

NCV8665

150 mA Very Low Iq Low Dropout Linear Regulator with Reset and Delay Reset

The NCV8665 is a precision 5.0 V fixed output, low dropout integrated voltage regulator with an output current capability of 150 mA. Careful management of light load current consumption, combined with a low leakage process, achieve a typical quiescent ground current of 30 μ A.

NCV8665 is pin for pin compatible with the NCV8675 and the NCV4275 and it could replace this part when lower output current, and very low quiescent current is required.

The output voltage is accurate within $\pm 2.0\%$, and maximum dropout voltage is 600 mV at full rated load current.

It is internally protected against 45 V input transients, input supply reversal, output overcurrent faults, and excess die temperature. No external components are required to enable these features.

Features

- 5 V Fixed Output (3.3 V and 2.5 V Versions are Also Available)
- $\pm 2.0\%$ Output Accuracy, Over Full Temperature Range
- 40 μ A Maximum Quiescent Current at $I_{OUT} = 100 \mu$ A
- 600 mV Maximum Dropout Voltage at 150 mA Load Current
- Wide Input Voltage Operating Range of 5.5 V to 45 V
- Internal Fault Protection
 - ◆ -42 V Reverse Voltage
 - ◆ Short Circuit
 - ◆ Thermal Overload
- NCV Prefix for Automotive and Other Applications Requiring Site and Control Changes
- These are Pb-Free Devices

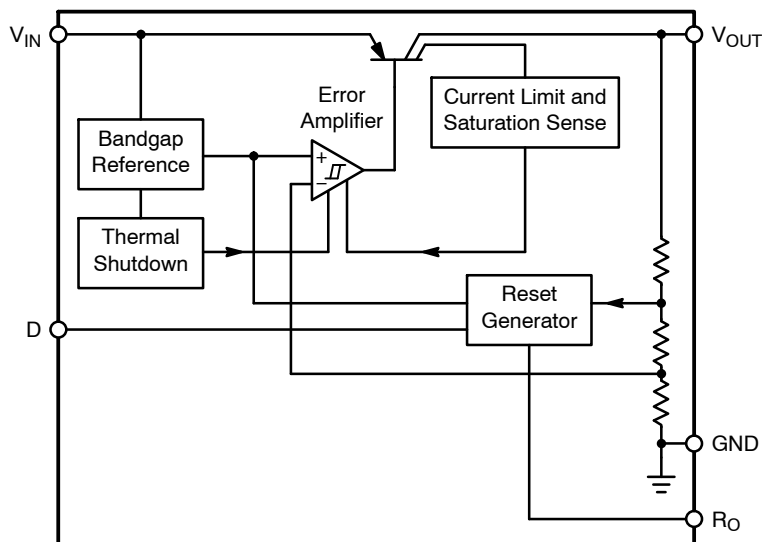


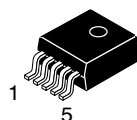
Figure 1. Block Diagram



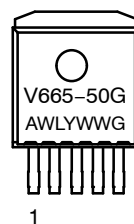
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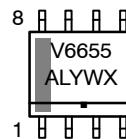
MARKING DIAGRAMS



D²PAK
5-PIN
DS SUFFIX
CASE 936A



SOIC-8
D SUFFIX
CASE 751



A = Assembly Location
WL, L = Wafer Lot
Y = Year
WW, W = Work Week
G or \blacksquare = Lead Free Indicator

PIN CONNECTIONS

D²PAK

| | |
|------|--------------|
| Pin | 1. V_{IN} |
| | 2. R_O |
| Tab, | 3. GND* |
| | 4. D |
| | 5. V_{OUT} |

* Tab is connected to Pin 3

SOIC-8

| | |
|-----|--------------|
| Pin | 1. V_{IN} |
| | 2. R_O |
| | 3. D |
| | 4. V_{OUT} |
| | 5-8. GND |

ORDERING INFORMATION

See detailed ordering and shipping information in the dimensions section on page 9 of this data sheet.

PIN DESCRIPTIONS

| Symbol | Function |
|------------------|--|
| V _{IN} | Unregulated input voltage; 5.5 V to 45 V; Battery Input Voltage. Bypass to GND with a 0.1 μ F ceramic capacitor. |
| R _O | Reset Output; open collector active Reset (Accurate when V _{OUT} > 1.0 V) |
| GND | Ground; Pin 3 internally connected to Tab |
| D | Reset Delay; timing capacitor to GND for Reset Delay function |
| V _{OUT} | Output; $\pm 2.0\%$, 150 mA. 10 μ F, ESR < 16 Ω |

