°C)	362mW
10-Pin μMAX (derate 5.6mW/°C above +70°C)	444mW
Operating Temperature Range40	0°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range65°	°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = 2.7V, \text{GND} = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = \infty \text{ to } V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
Operating Supply Voltage Range	Vcc	Inferred from PSRR	Test	2.7		5.5	V	
Quiescent Supply Current (Per	1	$V_{CC} = 5.5V$			1.3	1.8	mΛ	
Amplifier)	Icc	$V_{CC} = 2.7$			1.2		mA	
Input Offset Voltage	Vos	$V_{CM} = GND$ to V_{CC}			±0.6	±3	mV	
Input Bias Current	lΒ	V _{CM} = GND to V _{CC}	:		±100	±400	nA	
Input Offset Current	los	$V_{CM} = GND$ to V_{CC}	:		±7	±30	nA	
Differential Input Resistance	Dave:	$ V_{IN-} - V_{IN+} < 1.2V$			500		kΩ	
Differential input nesistance	R _{IN(Diff)}	$ V_{IN-} - V_{IN+} > 1.2V$			8.4		K22	
Input Common-Mode Voltage Range	V _{CM}	Inferred from CMRF	R Test	GND		V _C C	V	
Common-Mode Rejection Ratio	CMRR	V _{CM} = GND to V _{CC}		60	80		dB	
Power-Supply Rejection Ratio	PSRR	$V_{CC} = 2.7V \text{ to } 5.5V$		70	90		dB	
Output Resistance	R _{OUT}	$AV_{CL} = 1V/V$			0.05		Ω	
			$V_{CC} = 5V: R_L = 10k\Omega$ $V_{OUT} = 0.4V \text{ to } 4.6V$		95			
Large-Signal Voltage Gain	Avol	$V_{CC} = 5V$: $R_L = 100\Omega$ $V_{OUT} = 0.5V$ to 4.5V		70	84		dB	
		$V_{CC} = 2.7V$: $R_L = 32\Omega$ $V_{OUT} = 0.5V$ to 2.2V		62	72			
		$V_{CC} = 2.7V;$	VCC - VOH		100			
		$R_L = 10k\Omega$	VoL		100			
		V _{CC} = 2.7V;	VCC - VOH		220	400		
Output Voltage Swing	\/o=	$R_L = 32\Omega$	VoL		280	400	mV	
Output Voltage Swing	Vout	$V_{CC} = 5V;$	V _{CC} - V _{OH}		100		IIIV	
		$R_L = 10k\Omega$	VoL		100			
		V _{CC} = 5V;	V _{CC} - V _{OH}		190	350		
		$R_L = 100\Omega$	VoL		240	350		

DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = 2.7V, GND = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = \infty \text{ to } V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

PARAMETER	SYMBOL	CONDITIO	NS	MIN	TYP	MAX	UNITS
		V _{CC} = 2.7V;	V _{CC} - V _{OH}		270	500	
Outrout Drive	la	ISOURCE, ISINK = 50mA	VoL		360	500	\ /
Output Drive	lout	V _{CC} = 5V;	V _{CC} - V _{OH}		270	500	mV
		ISOURCE, ISINK = 50mA	VoL		360	500	
Short-Circuit Current	Isc				110		mA
CLIDNI Logic Lovels	VIH	Normal mode		0.7 x V _{CC}			V
SHDN Logic Levels	V_{IL}	Shutdown mode				0.3 x V _{CC}	V
SHDN Leakage Current	Iμ	$V_{CC} = 5V$, $GND < V$	SHDN < VCC			0.5	μΑ
Output Leakage Current in Shutdown	I _{OUT} (SHDN)	$V_{CC} = 5V$, $V_{\overline{SHDN}} = 0$, $V_{OUT} = 0$, V_{CC}			0.01	0.5	μΑ
Shutdown Supply Current (Per Amplifier)	I _{CC(SHDN)}	SHDN = GND; V _{CC}		<0.04	0.5	μΑ	

DC ELECTRICAL CHARACTERISTICS

(VCC = 2.7V, GND = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = ∞ to V_{CC}/2, VSHDN = V_{CC}, **T_A = -40°C to +85°C**, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Operating Supply Voltage Range	Vcc	Inferred from PSRR test	2.7		5.5	V
Quiescent Supply Current (Per Amplifier)	Icc	V _{CC} = 5.5V			2.25	mA
Input Offset Voltage	Vos	V _{CM} = GND to V _{CC}			±6	mV
Input Bias Current	ΙΒ	V _{CM} = GND to V _{CC}			±600	nA
Input Offset Current	Ios	$V_{CM} = GND \text{ to } V_{CC}$			±60	nA
Input Common-Mode Voltage Range	V _{CM}	Inferred from CMRR test	GND		V _C C	V
Common-Mode Rejection Ratio	CMRR	V _{CM} = GND to V _{CC}	50			dB
Power-Supply Rejection Ratio	PSRR	V _{CC} = 2.7V to 5.5V	64			dB
Lorge Cignel Veltage Coin		$V_{CC} = 5V$: $R_L = 100\Omega$, $V_{OUT} = 0.6V$ to 4.4V	66			٩D
Large-Signal Voltage Gain	Avol	$V_{CC} = 2.7V$: $R_L = 32\Omega$, $V_{OUT} = 0.6V$ to 2.1V	56			dB

DC ELECTRICAL CHARACTERISTICS (continued)

(VCC = 2.7V, GND = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = ∞ to V_{CC}/2, V_{SHDN} = V_{CC}, **Ta = -40°C to +85°C**, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP MAX	UNITS	
		V _{CC} = 2.7V;	Vcc - Voh		500		
Output Voltage Swing	V 0	$R_L = 32\Omega$	V _{OL}		500	1 ,	
Output Voltage Swing	Vout	$V_{CC} = 5V;$	VCC - VOH		400	mV	
		$R_L = 100\Omega$	V _{OL}		400		
Output Drive		V _{CC} = 2.7V;	V _{CC} - V _{OH}		650	mV	
	lout	ISOURCE, ISINK = 50mA	V _{OL}		650		
		V _{CC} = 5V; I _{SOURCE} , I _{SINK} = 50mA	VCC - VOH		650		
			VoL		650		
CLIDNI Lagia Laval	VIH	Normal mode		0.7 x V _C C		V	
SHDN Logic Level	VIL	Shutdown mode			0.3 x V _{CC}	V	
SHDN Leakage Current	IIL	V _{CC} = 5V, GND <	V SHDN < VCC		1	μΑ	
Output Leakage Current in Shutdown	I _{OUT(SHDN)}	V _{CC} = 5V, V _{SHDN} = 0, V _{OUT} = 0; V _{CC}			1	μΑ	
Shutdown Supply Current (Per Amplifier)	I _{CC(SHDN)}	VSHDN = 0; VCC =	5V		1	μΑ	

AC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = 2.7V, GND = 0, V_{CM} = V_{CC}/2, V_{OUT} = V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, A_{VCL} = 1V/V, C_L = 15pF, R_L = \infty \text{ to } V_{CC}/2, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

PARAMETER	SYMBOL	CONDIT	MIN TYP	MAX	UNITS	
Gain-Bandwidth Product	GBWP			5		MHz
Full-Power Bandwidth	FBWP	$V_{OUT} = 2V_{P-P}, V_{CC} = 5V_{P-P}$	1	280		kHz
Slew Rate	SR			1.8		V/µs
Phase Margin	PM			70		degrees
Gain Margin	GM			18		dB
		$V_{CC} = 5V$, $R_L = 100\Omega$,	f = 1kHz	0.005		
		V _{OUT} = 2V _{P-P}	f = 10kHz	0.02		
Total Harmonic Distortion	THD	$V_{CC} = 5V$, $R_L = 10k\Omega$, $V_{OUT} = 2V_{P-P}$, $f = 10kHz$		0.003		%
		V _{CC} = 2.7V;	f = 1kHz	0.01		
		$R_L = 32\Omega,$ $V_{OUT} = 2V_{P-P}$	f = 10kHz	0.03		

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AC ELECTRICAL CHARACTERISTICS (continued)