ABSOLUTE MAXIMUM RATINGS

| Pin Symbol, Parameter | Symbol | Min | Max | Unit |
|---|--------------------|------|------|------|
| V _{IN} , DC Input Voltage | V _{IN} | -42 | +45 | V |
| V _{OUT} , DC Voltage | V _{OUT} | -0.3 | +16 | V |
| Reset Output Voltage | V _{RO} | -0.3 | 25 | V |
| Reset Output Current | I _{RO} | -5.0 | 5.0 | mA |
| Reset Delay Voltage | V _D | -0.3 | 7.0 | V |
| Reset Delay Current | I _D | -2.0 | 2.0 | mA |
| Storage Temperature | T _{STG} | -55 | +150 | °C |
| ESD Capability, Human body Model (Note 1) | V _{ESDHB} | 4000 | | V |
| ESD Capability, Machine Model (Note 1) | V _{ESDMM} | 200 | | V |
| Moisture Sensitivity Level | MSL | 1 | | - |

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

- This device series incorporates ESD protection and is tested by the following methods:
 ESD Human Body Model (HBM) tested per AEC-Q100-002 (EIA/JESD22-A 114C)
 ESD Machine Model (MM) tested per AEC-Q100-003 (EIA/JESD22-A 115C)
- Latchup Current Maximum Rating: ≤ 100 mA per JEDEC standard: JESD78.

OPERATING RANGE

| Pin Symbol, Parameter | Symbol | Min | Max | Unit |
|-------------------------------|-----------------|-----|-----|------|
| Input Voltage Operating Range | V _{IN} | 5.5 | 45 | V |
| Junction Temperature | T _J | -40 | 150 | °C |

THERMAL RESISTANCE

| Parameter | | Symbol | Min | Max | Unit |
|------------------------------|--------------------|-------------------|-----|------|------|
| Junction to Ambient (Note 3) | D ² PAK | R _{θJA} | - | 85.4 | °C/W |
| Junction to Case (Note 3) | D ² PAK | R _{θJC} | - | 6.8 | |
| Junction to Ambient (Note 4) | SOIC-8 | R _{θJA} | - | 138 | |
| Junction to Lead 6 (Note 4) | SOIC-8 | Ψ _{θJL6} | - | 21 | |

- As mounted on a 35x35x1mm FR4 PCB with a single layer of 100 mm² of 1 oz copper heat spreading area.
- As mounted on a 35x35x1mm FR4 PCB with a single layer of 100 mm² of 1 oz copper heat spreading area including traces directly connected to the leads.

Pb SOLDERING TEMPERATURE AND MSL

| Parameter | Symbol | Min | Max | Unit |
|---|------------------|-----|--------|------|
| Lead Temperature Soldering Reflow (SMD styles only), Pb-Free (Note 5) | T _{sld} | - | 265 pk | °C |
| MSL, 8-Lead EP, LS Temperature 260°C | MSL | 1 | | - |

- This device series incorporates ESD protection and exceeds the following ratings:
 Human Body Model (HBM) ≤ 2.0 kV per JEDEC standard: JESD22-A114.
 Machine Model (MM) ≤ 200 V per JEDEC standard: JESD22-A115.

ELECTRICAL CHARACTERISTICS $V_{IN} = 13.5\text{ V}$, $T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$, unless otherwise specified

| Parameter | Symbol | Test Conditions | Min | Typ | Max | Unit |
|--------------------------------|-------------------------------------|---|-------|------------|------------|---------------------------|
| OUTPUT | | | | | | |
| Output Voltage | V_{OUT} | $0.1\text{ mA} \leq I_{OUT} \leq 150\text{ mA}$ (Note 6) $6\text{ V} \leq V_{IN} \leq 28\text{ V}$ | 4.900 | 5.000 | 5.100 | V |
| Output Voltage | V_{OUT} | $0\text{ mA} \leq I_{OUT} \leq 150\text{ mA}$ $5.5\text{ V} \leq V_{IN} \leq 28\text{ V}$ $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$ | 4.900 | 5.000 | 5.100 | V |
| Line Regulation | ΔV_{OUT} versus V_{IN} | $I_{OUT} = 5\text{ mA}$ $8\text{ V} \leq V_{IN} \leq 32\text{ V}$ | -25 | 5 | +25 | mV |
| Load Regulation | ΔV_{OUT} Vs. I_{OUT} | $1\text{ mA} \leq I_{OUT} \leq 150\text{ mA}$ (Note 6) | -35 | 5 | +35 | mV |
| Dropout Voltage | $V_{IN} - V_{OUT}$ | $I_{OUT} = 100\text{ mA}$ (Notes 6 and 7) $I_{OUT} = 150\text{ mA}$ (Notes 6 and 7) | | 200 250 | 500 600 | mV |
| Quiescent Current | I_q | $I_{OUT} = 100\mu\text{A}$ $T_J = 25^\circ\text{C}$ $T_J = -40^\circ\text{C}$ to $+85^\circ\text{C}$ | | 30 30 | 34 40 | μA |
| Active Ground Current | $I_{G(ON)}$ | $I_{OUT} = 50\text{ mA}$ (Note 6) $I_{OUT} = 150\text{ mA}$ (Note 6) | | 1.8 12 | 3.5 19 | mA |
| Power Supply Rejection | PSRR | $V_{RIPPLE} = 0.5\text{ V}_{PP}$, $F = 100\text{ Hz}$ | | 69 | | %/V |
| Output Capacitor for Stability | C_{OUT} ESR | $I_{OUT} = 0.1\text{ mA}$ to 150 mA | 10 | | 16 | μF Ω |