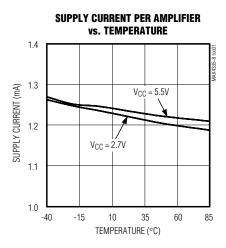
 $(V_{CC} = +2.7V, GND = 0, V_{CM} = V_{CC}/2, V_{OUT} = V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, A_{VCL} = 1V/V, C_L = 15pF, R_L = \infty \text{ to } V_{CC}/2, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

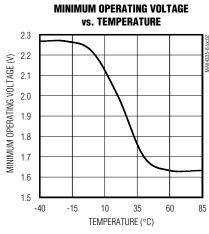
PARAMETER	SYMBOL	CONDITIONS	MIN TYP	MAX	UNITS	
Settling Time to 0.01%	ts	2V step	2		μs	
Crosstalk	CT	$V_{OUT} = 2V_{P-P}$; $f = 1kHz$	100		dB	
Input Capacitance	CIN		5		pF	
Input Voltage-Noise Density	0:-	f = 10kHz	26		nV/√ Hz	
	en	f = 1kHz			IIV/VMZ	
Input Current Naige Density		f = 10kHz	0.6		pA/√ Hz	
Input Current-Noise Density	In	f = 1kHz			μ <i>Α</i> γνπΖ	
Capacitive-Load Stability		No sustained oscillation	200		pF	
Shutdown Time	t <u>shdn</u>		1		μs	
Enable Time from Shutdown	tenable		1		μs	
Power-Up Time	tON		5		μs	

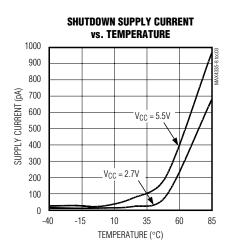
Note 1: All devices are 100% production tested at T_A = +25°C. All limits over temperature are guaranteed by design.

Typical Operating Characteristics

 $(V_{CC}=2.7V, GND=0, V_{CM}=0, V_{OUT}=V_{CC}/2, R_L=\infty \ to \ V_{CC}/2, V_{\overline{SHDN}}=V_{CC}, T_A=+25^{\circ}C, \ unless \ otherwise \ noted.)$

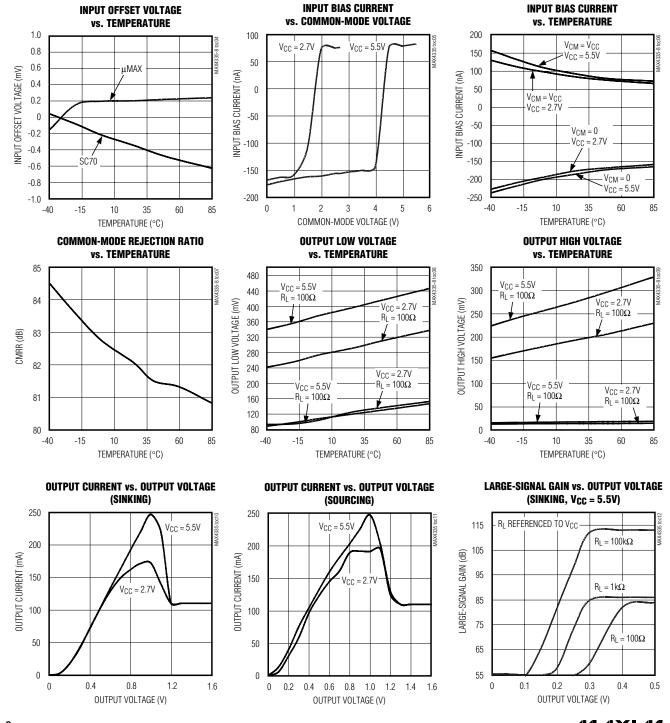






Typical Operating Characteristics (continued)

 $(V_{CC} = 2.7V, GND = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = \infty \text{ to } V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

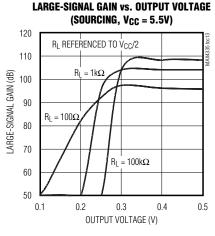


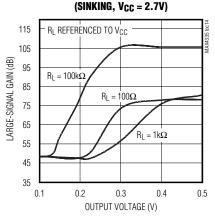
MAX4335-MAX4338

SC70/SOT23-8, 50mA IOUT, Rail-to-Rail I/O Op Amps with Shutdown/Mute

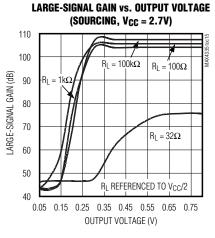
Typical Operating Characteristics (continued)

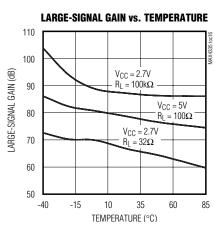
 $(V_{CC} = 2.7V, GND = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = \infty \text{ to } V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

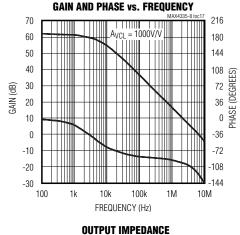


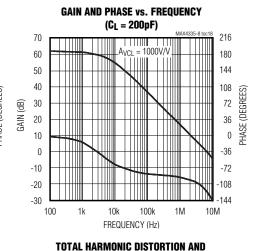


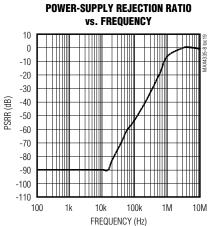
LARGE-SIGNAL GAIN vs. OUTPUT VOLTAGE

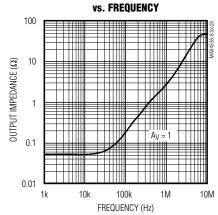


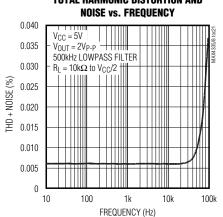






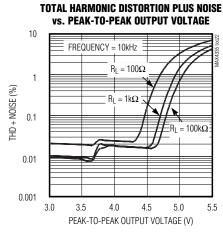


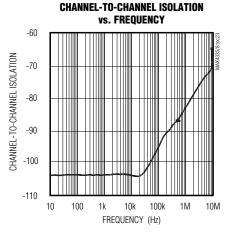


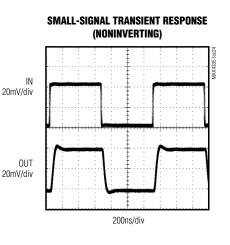


Typical Operating Characteristics (continued)

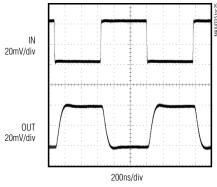
 $(V_{CC} = 2.7V, GND = 0, V_{CM} = 0, V_{OUT} = V_{CC}/2, R_L = \infty \text{ to } V_{CC}/2, V_{\overline{SHDN}} = V_{CC}, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$



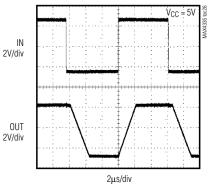




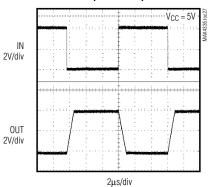
SMALL-SIGNAL TRANSIENT RESPONSE (INVERTING)



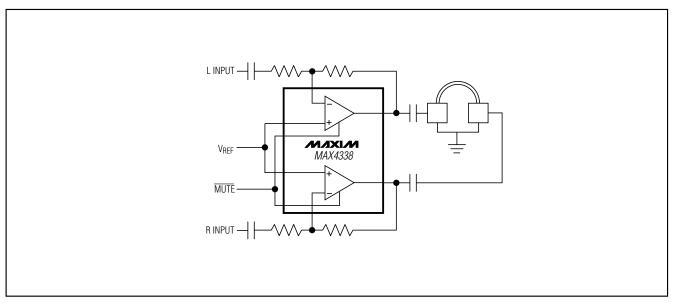




LARGE-SIGNAL TRANSIENT RESPONSE (INVERTING)