RESET TIMING D AND OUTPUT RO

| | | | | | | |
|------------------------------|--------------|---|------|------|------|---------------|
| Reset Switching Threshold | $V_{OUT,rt}$ | - | 4.50 | 4.65 | 4.80 | V |
| Reset Output Low Voltage | V_{ROL} | $R_{Ext} > 5.0\text{ k}$, $V_{OUT} > 1.0\text{ V}$ | - | 0.20 | 0.40 | V |
| Reset Output Leakage Current | I_{ROH} | $V_{ROH} = 5.0\text{ V}$ | - | 0 | 10 | μA |
| Reset Charging Current | $I_{D,C}$ | $V_D = 1.0\text{ V}$ | 2.0 | 4.0 | 6.5 | μA |
| Upper Timing Threshold | V_{DU} | - | 1.2 | 1.3 | 1.4 | V |
| Reset Delay Time | t_{rd} | $C_D = 47\text{ nF}$ | 10 | 16 | 22 | ms |
| Reset Reaction Time | t_{rr} | $C_D = 47\text{ nF}$ | | 1.5 | 4.0 | μs |

PROTECTION

| | | | | | | |
|-----------------------------|----------------|-----------------------------------|-----|--|-----|------------------|
| Current Limit | $I_{OUT(LIM)}$ | $V_{OUT} = 4.5\text{ V}$ (Note 6) | 150 | | 500 | mA |
| Short Circuit Current Limit | $I_{OUT(SC)}$ | $V_{OUT} = 0\text{ V}$ (Note 6) | 100 | | 500 | mA |
| Thermal shutdown threshold | T_{TSD} | (Note 8) | 150 | | 200 | $^\circ\text{C}$ |

6. Use pulse loading to limit power dissipation.

7. Dropout voltage = $(V_{IN} - V_{OUT})$, measured when the output voltage has dropped 100 mV relative to the nominal value obtained with $V_{IN} = 13.5\text{ V}$.