Typical Application Circuit



Pin Description

		PIN				
MAY 4005	MAYADOC	MAX4337		MAY4000	NAME	FUNCTION
MAX4335	MAX4336	SOT23	μМΑХ	MAX4338		
1	1	3, 5	3, 5	3, 7	IN1 ⁺ , IN2 ⁺	Noninverting Input
2	2	4	4	4	GND	Ground
3	3	2, 6	2, 6	2, 8	IN2 ⁻ , IN2 ⁻	Inverting Input
4	4	1, 7	1, 7	1, 9	OUT1, OUT2	Output(s)
5	_	_	_	_	N.C.	No Connection. Not internally connected.
_	5	_	_	5, 6	SHDN1, SHDN2	Drive SHDN low for shutdown. Drive SHDN high or connect to V _{CC} for normal operation.
6	6	8	8	10	V _{CC}	Positive Supply

Applications Information

Package Power Dissipation

Warning: Due to the high-output-current drive, this op amp can exceed the absolute maximum power-dissipation rating. As a general rule, as long as the peak current is less than or equal to 50mA, the maximum package power dissipation will not be exceeded for any of the package types offered. There are some exceptions to this rule, however. The absolute maximum power-dissipation rating of each package should always be verified using the following equations. The following equation gives an approximation of the package power dissipation:

$$P_{IC(DISS)} \cong V_{RMS} I_{RMS} COS \theta$$

where: V_{RMS} = the RMS voltage from V_{CC} to V_{OUT} when sourcing current

= the RMS voltage from Vout to VEE when sinking current

I_{RMS} = the RMS current flowing out of or into the op amp and the load

 θ = the phase difference between the voltage and the current. For resistive loads, COS θ = 1.

For example, the circuit in Figure 1 has a package power dissipation of 220mW.

$$\begin{split} V_{RMS} &\cong \left(V_{CC} - V_{DC}\right) - \frac{V_{PEAK}}{\sqrt{2}} \\ &= 5.5V - 2.75V - \frac{1V}{\sqrt{2}} = 2.043V_{RMS} \\ I_{RMS} &\cong I_{DC} + \frac{I_{PEAK}}{\sqrt{2}} = \frac{2.75V}{32\Omega} + \frac{1V/32\Omega}{\sqrt{2}} \\ &= 108\text{mA}_{RMS} \end{split}$$

Therefore, $P_{IC(DISS)} = V_{RMS} I_{RMS} COS \theta = 220 mW$ Adding a coupling capacitor improves the package power dissipation because there is no DC current to

the load, as shown in Figure 2.

$$V_{RMS} \cong \left(V_{CC} - V_{DC}\right) - \frac{V_{PEAK}}{\sqrt{2}}$$

$$= 5.5V - 2.75V - \frac{1V}{\sqrt{2}} = 2.043V_{RMS}$$

$$I_{RMS} \cong I_{DC} + \frac{I_{PEAK}}{\sqrt{2}} = 0A + \frac{1V/32\Omega}{\sqrt{2}}$$

$$= 22mA_{RMS}$$

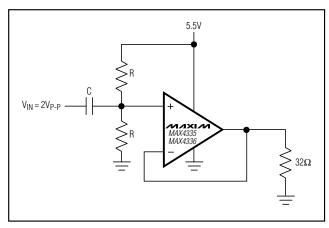


Figure 1. A Circuit Example where the MAX4335/MAX4336 is Dissipating High Power

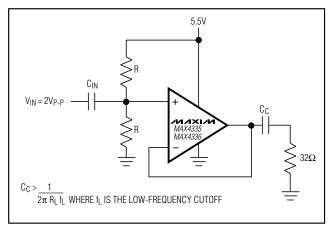


Figure 2. A Circuit Example where Adding a Coupling Capacitor Greatly Reduces the Power Dissipation of Its Package

Therefore,
$$P_{IC(DISS)} = V_{RMS} I_{RMS} COS \theta$$

= 45mW

The absolute maximum power-dissipation rating of the package may be exceeded if the configuration in Figure 1 is used with the MAX4335/MAX4336 amplifiers at a high ambient temperature of 79°C (220.6mW/°C plus a derating of 3.1mW/°C x 9°C = 247.9mW). Note that the 247.9mW just exceeds the absolute maximum power dissipation of 245mW for the 6-pin SC70 package.

Single-Supply Speaker Driver

The MAX4335/MAX4336 can be used as a single-supply speaker driver, as shown in the *Typical Operating Circuit*. Capacitor C1 is used for blocking DC (a 0.1µF ceramic capacitor can be used). When choosing resistors R3 and R4, take into consideration the input bias current as well as how much supply current can be tolerated. Choose resistors R1 and R2 according to the amount of gain and current desired. Capacitor C3 ensures unity gain for DC. A $10\mu\text{F}$ electrolytic capacitor is suitable for most applications. The coupling capacitor C2 sets a low-frequency pole and is fairly large in value. For a 32Ω load, a $100\mu\text{F}$ coupling capacitor gives a low-frequency pole at 50Hz. The low-frequency pole can be set according to the following equation:

$$f = 1 / 2\pi (R_LC2)$$

Rail-to-Rail Input Stage

Devices in the MAX4335–MAX4338 family of high-output-current amplifiers have rail-to-rail input and output stages designed for low-voltage, single-supply operation. The input stage consists of separate NPN and PNP differential stages that combine to provide an input common-mode range that extends 0.25V beyond the supply rails. The PNP stage is active for input voltages close to the negative rail, and the NPN stage is active for input voltages near the positive rail. The switchover transition region, which occurs near VCC/2, has been extended to minimize the slight degradation in common-mode rejection ratio caused by mismatch of the input pairs.

Since the input stage switches between the NPN and PNP pairs, the input bias current changes polarity as the input voltage passes through the transition region. Match the effective impedance seen by each input to reduce the offset error caused by input bias currents flowing through external source impedances (Figures 3 and 5).

High source impedances, together with input capacitance, can create a parasitic pole that produces an underdamped signal response. Reducing the input impedance or placing a small (2pF to 10pF) capacitor across the feedback resistor improves response.

The MAX4335–MAX4338's inputs are protected from large differential input voltages by $1 \mathrm{k}\Omega$ series resistors and back-to-back double diodes across the inputs (Figure 5).

For differential voltages less than 1.2V, input resistance is typically $500k\Omega$. For differential input voltages greater than 1.2V, input resistance is approximately $8.4k\Omega$. The input bias current is given by the following equation:

 $I_{BIAS} = (V_{DIFF} - 1.2V) / 8.4k\Omega$

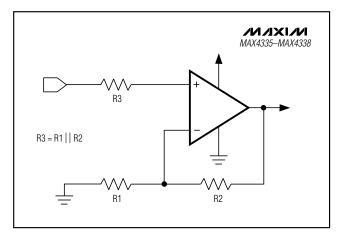


Figure 3. Reducing Offset Error Due to Bias Current (Noninverting)

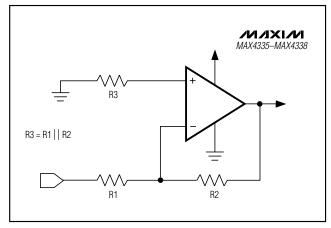


Figure 4. Reducing Offset Error Due to Bias Current (Inverting)

Rail-to-Rail Output Stage

The minimum output is within millivolts of ground for single-supply operation, where the load is referenced to ground (GND). Figure 6 shows the input voltage range and the output voltage swing of a MAX4335 connected as a voltage follower. The maximum output voltage swing is load dependent; however, it is guaranteed to be within 400mV of the positive rail (VCC = 2.7V) even with maximum load (32 Ω to VCC/2).

Driving Capacitive Loads

The MAX4335–MAX4338 have a high tolerance for capacitive loads. They are stable with capacitive loads up to 200pF. Figure 7 is a graph of the stable operating region for various capacitive loads vs. resistive loads.