8. Not tested in production. Limits are guaranteed by design.

NCV8665

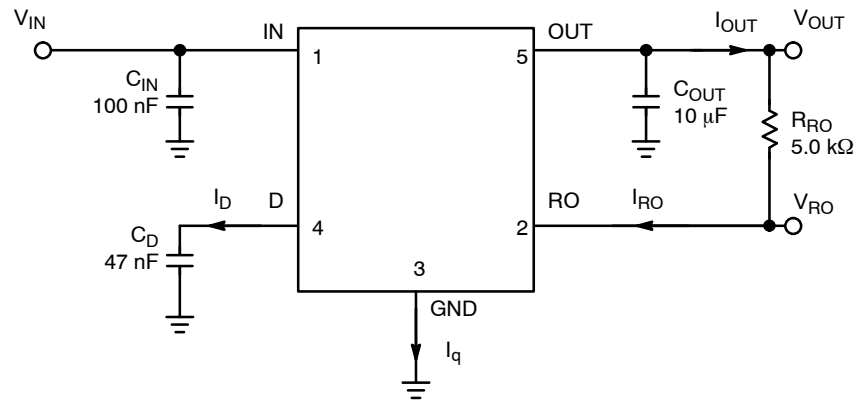


Figure 2. Application Circuit

TYPICAL CHARACTERISTIC CURVES

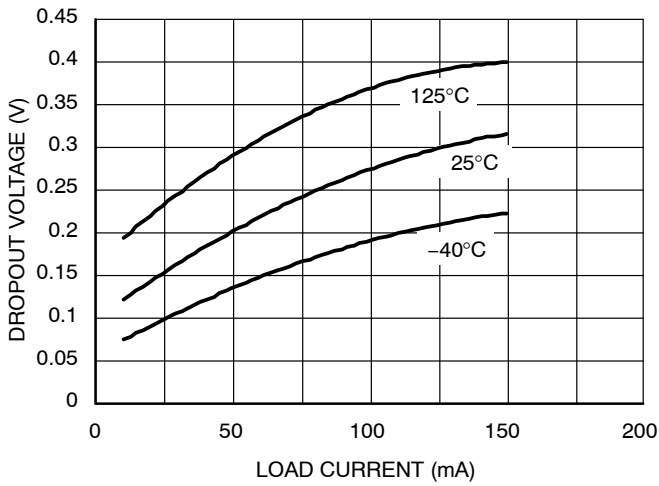


Figure 3. NCV8665 Dropout Voltage vs. Load Current

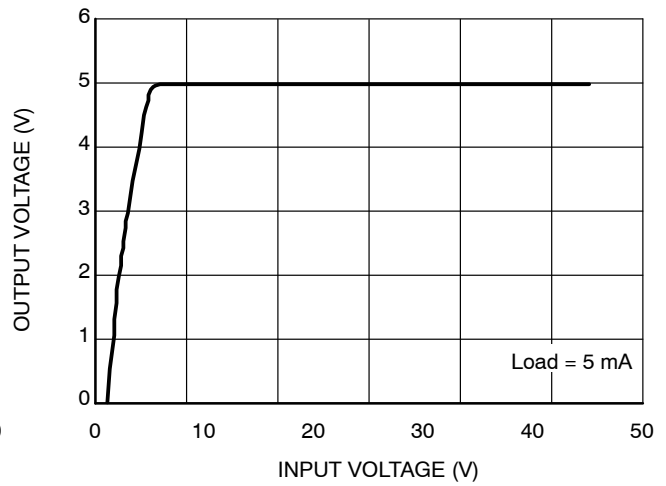


Figure 4. NCV8665 Input Voltage vs. Output Voltage (Full Range)

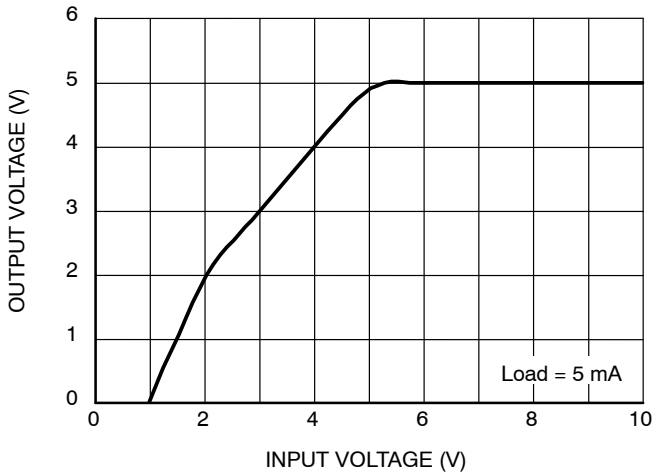


Figure 5. NCV8665 Input Voltage vs. Output Voltage (Low Voltage)

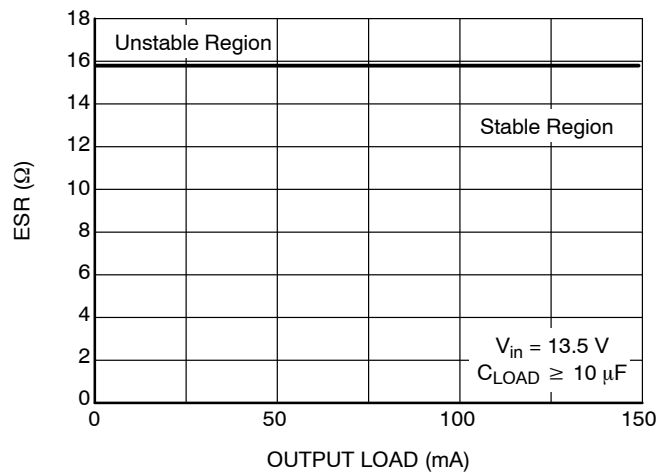


Figure 6. NCV8665 Stability Curve

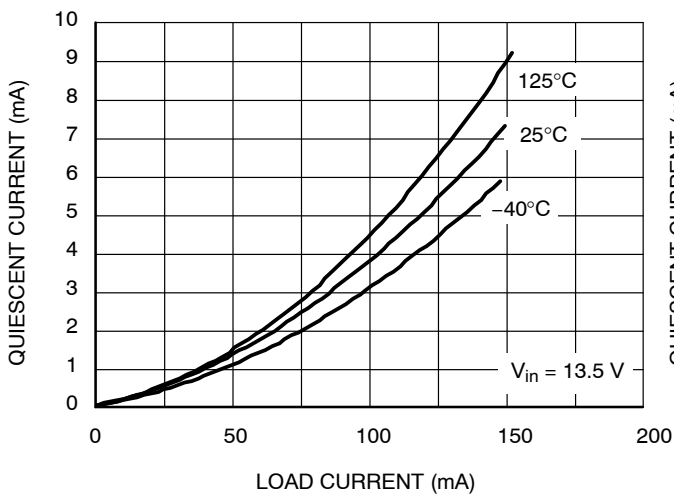


Figure 7. NCV8665 Quiescent Current vs. Load Current (Full Range)

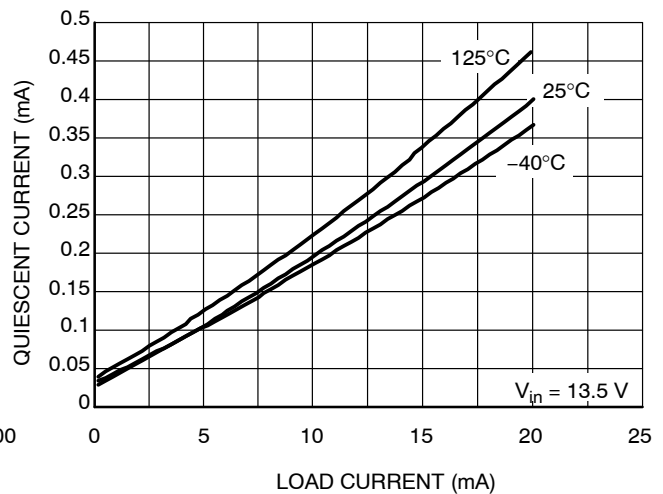


Figure 8. NCV8665 Quiescent Current vs. Load Current (Light Load)

TYPICAL CHARACTERISTIC CURVES

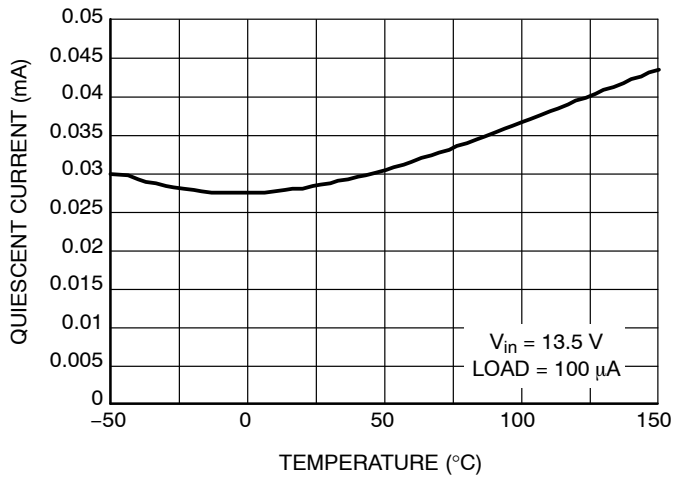


Figure 9. NCV8665 Quiescent Current vs. Temperature

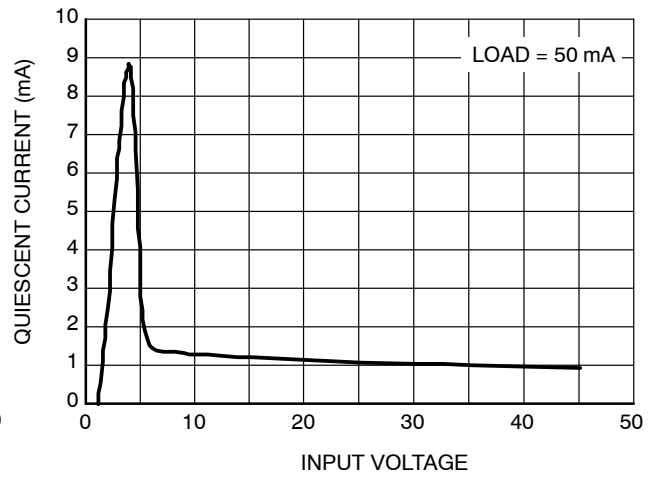


Figure 10. NCV8665 Quiescent Current vs. Input Voltage

Circuit Description

The NCV8665 is an integrated low dropout regulator that provides 5.0 V, 150 mA protected output and a signal for power on reset. The regulation is provided by a PNP pass transistor controlled by an error amplifier with a bandgap reference, which gives it the lowest possible drop out voltage and best possible temperature stability. The output current capability is 150 mA, and the base drive quiescent current is controlled to prevent over saturation when the input voltage is low or when the output is overloaded. The regulator is protected by both current limit and thermal shutdown. Thermal shutdown occurs above 150°C to protect the IC during overloads and extreme ambient temperatures. The delay time for the reset output is adjustable by selection of the timing capacitor. See Figure 2, Application Circuit, for circuit element nomenclature illustration.

Regulator

The error amplifier compares the reference voltage to a sample of the output voltage (V_{OUT}) and drives the base of a PNP series pass transistor by a buffer. The reference is a bandgap design to give it a temperature-stable output. Saturation control of the PNP is a function of the load current and input voltage. Oversaturation of the output power device is prevented, and quiescent current in the ground pin is minimized.