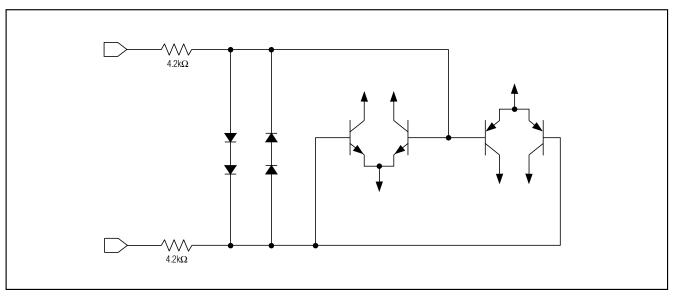


Figure 5. Input Protection Circuit

Figures 8 and 9 show the transient response with excessive capacitive loads (330pF), with and without the addition of an isolation resistor in series with the output. Figure 10 shows a typical noninverting capacitive-load-driving circuit in the unity-gain configuration. The resistor improves the circuit's phase margin by isolating the load capacitor from the op amp's output.

Power-Up and Shutdown/Mute Modes

The MAX4336/MAX4338 have a shutdown option. When the shutdown pin $\overline{(SHDN)}$ is pulled low, supply current drops to 0.04µA per amplifier (V_{CC} = 5V), the amplifiers are disabled, and their outputs are placed in a high-impedance state. Pulling \overline{SHDN} high enables the amplifier. In the dual MAX4338, the two amplifiers shut down independently. Figure 11 shows the MAX4336's output voltage response to a shutdown pulse. The MAX4335–MAX4338 typically settle within 5µs after power-up (Figure 12).

Power Supplies and Layout

The MAX4335–MAX4338 can operate from a single 2.7V to 5.5V supply. Bypass the power supply with a 0.1 μ F ceramic capacitor in parallel with at least 1 μ F. Good layout improves performance by decreasing the amount of stray capacitance at the op amps' inputs and outputs. Decrease stray capacitance by placing external components close to the op amps' input/output pins, minimizing trace and lead lengths.

Thermal Overload Protection

The MAX4335–MAX4338 includes thermal overload protection circuitry. When the junction temperature of the device exceeds +140°C, the supply current drops to 120 μ A per amplifier (V_{CC} = 5V) and the outputs are placed in a high-impedance state. The device returns to normal operation when the junction temperature falls to below +120°C.

Short-Circuit Current Protection

The MAX4335–MAX4338 incorporate a smart short-circuit protection feature. Figure 7 shows the output voltage region where the protection circuitry is active. A fault condition occurs when IOUT > 110mA and VOUT > 1V (sinking current) or when IOUT > 110mA and (VCC - VOUT) > 1V (sourcing current). When a fault is detected, the short-circuit protection circuitry is activated and the output current is limited to 110mA, protecting the device and the application circuitry. When the smart short circuit is not active, the output current can safely exceed 110mA (see the Output Current vs. Output Voltage Graph in the *Typical Operating Characteristics*).

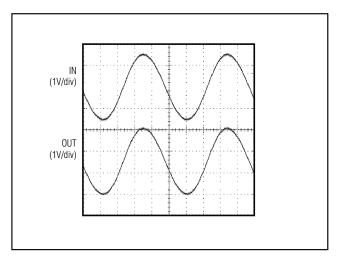


Figure 6. Rail-to-Rail Input/Output Range

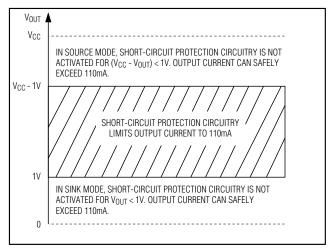


Figure 7. Short-Circuit Protection

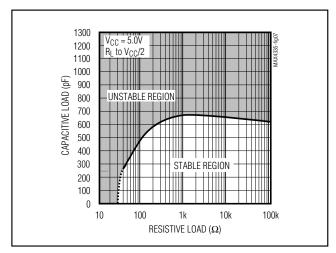


Figure 8. Capacitive-Load Stability

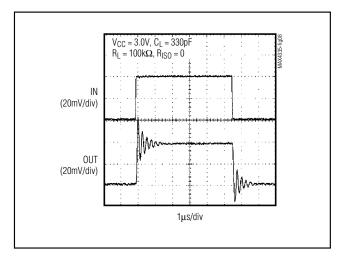


Figure 9. Small-Signal Transient Response with Excessive Capacitive Load

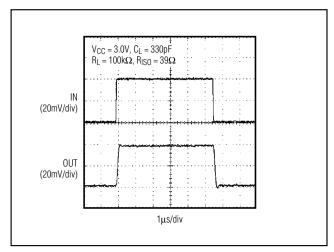


Figure 10. Small-Signal Transient Response with Excessive Capacitive Load with Isolation Resistor