Regulator Stability Considerations

The input capacitor (C_{IN}) is necessary to stabilize the input impedance to avoid voltage line influences. The output capacitor helps determine three main characteristics of a linear regulator: startup delay, load transient response and loop stability. The capacitor value and type should be based on cost, availability, size and temperature constraints.

Tantalum, aluminum electrolytic, film, or ceramic capacitors are all acceptable solutions, however attention must be paid to ESR constraints. The aluminum electrolytic capacitor is the least expensive solution, but, if the circuit operates at low temperatures (-25°C to -40°C), both the capacitance and ESR of the capacitor will vary considerably. The capacitor manufacturer's data sheet usually provides this information. The value for the output capacitor C_{OUT} shown in Figure 2, Application Circuit, should work for most applications; however, it is not necessarily the optimized solution.

Reset Output

The reset output is used as the power on indicator to the microcontroller. This signal indicates when the output voltage is suitable for reliable operation of the controller. It pulls low when the output is not considered to be ready. RO is pulled up to V_{OUT} by an external resistor, typically 5.0 k Ω in value. The input and output conditions that control the Reset Output and the relative timing are illustrated in Figure 11, Reset Timing. Output voltage regulation must be maintained for the delay time before the reset output signals a valid condition. The delay for the reset output is defined as the amount of time it takes the timing capacitor on the delay pin to charge from a residual voltage of 0 V to the upper timing threshold voltage V_{DU} of 1.8 V. The charging current for this is I_D of 5.5 μA . By using typical IC parameters with a 47 nF capacitor on the D Pin, the following time delay is derived:

$$t_{RD} = C_D * V_{DU} / I_D$$

$$t_{RD} = 47 \text{ nF} * (1.8 \text{ V}) / 5.5 \mu\text{A} = 15.4 \text{ ms}$$

Other time delays can be obtained by changing the C_D capacitor value.

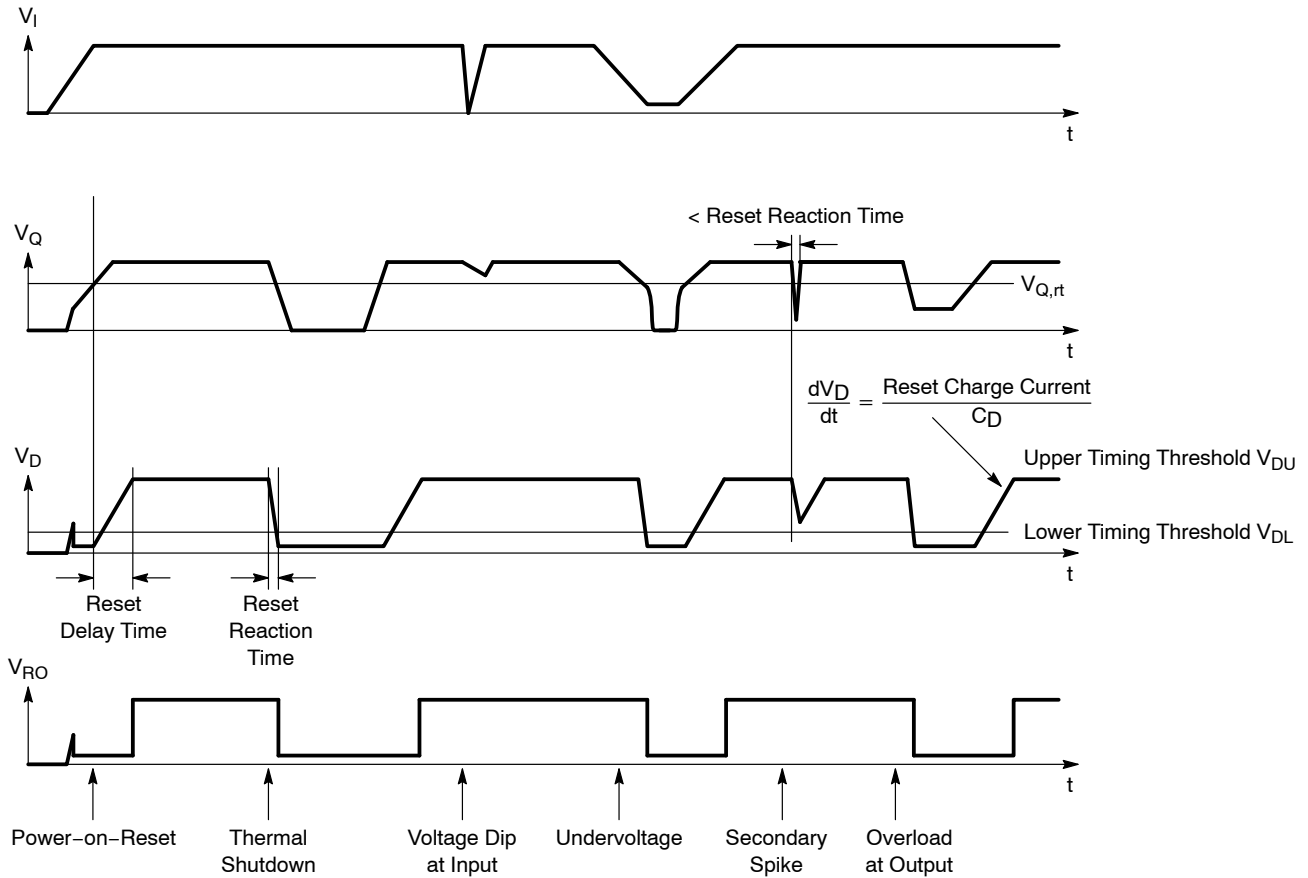


Figure 11. Reset Timing

Calculating Power Dissipation in a Single Output Linear Regulator

The maximum power dissipation for a single output regulator (Figure 12) is:

$$PD(max) = [V_{I(max)} - V_{Q(min)}]I_{Q(max)} + V_{I(max)}I_q \quad (1)$$

where

$V_{I(max)}$ is the maximum input voltage,

$V_{Q(min)}$ is the minimum output voltage,

$I_{Q(max)}$ is the maximum output current for the application,

I_q is the quiescent current the regulator consumes at $I_{Q(max)}$.

Once the value of $P_{D(max)}$ is known, the maximum permissible value of $R_{\theta JA}$ can be calculated:

$$R_{\theta JA} = \frac{150^{\circ}C - T_A}{P_D} \quad (2)$$

The value of $R_{\theta JA}$ can then be compared with those in the package section of the data sheet. Those packages with $R_{\theta JA}$'s less than the calculated value in Equation NO TAG will keep the die temperature below 150°C.

In some cases, none of the packages will be sufficient to dissipate the heat generated by the IC, and an external heatsink will be required.

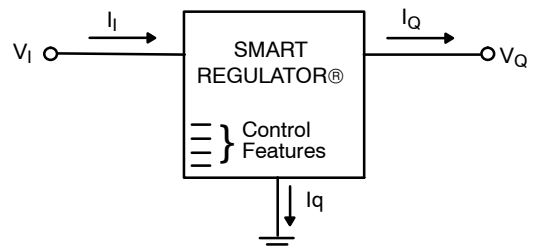


Figure 12. Single Output Regulator with Key Performance Parameters Labeled

Heatsinks

A heatsink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment will have a thermal resistance. Like series electrical resistances, these resistances are summed to determine the value of $R_{\theta JA}$:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CS} + R_{\theta SA} \quad (3)$$

where

$R_{\theta JC}$ is the junction-to-case thermal resistance,

$R_{\theta CS}$ is the case-to-heatsink thermal resistance,

$R_{\theta SA}$ is the heatsink-to-ambient thermal resistance.

$R_{\theta JC}$ appears in the package section of the data sheet. Like $R_{\theta JA}$, it too is a function of package type. $R_{\theta CS}$ and $R_{\theta SA}$ are

functions of the package type, heatsink and the interface between them. These values appear in heatsink data sheets of heatsink manufacturers.

Thermal, mounting, and heatsinking considerations are discussed in the ON Semiconductor application note AN1040/D.