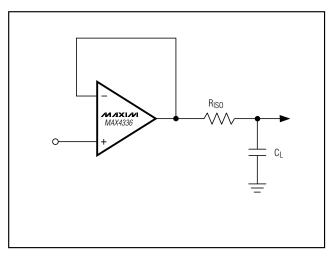


Figure 11. Capacitive-Load-Driving Circuit

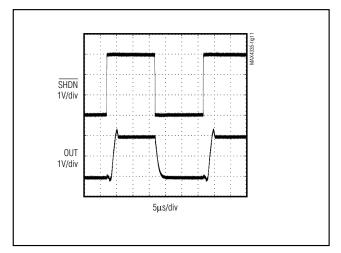


Figure 12. Shutdown Output Voltage Enable/Disable

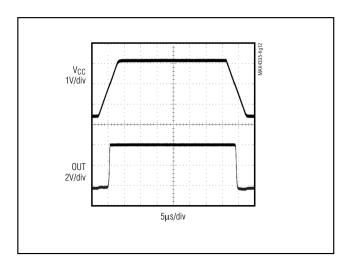
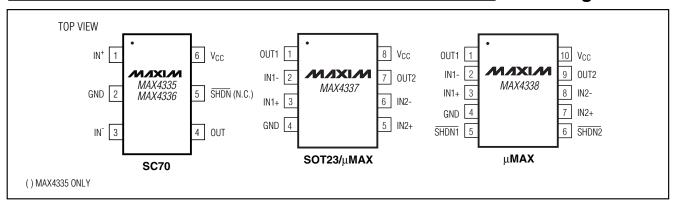


Figure 13. Power-Up/Down Output Voltage

Pin Configurations



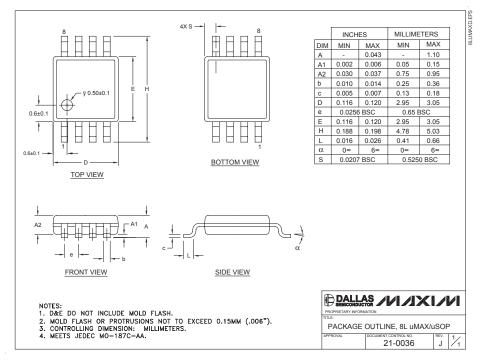
Chip Information

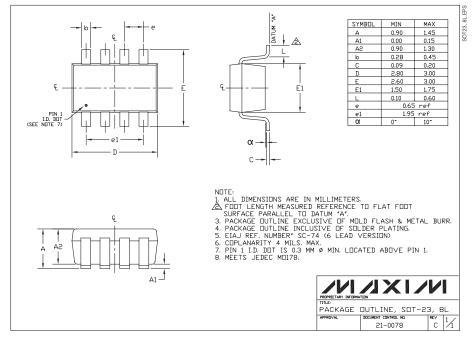
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PROCESS: BICMOS

Package Information

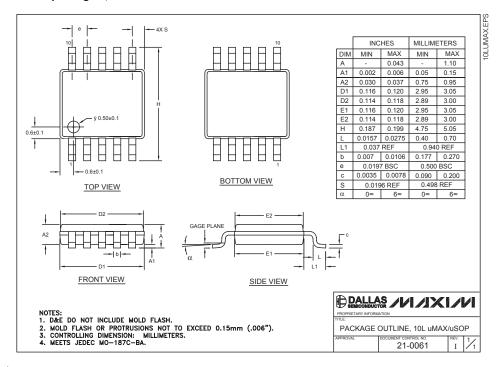
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