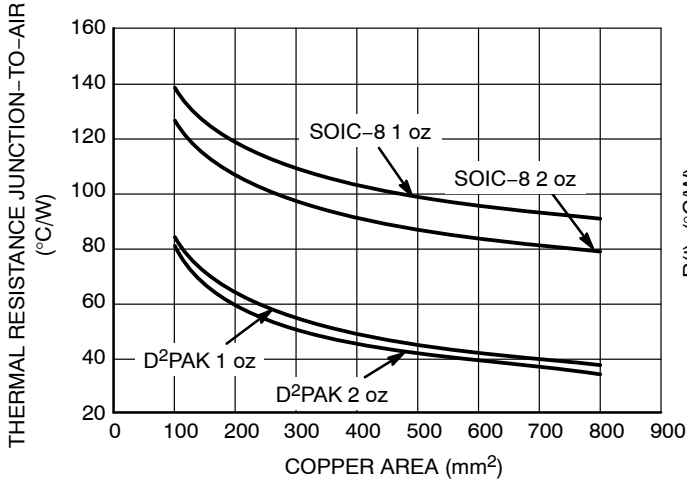


Figure 13. Thermal Resistance vs. PCB Area

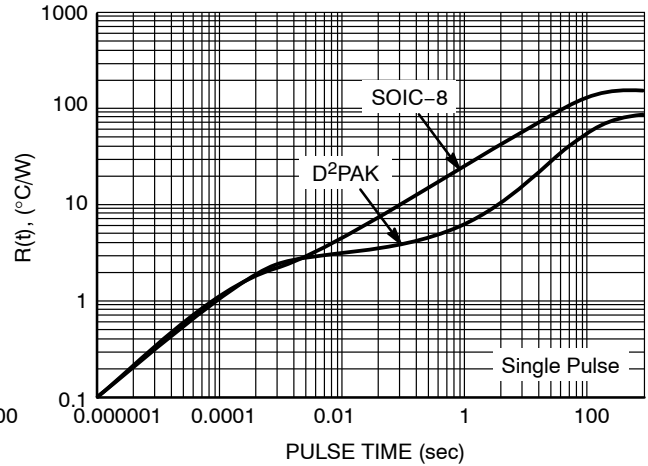


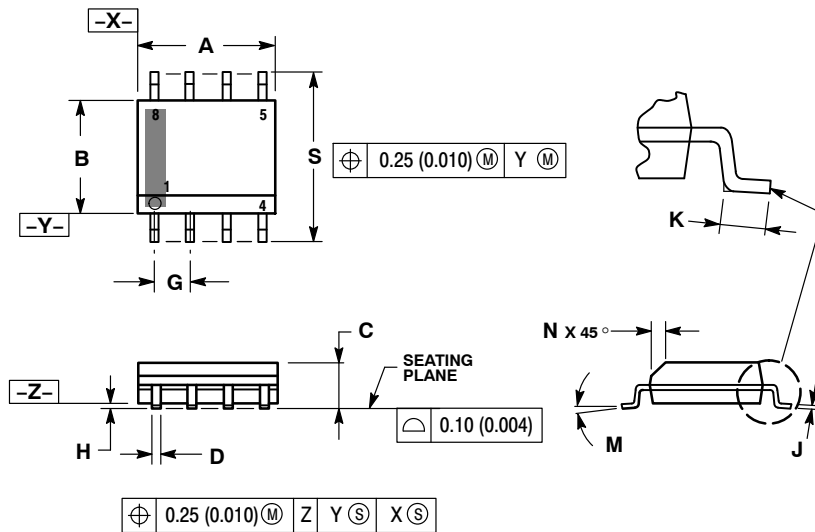
Figure 14. NCV8675 @ PCB Cu Area 100 mm²
PCB Cu thk 1 oz

ORDERING INFORMATION

| Device | Package | Shipping [†] |
|----------------|---------------------|-----------------------|
| NCV8665DS50G | D2PAK (Pb-Free) | 50 Units / Rail |
| NCV8665DS50R4G | D2PAK (Pb-Free) | 800 / Tape & Reel |
| NCV8665D50G | SOIC-8 (Pb-Free) | 98 Units / Rail |
| NCV8665D50R2G | SOIC-8 (Pb-Free) | 2500 / Tape & Reel |

[†]For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

PACKAGE DIMENSIONS

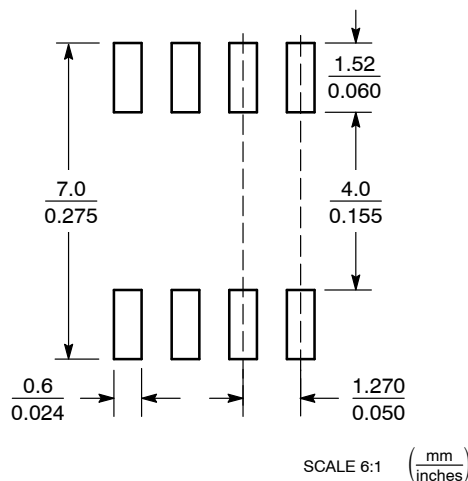
SOIC-8 NB
CASE 751-07
ISSUE AJ

NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

| DIM | MILLIMETERS | | INCHES | |
|-----|-------------|------|-----------|-------|
| | MIN | MAX | MIN | MAX |
| A | 4.80 | 5.00 | 0.189 | 0.197 |
| B | 3.80 | 4.00 | 0.150 | 0.157 |
| C | 1.35 | 1.75 | 0.053 | 0.069 |
| D | 0.33 | 0.51 | 0.013 | 0.020 |
| G | 1.27 BSC | | 0.050 BSC | |
| H | 0.10 | 0.25 | 0.004 | 0.010 |
| J | 0.19 | 0.25 | 0.007 | 0.010 |
| K | 0.40 | 1.27 | 0.016 | 0.050 |
| M | 0° | 8° | 0° | 8° |
| N | 0.25 | 0.50 | 0.010 | 0.020 |
| S | 5.80 | 6.20 | 0.228 | 0.244 |

SOLDERING FOOTPRINT*



*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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